

# ANTREX Electronics

Electronic Components Sourcing Information

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## Product Reference Information

**Part Number:** MCP48CMD12-E/MG  
**Manufacturer:** Microchip Technology  
**Package:** 16-VFQFN Exposed Pad  
**Availability:** Please confirm with sales

### Product Page:

<https://antrexco.com/product/details/microchip-technology/mcp48cmd12-e-mg.html>

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### Note:

This PDF includes an ANTREX product reference cover page.

The original manufacturer datasheet follows from the next page.

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## 8/10/12-Bit Digital-to-Analog Converters, 1 LSb INL, Single/Dual Voltage Outputs with SPI Interface

### Features

- Memory Options:
  - Volatile memory: MCP48CVDX1/2
  - Nonvolatile memory: MCP48CMDX1/2
- Operating Voltage Range:
  - 2.7V to 5.5V – Full specifications
  - 1.8V to 2.7V – Reduced device specifications
- Output Voltage Resolutions:
  - 8-bit: MCP48CXD0X (256 steps)
  - 10-bit: MCP48CXD1X (1024 steps)
  - 12-bit: MCP48CXD2X (4096 steps)
- Nonvolatile Memory (MTP) Size: 32 Locations
- 1 LSb Integral Nonlinearity (INL) Specification
- DAC Voltage Reference Source Options:
  - Device  $V_{DD}$
  - External  $V_{REF}$  pin (buffered or unbuffered)
  - Internal band gap (1.214V typical)
- Output Gain Options:
  - 1x (unity)
  - 2x (available when not using internal  $V_{DD}$  as voltage source)
- Power-on/Brown-out Reset (POR/BOR) Protection
- Power-Down Modes:
  - Disconnects output buffer (high-impedance)
  - Selection of  $V_{OUT}$  pull-down resistors (100 k $\Omega$  or 1 k $\Omega$ )
- SPI Interface:
  - Supports '00' and '11' modes
  - 50 MHz write speed
  - 25 MHz read speed
- Package Types:
  - Dual: 16-lead 3 x 3 QFN, 10-lead MSOP, 10-lead 3 x 3 DFN
  - Single: 16-lead 3 x 3 QFN, 10-lead MSOP, 10-lead 3 x 3 DFN
- Extended Temperature Range: -40°C to +125°C

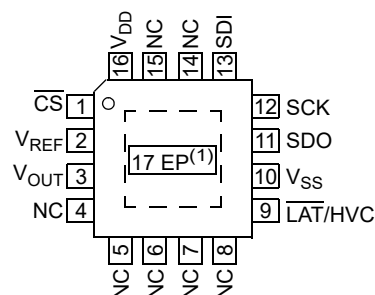
### Package Types

#### MCP48CXX1 (Single)

##### MSOP-10, DFN-10 (3 x 3)

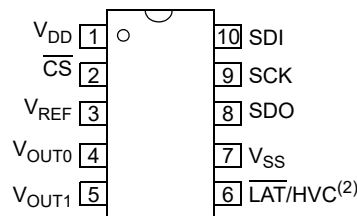


##### QFN-16 (3 x 3)

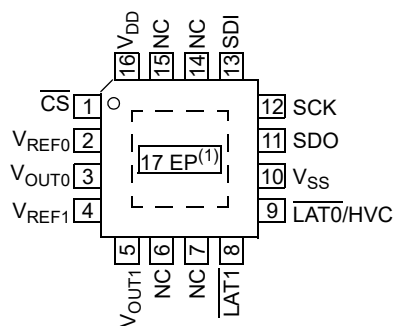


#### MCP48CXX2 (Dual)

##### MSOP-10, DFN-10 (3 x 3)



##### QFN-16 (3 x 3)



**Note 1:** Exposed pad (substrate paddle).

**Note 2:** This pin's signal can be connected to DAC0 and/or DAC1.

# MCP48CXDX1/2

## General Description

The MCP48CXDX1/2 are single and dual channel 8-bit, 10-bit and 12-bit buffered voltage output Digital-to-Analog Converters (DAC), with volatile or MTP memory and an SPI serial interface.

The MTP memory can be written by the user up to 32 times, for each specific register. It requires a high-voltage level on the HVC pin, typically 7.5V, in order to successfully program the desired memory location. The nonvolatile memory includes power-up output values, device configuration registers and general purpose memory.

The  $V_{REF}$  pin, the device  $V_{DD}$  or the internal band gap voltage can be selected as the DAC's reference voltage. When  $V_{DD}$  is selected,  $V_{DD}$  is internally connected to the DAC reference circuit.

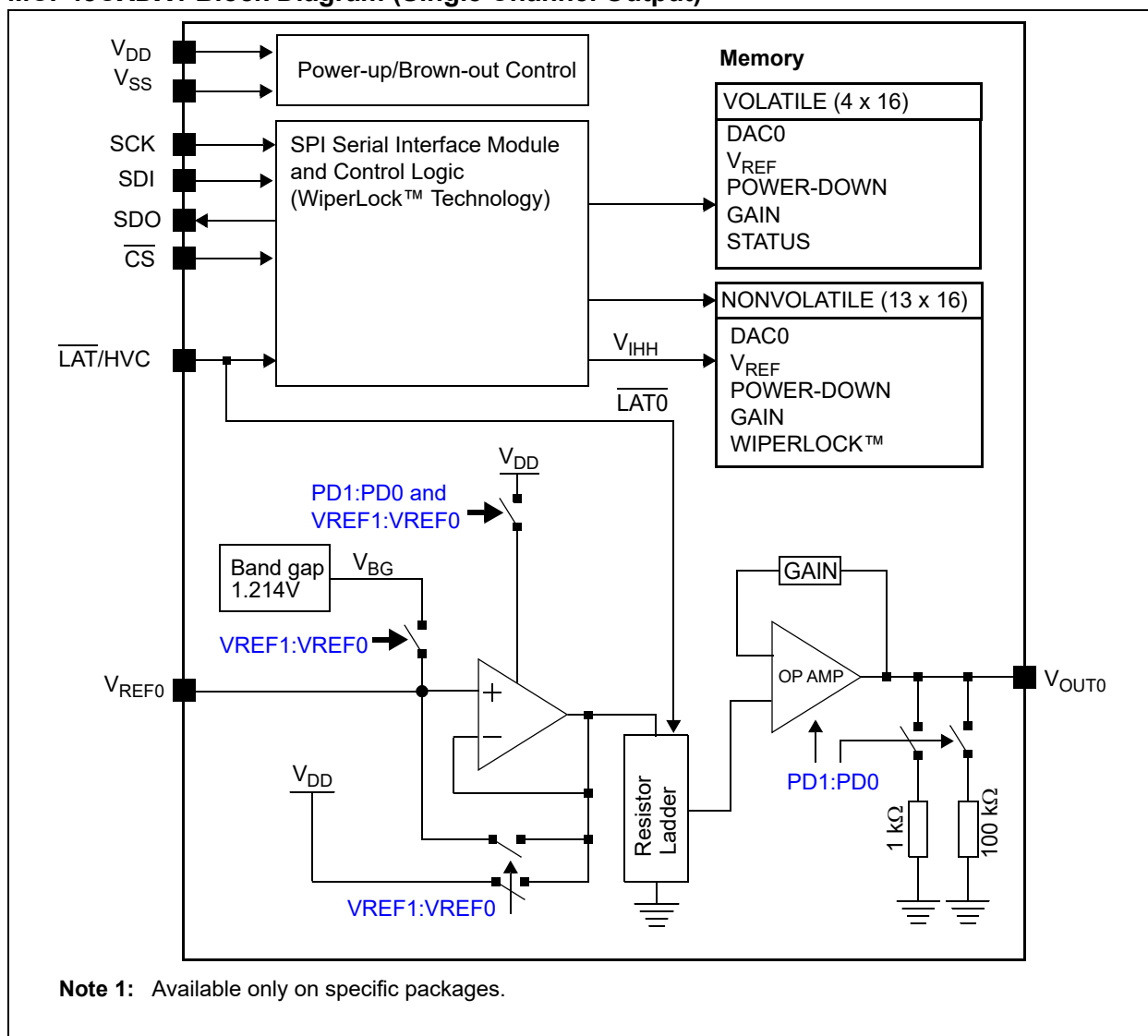
When the  $V_{REF}$  pin is used with an external voltage reference, the user can select between a gain of 1 or 2 and can have the reference buffer enabled or disabled. When the gain is 2, the  $V_{REF}$  pin voltage must be limited to a maximum of  $V_{DD}/2$ .

These devices have a four-wire SPI-compatible serial interface with speeds up to 50 MHz for write and 25 MHz for read operations.

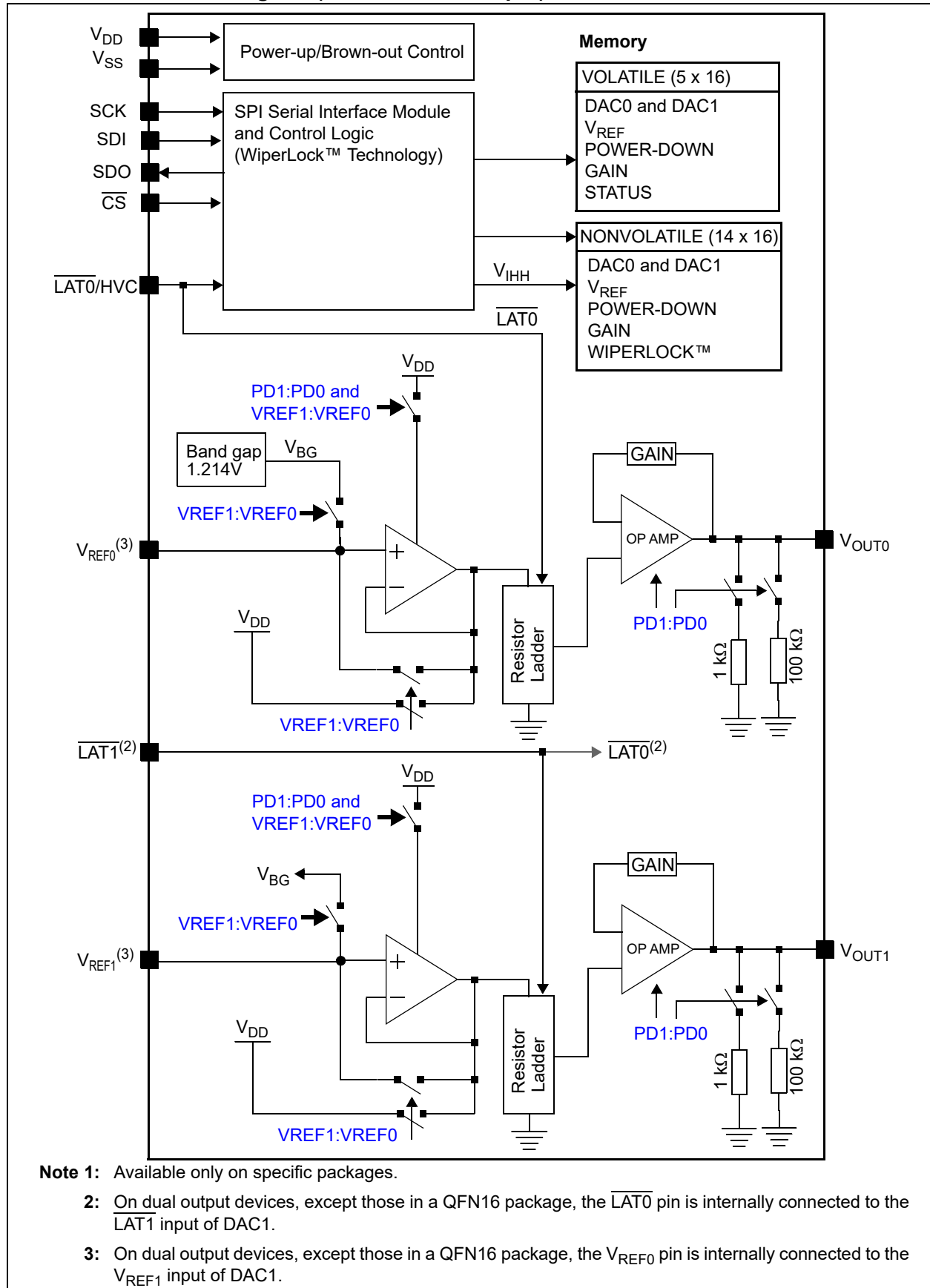
## Applications

- Set Point or Offset Trimming
- Sensor Calibration
- Low-Power Portable Instrumentation
- PC Peripherals
- Data Acquisition Systems

## MCP48CXDX1 Block Diagram (Single Channel Output)



**MCP48CXDX2 Block Diagram (Dual Channel Output)**



# MCP48CXDX1/2

## Family Device Features

| Device     | Package Type   | # of Channels | Resolution (bits) | DAC Output<br>POR/BOR<br>Setting <sup>(1)</sup> | # of V <sub>REF</sub> Inputs | # of $\overline{\text{LAT}}$ Inputs <sup>(3)</sup> | Memory <sup>(2)</sup> | GP MTP Locations |
|------------|----------------|---------------|-------------------|-------------------------------------------------|------------------------------|----------------------------------------------------|-----------------------|------------------|
| MCP48CVD01 | MSOP, QFN, DFN | 1             | 8                 | 7Fh                                             | 1                            | 1                                                  | RAM                   | —                |
| MCP48CVD11 | MSOP, QFN, DFN | 1             | 10                | 1FFh                                            | 1                            | 1                                                  | RAM                   | —                |
| MCP48CVD21 | MSOP, QFN, DFN | 1             | 12                | 7FFh                                            | 1                            | 1                                                  | RAM                   | —                |
| MCP48CVD02 | QFN            | 2             | 8                 | 7Fh                                             | 2                            | 2                                                  | RAM                   | —                |
|            | MSOP, DFN      | 2             | 8                 | 7Fh                                             | 1                            | 1                                                  | RAM                   | —                |
| MCP48CVD12 | QFN            | 2             | 10                | 1FFh                                            | 2                            | 2                                                  | RAM                   | —                |
|            | MSOP, DFN      | 2             | 10                | 1FFh                                            | 1                            | 1                                                  | RAM                   | —                |
| MCP48CVD22 | QFN            | 2             | 12                | 7FFh                                            | 2                            | 2                                                  | RAM                   | —                |
|            | MSOP, DFN      | 2             | 12                | 7FFh                                            | 1                            | 1                                                  | RAM                   | —                |
| MCP48CMD01 | MSOP, QFN, DFN | 1             | 8                 | 7Fh                                             | 1                            | 1                                                  | MTP                   | 8                |
| MCP48CMD11 | MSOP, QFN, DFN | 1             | 10                | 1FFh                                            | 1                            | 1                                                  | MTP                   | 8                |
| MCP48CMD21 | MSOP, QFN, DFN | 1             | 12                | 7FFh                                            | 1                            | 1                                                  | MTP                   | 8                |
| MCP48CMD02 | QFN            | 2             | 8                 | 7Fh                                             | 2                            | 2                                                  | MTP                   | 8                |
|            | MSOP, DFN      | 2             | 8                 | 7Fh                                             | 1                            | 1                                                  | MTP                   | 8                |
| MCP48CMD12 | QFN            | 2             | 10                | 1FFh                                            | 2                            | 2                                                  | MTP                   | 8                |
|            | MSOP, DFN      | 2             | 10                | 1FFh                                            | 1                            | 1                                                  | MTP                   | 8                |
| MCP48CMD22 | QFN            | 2             | 12                | 7FFh                                            | 2                            | 2                                                  | MTP                   | 8                |
|            | MSOP, DFN      | 2             | 12                | 7FFh                                            | 1                            | 1                                                  | MTP                   | 8                |

**Note 1:** The factory default value.

**2:** Each nonvolatile memory location can be written 32 times. For subsequent writes to the MTP, the device will ignore the commands and the memory will not be modified.

**3:** If the product is a dual device and the package has only one  $\overline{\text{LAT}}$  pin, it is associated with both DAC0 and DAC1.

## 1.0 ELECTRICAL CHARACTERISTICS

### Absolute Maximum Ratings<sup>(†)</sup>

|                                                                                         |                        |
|-----------------------------------------------------------------------------------------|------------------------|
| Voltage on $V_{DD}$ with Respect to $V_{SS}$                                            | -0.6V to +6.5V         |
| Voltage on all Pins with Respect to $V_{SS}$                                            | -0.6V to $V_{DD}+0.3V$ |
| Input Clamp Current, $I_{IK}$ ( $V_I < 0$ , $V_I > V_{DD}$ , $V_I > V_{PP}$ on HV Pins) | $\pm 20$ mA            |
| Output Clamp Current, $I_{OK}$ ( $V_O < 0$ or $V_O > V_{DD}$ )                          | $\pm 20$ mA            |
| Maximum Current out of the $V_{SS}$ Pin (Single)                                        | 50 mA                  |
| (Dual)                                                                                  | 100 mA                 |
| Maximum Current into the $V_{DD}$ Pin (Single)                                          | 50 mA                  |
| (Dual)                                                                                  | 100 mA                 |
| Maximum Current Sourced by the $V_{OUT}$ Pin                                            | 20 mA                  |
| Maximum Current Sunk by the $V_{OUT}$ Pin                                               | 20 mA                  |
| Maximum Current Sourced/Sunk by the $V_{REF(0)}$ Pin (in Band Gap mode)                 | 20 mA                  |
| Maximum Current Sunk by the $V_{REFx}$ Pin (when $V_{REF}$ is in Unbuffered mode)       | 175 $\mu$ A            |
| Maximum Current Sourced by the $V_{REFx}$ Pin                                           | 20 $\mu$ A             |
| Maximum Current Sunk by the $V_{REF}$ Pin                                               | 125 $\mu$ A            |
| Maximum Output Current Sunk by SDO Output Pin                                           | 25 mA                  |
| Maximum Output Current Sourced by SDO Output Pin                                        | 25 mA                  |
| Total Power Dissipation <sup>(1)</sup>                                                  | 400 mW                 |
| ESD Protection on all Pins                                                              | $\geq \pm 6$ kV (HBM)  |
|                                                                                         | $\geq \pm 400V$ (MM)   |
|                                                                                         | $\geq \pm 2$ kV (CDM)  |
| Latch-Up (per JEDEC <sup>®</sup> JESD78A) at +125°C                                     | $\pm 100$ mA           |
| Storage Temperature                                                                     | -65°C to +150°C        |
| Ambient Temperature with Power Applied                                                  | -55°C to +125°C        |
| Soldering Temperature of Leads (10 seconds)                                             | +300°C                 |
| Maximum Junction Temperature ( $T_J$ )                                                  | +150°C                 |

**† Notice:** Stresses above those listed under “Maximum Ratings” may cause permanent damage to the device. This is a stress rating only and functional operation of the device at those or any other conditions above those indicated in the operational listings of this specification is not implied. Exposure to maximum rating conditions for extended periods may affect device reliability.

**Note 1:** Power dissipation is calculated as follows:

$$P_{DIS} = V_{DD} \times \{I_{DD} - \sum I_{OH}\} + \sum \{(V_{DD} - V_{OH}) \times I_{OH}\} + \sum (V_{OL} \times I_{OL})$$

# MCP48CXDX1/2

## DC CHARACTERISTICS

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,  $R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                  | Sym.         | Min.      | Typ. | Max. | Units         | Conditions                                                                                                                                   |
|-------------------------------------------------------------|--------------|-----------|------|------|---------------|----------------------------------------------------------------------------------------------------------------------------------------------|
| Supply Voltage                                              | $V_{DD}$     | 2.7       | —    | 5.5  | V             | —                                                                                                                                            |
|                                                             |              | 1.8       | —    | 2.7  | V             | DAC operation (reduced analog specifications) and serial interface                                                                           |
| $V_{DD}$ Voltage (rising) to Ensure Device Power-on Reset   | $V_{POR}$    | —         | —    | 1.75 | V             | RAM retention voltage ( $V_{RAM}$ ) < $V_{POR}$<br>$V_{DD}$ voltages greater than the $V_{POR}$ limit ensure that the device is out of reset |
| $V_{DD}$ Voltage (falling) to Ensure Device Brown-out Reset | $V_{BOR}$    | $V_{RAM}$ | —    | 1.61 | V             | RAM retention voltage ( $V_{RAM}$ ) < $V_{BOR}$                                                                                              |
| $V_{DD}$ Rise Rate to Ensure Power-on Reset                 | $V_{DDRR}$   | Note 3    |      |      | V/ms          | —                                                                                                                                            |
| Power-on Reset to Output-Driven Delay (Note 1)              | $T_{POR2OD}$ | —         | —    | 130  | $\mu\text{s}$ | $V_{DD}$ rising, $V_{DD} > V_{POR}$<br>single output                                                                                         |
|                                                             |              | —         | —    | 145  | $\mu\text{s}$ | $V_{DD}$ rising, $V_{DD} > V_{POR}$<br>dual output                                                                                           |

**Note 1** This parameter is ensured by design.

**Note 3** POR/BOR voltage trip point is not slope dependent. Hysteresis implemented with time delay.

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                               | Sym.                    | Min. | Typ. | Max. | Units | Conditions |                                                                                                                                                                                                                                |                                                                                                                                                                                  |
|------------------------------------------|-------------------------|------|------|------|-------|------------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| Supply Current                           | I <sub>DD</sub>         | —    | —    | 190  | μA    | Single     | 1 MHz                                                                                                                                                                                                                          | Serial Interface Active<br>VRxB:VRxA = 10 <sup>(4)</sup> ,<br>V <sub>OUT</sub> is unloaded,<br>V <sub>REF</sub> = V <sub>DD</sub> = 5.5V<br>Volatile DAC<br>register = mid-scale |
|                                          |                         | —    | —    | 530  |       |            | 10 MHz <sup>(2)</sup>                                                                                                                                                                                                          |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 1270 |       |            | 50 MHz <sup>(2)</sup>                                                                                                                                                                                                          |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 290  | μA    | Dual       | 1 MHz                                                                                                                                                                                                                          |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 630  |       |            | 10 MHz <sup>(2)</sup>                                                                                                                                                                                                          |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 1370 |       |            | 50 MHz <sup>(2)</sup>                                                                                                                                                                                                          |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 140  | μA    | Single     | Serial Interface Inactive<br>VRxB:VRxA = 10, V <sub>REF</sub> = V <sub>DD</sub> = 5.5V<br>SCK = SDI = V <sub>SS</sub> ,<br>V <sub>OUT</sub> is unloaded,<br>Volatile DAC register = mid-scale                                  |                                                                                                                                                                                  |
|                                          |                         | —    | —    | 240  | μA    | Dual       |                                                                                                                                                                                                                                |                                                                                                                                                                                  |
| LAT/HVC Pin Write Current <sup>(1)</sup> | I <sub>DD(MTP_WR)</sub> | —    | —    | 6.40 | mA    | —          | Serial Interface Inactive<br>(MTP Write Active),<br>VRxB:VRxA = 10 (valid for all modes)<br>V <sub>DD</sub> = 5.5V, LAT/HVC = V <sub>IHH</sub> ,<br>Write all '1's to nonvolatile DAC0,<br>V <sub>OUT</sub> pins are unloaded. |                                                                                                                                                                                  |
| Power-Down Current                       | I <sub>DDP</sub>        | —    | 0.5  | 2.4  | μA    | —          | PDxB:PDxA = 01 <sup>(5)</sup> ,<br>VRxB:VRxA = 10,<br>V <sub>OUT</sub> not connected                                                                                                                                           |                                                                                                                                                                                  |

**Note 1** This parameter is ensured by design.

**Note 2** This parameter is ensured by characterization.

**Note 4** Supply current is independent of current through the resistor ladder in mode  $VRxB:VRxA = 10$ .

**Note 5** The  $PDxB:PDxA = 01$ ,  $10$  and  $11$  configurations must have the same current.

# MCP48CXDX1/2

## DC CHARACTERISTICS (CONTINUED)

**Standard Operating Conditions (unless otherwise specified):**

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                                           | Sym.                                    | Min.     | Typ.  | Max.     | Units                   | Conditions                                                                 |
|------------------------------------------------------------------------------------------------------|-----------------------------------------|----------|-------|----------|-------------------------|----------------------------------------------------------------------------|
| Resistor Ladder Resistance <sup>(6)</sup>                                                            | $R_L$                                   | 62.475   | 73.5  | 84.525   | $\text{k}\Omega$        | $VRxB:VRxA = 10$ ,<br>$V_{REF} = V_{DD}$                                   |
| Resolution<br>(# of Resistors and<br># of Taps)<br>(see <b>B.1 “Resolution”</b> )                    | N                                       | 256      |       |          | Taps                    | 8-bit No Missing Codes                                                     |
|                                                                                                      |                                         | 1024     |       |          | Taps                    | 10-bit No Missing Codes                                                    |
|                                                                                                      |                                         | 4096     |       |          | Taps                    | 12-bit No Missing Codes                                                    |
| Nominal $V_{OUT}$ Match <sup>(10)</sup>                                                              | $ V_{OUT} - V_{OUTMEAN}  / V_{OUTMEAN}$ | —        | 0.026 | 0.3      | %                       | $1.8\text{V} \leq V_{DD} \leq 5.5\text{V}$ <sup>(2)</sup>                  |
| $V_{OUT}$ Tempco <sup>(2)</sup><br>(see <b>B.19 “<math>V_{OUT}</math> Temperature Coefficient”</b> ) | $\Delta V_{OUT} / \Delta T$             | —        | 3     | —        | ppm/ $^{\circ}\text{C}$ | Code = mid-scale<br>(7Fh, 1FFh or 7FFh),<br>$VRxB:VRxA =$<br>00, 10 and 11 |
| $V_{REF}$ Pin<br>Input Voltage Range                                                                 | $V_{REF}$                               | $V_{SS}$ | —     | $V_{DD}$ | V                       | $1.8\text{V} \leq V_{DD} \leq 5.5\text{V}$ <sup>(1)</sup>                  |

**Note 1** This parameter is ensured by design.

**Note 2** This parameter is ensured by characterization.

**Note 6** Resistance is defined as the resistance between the  $V_{REF}$  pin (mode  $VRxB:VRxA = 10$ ) and the  $V_{SS}$  pin. For dual channel devices (MCP48CXDX2), this is the effective resistance of each resistor ladder. The resistance measurement is one of the two resistor ladders measured in parallel.

**Note 10** Variation of one output voltage to mean output voltage for dual devices only.

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                                               | Sym.              | Min. | Typ. | Max.  | Units       | Conditions                                                                                              |                                                                                        |
|----------------------------------------------------------------------------------------------------------|-------------------|------|------|-------|-------------|---------------------------------------------------------------------------------------------------------|----------------------------------------------------------------------------------------|
| Zero-Scale Error<br>(Code = 000h)<br><br>(see B.5<br>"Zero-Scale<br>Error (E <sub>ZS</sub> )")           | E <sub>ZS</sub>   | —    | —    | 0.5   | LSb         | 8-bit                                                                                                   | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , No Load.                |
|                                                                                                          |                   | —    | —    | 2     | LSb         | 10-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load                 |
|                                                                                                          |                   | —    | —    | 8     | LSb         | 12-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load                 |
| Offset Error<br>(see B.7 "Offset<br>Error (E <sub>OS</sub> )")                                           | E <sub>OS</sub>   | -9   | ±1.1 | +9    | mV          | VRxB:VRxA = 10, G <sub>x</sub> = 0, no load<br>8-bit: Code = 4; 10-bit: Code = 16;<br>12-bit: Code = 64 |                                                                                        |
| Offset Voltage<br>Temperature<br>Coefficient <sup>(2, 9)</sup>                                           | V <sub>OSTC</sub> | —    | ±9   | —     | μV/°C       | —                                                                                                       |                                                                                        |
| Full-Scale Error<br><br>(see B.4<br>"Full-Scale Error<br>(E <sub>FS</sub> )")                            | E <sub>FS</sub>   | —    | —    | 0.625 | LSb         | 8-bit                                                                                                   | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load                 |
|                                                                                                          |                   | —    | —    | 2.5   | LSb         | 10-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load                 |
|                                                                                                          |                   | —    | —    | 10    | LSb         | 12-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load                 |
| Gain Error <sup>(7)</sup><br><br>(see B.9 "Gain<br>Error (E <sub>G</sub> )")                             | E <sub>G</sub>    | -0.6 | ±0.1 | +0.6  | % of<br>FSR | 8-bit                                                                                                   | VRxB:VRxA = 10, G = 0,<br>Code = 252,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load  |
|                                                                                                          |                   | -0.6 | ±0.1 | +0.6  | % of<br>FSR | 10-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>Code = 1008,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load |
|                                                                                                          |                   | -0.6 | ±0.1 | +0.6  | % of<br>FSR | 12-bit                                                                                                  | VRxB:VRxA = 10, G = 0,<br>Code = 4032,<br>V <sub>REF</sub> = V <sub>DD</sub> , no load |
| Gain-Error Drift <sup>(2)</sup><br>(see B.10 "Gain<br>Error Drift<br>(E <sub>GD</sub> )") <sup>(9)</sup> | ΔG/°C             | —    | -10  | —     | ppm/°C      | —                                                                                                       |                                                                                        |

**Note 2** This parameter is ensured by characterization.

**Note 7** This gain error does not include the offset error.

**Note 9** Code range dependent on resolution: 8-bit, codes 4 to 252; 10-bit, codes 16 to 1008; 12-bit, codes 64 to 4032.

# MCP48CXDX1/2

## DC CHARACTERISTICS (CONTINUED)

**Standard Operating Conditions (unless otherwise specified):**

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                           | Sym.  | Min.                                            | Typ. | Max.  | Units | Conditions |                                                                                                                         |
|--------------------------------------------------------------------------------------|-------|-------------------------------------------------|------|-------|-------|------------|-------------------------------------------------------------------------------------------------------------------------|
| Total Unadjusted Error<br>(2, 9)<br><br>(see B.6 “Total Unadjusted Error (ET)”)      | $E_T$ | -1.375                                          | —    | 0.625 | LSb   | 8-bit      | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | -5.5                                            | —    | 2.5   | LSb   | 10-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | -22                                             | —    | 10    | LSb   | 12-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | See Section 2.0<br>“Typical Performance Curves” |      |       | LSb   | 12-bit     | $VRxB:VRxA = 00$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$                        |
|                                                                                      |       |                                                 |      |       |       |            | $VRxB:VRxA = 01$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$ , $G = 0$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.) |
|                                                                                      |       |                                                 |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$ , $G = 1$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.)                                |
|                                                                                      |       |                                                 |      |       |       |            | $VRxB:VRxA = 10$<br>$VRxB:VRxA = 11$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$    |
|                                                                                      |       |                                                 |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$<br>$V_{REF} = 0.5 \times V_{DD}$ , $G = 1$                                            |
| Integral Nonlinearity <sup>(9)</sup><br><br>(see B.11 “Integral Nonlinearity (INL)”) | INL   | -0.10                                           | —    | +0.10 | LSb   | 8-bit      | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | -0.25                                           | —    | +0.25 | LSb   | 10-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | -1                                              | —    | +1    | LSb   | 12-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                      |       | See Section 2.0<br>“Typical Performance Curves” |      |       | LSb   | 12-bit     | $VRxB:VRxA = 00$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$                        |
|                                                                                      |       |                                                 |      |       |       |            | $VRxB:VRxA = 01$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$ , $G = 0$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.) |
|                                                                                      |       |                                                 |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$ , $G = 1$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.)                                |
|                                                                                      |       |                                                 |      |       |       |            | $VRxB:VRxA = 10$<br>$VRxB:VRxA = 11$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$    |
|                                                                                      |       |                                                 |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$<br>$V_{REF} = 0.5 \times V_{DD}$ , $G = 1$                                            |

**Note 2** This parameter is ensured by characterization.

**Note 9** Code range dependent on resolution: 8-bit, codes 4 to 252; 10-bit, codes 16 to 1008; 12-bit, codes 64 to 4032.

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                                                   | Sym. | Min.                                                         | Typ. | Max.  | Units | Conditions |                                                                                                                         |
|--------------------------------------------------------------------------------------------------------------|------|--------------------------------------------------------------|------|-------|-------|------------|-------------------------------------------------------------------------------------------------------------------------|
| Differential Nonlinearity <sup>(9)</sup><br><br>(see <a href="#">B.12</a> “Differential Nonlinearity (DNL)”) | DNL  | -0.1                                                         | —    | +0.1  | LSb   | 8-bit      | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                                              |      | -0.25                                                        | —    | +0.25 | LSb   | 10-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                                              |      | -1.0                                                         | —    | +1.0  | LSb   | 12-bit     | $VRxB:VRxA = 10$ , $G = 0$ ,<br>$V_{REF} = V_{DD}$ , no load                                                            |
|                                                                                                              |      | See <a href="#">Section 2.0</a> “Typical Performance Curves” |      |       | LSb   | 12-bit     | $VRxB:VRxA = 00$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$                        |
|                                                                                                              |      |                                                              |      |       |       |            | $VRxB:VRxA = 01$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$ , $G = 0$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.) |
|                                                                                                              |      |                                                              |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$ , $G = 1$ ,<br>$V_{REF} = V_{BG} = 1.2\text{V}$ (typ.)                                |
|                                                                                                              |      |                                                              |      |       |       |            | $VRxB:VRxA = 10$<br>$VRxB:VRxA = 11$<br>No Load<br>$V_{DD} = 1.8\text{-}5.5\text{V}$<br>$V_{REF} = V_{DD}$ , $G = 0$    |
|                                                                                                              |      |                                                              |      |       |       |            | $V_{DD} = 2.7\text{-}5.5\text{V}$<br>$V_{REF} = 0.5 \times V_{DD}$ , $G = 1$                                            |

**Note 9** Code range dependent on resolution: 8-bit, codes 4 to 252; 10-bit, codes 16 to 1008; 12-bit, codes 64 to 4032.

# MCP48CXDX1/2

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                      | Sym.           | Min.  | Typ.  | Max.  | Units      | Conditions                                                 |                                           |
|-----------------------------------------------------------------|----------------|-------|-------|-------|------------|------------------------------------------------------------|-------------------------------------------|
| -3 dB Bandwidth<br>(see <a href="#">B.16</a> “-3 dB Bandwidth”) | BW             | —     | 270   | —     | kHz        | $V_{REF} = 3.0V \pm 2V$ ,<br>$VRxB:VRxA = 10$ , $Gx = 0$   |                                           |
|                                                                 |                | —     | 170   | —     | kHz        | $V_{REF} = 3.5V \pm 1.5V$ ,<br>$VRxB:VRxA = 10$ , $Gx = 1$ |                                           |
| Output Amplifier (Op Amp)                                       |                |       |       |       |            |                                                            |                                           |
| Phase Margin <sup>(1)</sup>                                     | PM             | —     | 76    | —     | °          | $R_L = \infty$                                             |                                           |
| Slew Rate                                                       | SR             | —     | 0.7   | —     | V/ $\mu$ s | $R_L = 2\text{ k}\Omega$                                   |                                           |
| Load Regulation                                                 | —              | —     | 147   | —     | $\mu$ V/mA | 1 mA < I < 6 mA                                            | $V_{DD} = 5.5V$ ,<br>DAC code = mid-scale |
|                                                                 |                | —     | 300   | —     | $\mu$ V/mA | -6 mA < I < -1 mA                                          |                                           |
| Short-Circuit Current                                           | $I_{SC\_OA}$   | 6     | 10    | 14    | mA         | Short to $V_{SS}$                                          | DAC code = full-scale                     |
|                                                                 |                | 6     | 10    | 14    | mA         | Short to $V_{DD}$                                          | DAC code = zero-scale                     |
| Settling Time <sup>(8)</sup>                                    | $t_{SETTLING}$ | —     | 4     | —     | $\mu$ s    | $R_L = 2\text{ k}\Omega$                                   |                                           |
| Internal Band Gap                                               |                |       |       |       |            |                                                            |                                           |
| Band Gap Voltage                                                | $V_{BG}$       | 1.180 | 1.214 | 1.260 | V          | $1.8 < V_{DD} < 5.5V$                                      |                                           |
| Short-Circuit Current                                           | $I_{SC\_BG}$   | 6     | 10    | 14    | mA         | Short to $V_{SS}$                                          |                                           |
|                                                                 |                | 6     | 10    | 14    | mA         | Short to $V_{DD}$                                          |                                           |
| Band Gap Voltage Temperature Coefficient                        | $V_{BGTC}$     | —     | 18    | —     | ppm/°C     | $1.8V \leq V_{DD} < 5.5V$                                  |                                           |
| Band Gap Mode $V_{REF}$ Pin Load Regulation                     | $I_{BG}$       | —     | 52    | —     | $\mu$ V/mA | 1 mA < I < 6 mA                                            | $V_{DD} = 5.5V$ ,<br>DAC code = mid-scale |
|                                                                 |                | —     | 374   | —     | $\mu$ V/mA | -6 mA < I < -1 mA                                          |                                           |

**Note 1** This parameter is ensured by design.

**Note 8** Within 1/2 LSB of the final value when code changes from 1/4 to 3/4 of FSR. (Example: 400h to C00h in a 12-bit device.)

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                              | Sym.       | Min.                                                         | Typ. | Max.     | Units         | Conditions                                                                                        |
|-----------------------------------------------------------------------------------------|------------|--------------------------------------------------------------|------|----------|---------------|---------------------------------------------------------------------------------------------------|
| <b>External Reference (<math>V_{REF}</math>)</b>                                        |            |                                                              |      |          |               |                                                                                                   |
| Input Range <sup>(1)</sup>                                                              | $V_{REF}$  | $V_{SS}$                                                     | —    | $V_{DD}$ | V             | $VRxB:VRxA = 10$<br>(Unbuffered mode)                                                             |
| Input Capacitance                                                                       | $C_{REF}$  | —                                                            | 29   | —        | pF            | $VRxB:VRxA = 10$<br>(Unbuffered mode)                                                             |
| Input Impedance                                                                         | $R_L$      | See <a href="#">Resistor Ladder Resistance<sup>(6)</sup></a> |      |          | k $\Omega$    | $2.7\text{V} \leq V_{DD} \leq 5.5\text{V}$<br>$VRxB:VRxA = 10$<br>$V_{REF} = V_{DD}$              |
| Current through $V_{REF}$ <sup>(1)</sup>                                                | $I_{VREF}$ | —                                                            | —    | 176.07   | $\mu\text{A}$ | Mathematically from $R_{VREF(min)}$ spec (at 5.5V)                                                |
| Total Harmonic Distortion <sup>(1)</sup>                                                | THD        | —                                                            | -76  | —        | dB            | $V_{REF} = 2.048\text{V} \pm 0.1\text{V}$ ,<br>$VRxB:VRxA = 10$ , $Gx = 0$ ,<br>Frequency = 1 kHz |
| <b>Dynamic Performance</b>                                                              |            |                                                              |      |          |               |                                                                                                   |
| Major Code Transition Glitch (see <a href="#">B.14 “Major-Code Transition Glitch”</a> ) | —          | —                                                            | 60   | —        | nV-s          | 1 LSb change around major carry (7FFh to 800h)                                                    |
| Digital Feedthrough (see <a href="#">B.15 “Digital Feed-Through”</a> )                  | —          | —                                                            | <2   | —        | nV-s          | —                                                                                                 |

**Note 1** This parameter is ensured by design.

**Note 6** Resistance is defined as the resistance between the  $V_{REF}$  pin (mode  $VRxB:VRxA = 10$ ) and the  $V_{SS}$  pin. For dual channel devices (MCP48CXDX2), this is the effective resistance of each resistor ladder. The resistance measurement is one of the two resistor ladders measured in parallel.

# MCP48CXDX1/2

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                                                                | Sym.                               | Min.                 | Typ.                | Max.                | Units         | Conditions                                                                                                                                                                                 |                  |
|---------------------------------------------------------------------------------------------------------------------------|------------------------------------|----------------------|---------------------|---------------------|---------------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------------------|
| Digital Inputs/Outputs ( $\overline{\text{CS}}$ , $\text{SDI}$ , $\text{SDO}$ , $\text{SCK}$ , $\overline{\text{LATx}}$ ) |                                    |                      |                     |                     |               |                                                                                                                                                                                            |                  |
| Schmitt Trigger High-Input Threshold                                                                                      | $V_{\text{IH}}$                    | $0.45 V_{\text{DD}}$ | —                   | —                   | V             | $1.8\text{V} \leq V_{\text{DD}} \leq 5.5\text{V}$ (allows 2.7V digital $V_{\text{DD}}$ with 5.5V analog $V_{\text{DD}}$ or 1.8V digital $V_{\text{DD}}$ with 3.0V analog $V_{\text{DD}}$ ) |                  |
| Schmitt Trigger Low-Input Threshold                                                                                       | $V_{\text{IL}}$                    | —                    | —                   | $0.2 V_{\text{DD}}$ | V             | —                                                                                                                                                                                          |                  |
| Hysteresis of Schmitt Trigger Inputs                                                                                      | $V_{\text{HYS}}$                   | —                    | $0.1 V_{\text{DD}}$ | —                   | V             | —                                                                                                                                                                                          |                  |
| Output Low Voltage (SDO)                                                                                                  | $V_{\text{OL}}$                    | $V_{\text{SS}}$      | —                   | $0.3 V_{\text{DD}}$ | V             | $I_{\text{OL}} = 200\text{ }\mu\text{A}$                                                                                                                                                   |                  |
| Output High Voltage (SDO)                                                                                                 | $V_{\text{OH}}$                    | $0.7 V_{\text{DD}}$  | —                   | $V_{\text{DD}}$     | V             | $I_{\text{OH}} = -200\text{ }\mu\text{A}$                                                                                                                                                  |                  |
| Input Leakage Current                                                                                                     | $I_{\text{IL}}$                    | -1                   | —                   | 1                   | $\mu\text{A}$ | $V_{\text{IN}} = V_{\text{DD}}$ and $V_{\text{IN}} = V_{\text{SS}}$                                                                                                                        |                  |
| Pin Capacitance                                                                                                           | $C_{\text{IN}}$ , $C_{\text{OUT}}$ | —                    | 10                  | —                   | pF            | —                                                                                                                                                                                          |                  |
| RAM Value                                                                                                                 |                                    |                      |                     |                     |               |                                                                                                                                                                                            |                  |
| Value Range                                                                                                               | N                                  | 0h                   | —                   | FFh                 | hex           | 8-bit                                                                                                                                                                                      | —                |
|                                                                                                                           |                                    | 0h                   | —                   | 3FFh                | hex           | 10-bit                                                                                                                                                                                     | —                |
|                                                                                                                           |                                    | 0h                   | —                   | FFFh                | hex           | 12-bit                                                                                                                                                                                     | —                |
| DAC Register POR/BOR Value                                                                                                | N                                  | See Table 4-2        |                     |                     | hex           | 8-bit                                                                                                                                                                                      | —                |
|                                                                                                                           |                                    | See Table 4-2        |                     |                     | hex           | 10-bit                                                                                                                                                                                     | —                |
|                                                                                                                           |                                    | See Table 4-2        |                     |                     | hex           | 12-bit                                                                                                                                                                                     | —                |
| PDCON Initial Factory Setting                                                                                             | —                                  | See Table 4-2        |                     |                     | hex           | —                                                                                                                                                                                          |                  |
| Power Requirements                                                                                                        |                                    |                      |                     |                     |               |                                                                                                                                                                                            |                  |
| Power Supply Sensitivity (B.17 “Power-Supply Sensitivity (PSS)”)                                                          | PSS                                | —                    | 0.0010              | 0.0070              | %/%           | 8-bit                                                                                                                                                                                      | Code = mid-scale |
|                                                                                                                           |                                    |                      |                     |                     |               | 10-bit                                                                                                                                                                                     |                  |
|                                                                                                                           |                                    |                      |                     |                     |               | 12-bit                                                                                                                                                                                     |                  |

## DC CHARACTERISTICS (CONTINUED)

### Standard Operating Conditions (unless otherwise specified):

Operating Temperature:  $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$  (Extended)

All parameters apply across the specified operating ranges unless noted.

$V_{DD} = +2.7\text{V}$  to  $5.5\text{V}$ ,  $V_{REF} = +1.000\text{V}$  to  $V_{DD}$ ,  $V_{SS} = 0\text{V}$ ,

$R_L = 2\text{ k}\Omega$  from  $V_{OUT}$  to GND,  $C_L = 100\text{ pF}$ .

Typical specifications represent values for  $V_{DD} = 5.5\text{V}$ ,  $T_A = +25^{\circ}\text{C}$ .

| Parameters                                                                     | Sym.                 | Min.                          | Typ. | Max.               | Units  | Conditions                                                                                                                                                                                                     |
|--------------------------------------------------------------------------------|----------------------|-------------------------------|------|--------------------|--------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| <b>Multi-Time Programming Memory (MTP)</b>                                     |                      |                               |      |                    |        |                                                                                                                                                                                                                |
| MTP Programming Voltage ( <a href="#">Note 1</a> )                             | $V_{PG\_MTP}$        | 2.0                           | —    | 5.5                | V      | $HVC = V_{IH\overline{H}}$ , $-20^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$                                                                                                                          |
| LAT/HVC pin Voltage for MTP Programming (High-Voltage Commands) <sup>(1)</sup> | $V_{IH\overline{H}}$ | 7.25                          | 7.5  | 7.75V              | V      | The LAT/HVC pin will be at one of the three input levels ( $V_{IL}$ , $V_{IH}$ or $V_{IH\overline{H}}$ ) <sup>(1,11)</sup><br>The LAT/HVC pin must supply the required MTP programming current (up to 6.4 mA). |
| Writes Cycles                                                                  | —                    | —                             | —    | 32 <sup>(12)</sup> | Cycles | <a href="#">Note 1</a>                                                                                                                                                                                         |
| Data Retention                                                                 | $DR_{MTP}$           | 10                            | —    | —                  | Years  | at $+85^{\circ}\text{C}$ <sup>(1)</sup>                                                                                                                                                                        |
| MTP Range                                                                      | N                    | 0h                            | —    | FFh                | hex    | 8-bit                                                                                                                                                                                                          |
|                                                                                |                      | 0h                            | —    | 3FFh               | hex    | 10-bit                                                                                                                                                                                                         |
|                                                                                |                      | 0h                            | —    | FFFh               | hex    | 12-bit                                                                                                                                                                                                         |
|                                                                                |                      | 0000h                         | —    | 7FFFh              | hex    | All general purpose memory                                                                                                                                                                                     |
| Initial Factory Setting                                                        | N                    | See <a href="#">Table 4-2</a> |      |                    | —      | —                                                                                                                                                                                                              |
| MTP Programming Write Cycle Time                                               | $t_{WC(MTP)}$        | —                             | —    | 250                | us     | $V_{DD} = +2.0\text{V}$ to $5.5\text{V}$ , $-20^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ ( <a href="#">Note 1</a> )                                                                                 |

**Note 1** This parameter is ensured by design.

**Note 11** High voltage on the LAT/HVC pin must be limited to the command plus programming time. After the programming cycle, the LAT/HVC pin voltage must be returned to 5.5V or lower.

**Note 12** After 32 MTP write cycles, writes are inhibited and the 32nd write value is retained (not corrupted).

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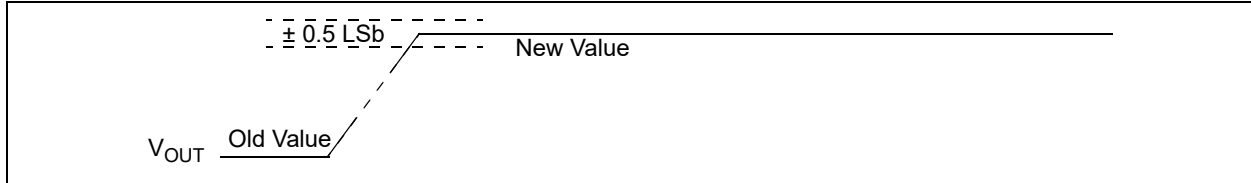
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## DC Characteristics Notes:

1. This parameter is ensured by design.
2. This parameter is ensured by characterization.
3. POR/BOR voltage trip point is not slope dependent. Hysteresis implemented with time delay.
4. Supply current is independent of current through the resistor ladder in mode VRxB:VRxA = 10.
5. The PDxB:PDxA = 01, 10 and 11 configurations must have the same current.
6. Resistance is defined as the resistance between the V<sub>REF</sub> pin (mode VRxB:VRxA = 10) and the V<sub>SS</sub> pin. For dual channel devices (MCP48CXDX2), this is the effective resistance of each resistor ladder. The resistance measurement is one of the two resistor ladders measured in parallel.
7. This gain error does not include the offset error.
8. Within 1/2 LSb of the final value when code changes from 1/4 to 3/4 of FSR. (Example: 400h to C00h in a 12-bit device.)
9. Code range dependent on resolution: 8-bit, codes 4 to 252; 10-bit, codes 16 to 1008; 12-bit, codes 64 to 4032.
10. Variation of one output voltage to mean output voltage for dual devices only.
11. High voltage on the  $\overline{\text{LAT}}$ /HVC pin must be limited to the command plus programming time. After the programming cycle, the LAT/HVC pin voltage must be returned to 5.5V or lower.
12. After 32 MTP write cycles, writes are inhibited and the 32<sup>nd</sup> write value is retained (not corrupted).

## 1.1 Timing Waveforms and Requirements

### 1.1.1 WIPER SETTTLING TIME



**FIGURE 1-1:**  $V_{OUT}$  Settling Time Waveforms.

**TABLE 1-1: WIPER SETTTLING TIMING**

| Timing Characteristics                                                 |       | Standard Operating Conditions (unless otherwise specified):                                                                                                                                                                                                                                                                 |      |      |               |            |                                                                        |
|------------------------------------------------------------------------|-------|-----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------|------|---------------|------------|------------------------------------------------------------------------|
|                                                                        |       | Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ (Extended)                                                                                                                                                                                                                                  |      |      |               |            |                                                                        |
|                                                                        |       | All parameters apply across the specified operating ranges unless noted.<br>$V_{DD} = +2.7\text{V}$ to $5.5\text{V}$ , $V_{SS} = 0\text{V}$ , $R_L = 2\text{ k}\Omega$ from $V_{OUT}$ to GND, $C_L = 100\text{ pF}$ .<br>Typical specifications represent values for $V_{DD} = 5.5\text{V}$ , $T_A = +25^{\circ}\text{C}$ . |      |      |               |            |                                                                        |
| Parameters                                                             | Sym.  | Min.                                                                                                                                                                                                                                                                                                                        | Typ. | Max. | Units         | Conditions |                                                                        |
| $V_{OUT}$ Settling Time<br>(see <a href="#">B.13 "Settling Time"</a> ) | $t_S$ | —                                                                                                                                                                                                                                                                                                                           | 4    | —    | $\mu\text{s}$ | 12-bit     | Code = 400h $\rightarrow$ C00h; C00h $\rightarrow$ 400h <sup>(3)</sup> |

**Note 3** Within 1/2 LSB of the final value when code changes from 1/4 to 3/4 of FSR.

### 1.1.2 LATCH PIN ( $\overline{\text{LAT}}$ ) TIMING



**FIGURE 1-2:**  $\overline{\text{LAT}}$  Pin Waveforms.

**TABLE 1-2:  $\overline{\text{LAT}}$  PIN TIMING**

| Timing Characteristics                   |                  | Standard Operating Conditions (unless otherwise specified):                                                                                                                                                                                                                                                                 |      |      |       |            |  |
|------------------------------------------|------------------|-----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------|------|-------|------------|--|
|                                          |                  | Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ (Extended)                                                                                                                                                                                                                                  |      |      |       |            |  |
|                                          |                  | All parameters apply across the specified operating ranges unless noted.<br>$V_{DD} = +2.7\text{V}$ to $5.5\text{V}$ , $V_{SS} = 0\text{V}$ , $R_L = 2\text{ k}\Omega$ from $V_{OUT}$ to GND, $C_L = 100\text{ pF}$ .<br>Typical specifications represent values for $V_{DD} = 5.5\text{V}$ , $T_A = +25^{\circ}\text{C}$ . |      |      |       |            |  |
| Parameters                               | Sym.             | Min.                                                                                                                                                                                                                                                                                                                        | Typ. | Max. | Units | Conditions |  |
| $\overline{\text{LATx}}$ Pin Pulse Width | $t_{\text{LAT}}$ | 20                                                                                                                                                                                                                                                                                                                          | —    | —    | ns    | —          |  |

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## 1.1.3 RESET AND POWER-DOWN TIMING



**FIGURE 1-3:** Power-on and Brown-out Reset Waveforms.



**FIGURE 1-4:** SPI Power-Down Waveforms.

**TABLE 1-3: RESET AND POWER-DOWN TIMING**

| Timing Characteristics               |               | Standard Operating Conditions (unless otherwise specified):                                                                                                                                                                                                                                                                  |      |      |               |                                                                                                                                                                                                         |
|--------------------------------------|---------------|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|------|------|---------------|---------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
|                                      |               | Operating Temperature: $-40^{\circ}\text{C} \leq T_A \leq +125^{\circ}\text{C}$ (Extended)                                                                                                                                                                                                                                   |      |      |               |                                                                                                                                                                                                         |
|                                      |               | All parameters apply across the specified operating ranges, unless noted.<br>$V_{DD} = +2.7\text{V}$ to $5.5\text{V}$ , $V_{SS} = 0\text{V}$ , $R_L = 2\text{ k}\Omega$ from $V_{OUT}$ to GND, $C_L = 100\text{ pF}$ .<br>Typical specifications represent values for $V_{DD} = 5.5\text{V}$ , $T_A = +25^{\circ}\text{C}$ . |      |      |               |                                                                                                                                                                                                         |
| Parameters                           | Sym.          | Min.                                                                                                                                                                                                                                                                                                                         | Typ. | Max. | Units         | Conditions                                                                                                                                                                                              |
| Power-on Reset Delay <sup>(1)</sup>  | $t_{POR2SIA}$ | —                                                                                                                                                                                                                                                                                                                            | —    | 130  | $\mu\text{s}$ | Single $V_{DD}$ transitions from $V_{DD(MIN)} \rightarrow > V_{POR}$ ;                                                                                                                                  |
|                                      |               | —                                                                                                                                                                                                                                                                                                                            | —    | 145  |               | Dual $V_{OUT}$ disabled to $V_{OUT}$ driven                                                                                                                                                             |
| Brown-out Reset Delay                | $t_{BORD}$    | —                                                                                                                                                                                                                                                                                                                            | 30   | —    | $\mu\text{s}$ | $V_{DD}$ transitions from $V_{DD(MIN)} \rightarrow V_{DD} > V_{POR}$ ,<br>$V_{OUT}$ driven to $V_{OUT}$ disabled                                                                                        |
| Power-Down Output Disable Time Delay | $T_{PDD}$     | —                                                                                                                                                                                                                                                                                                                            | 0.02 | —    | $\mu\text{s}$ | $PDxB:PDxA = 00 \rightarrow 11, 10$ or $01$ started from the rising edge of the 24 <sup>th</sup> SCK clock cycle;<br>$V_{OUT} = V_{OUT} - 10\text{ mV}$ ,<br>$V_{OUT}$ not connected                    |
| Power-Down Output Enable Time Delay  | $T_{PDE}$     | —                                                                                                                                                                                                                                                                                                                            | 7    | —    | $\mu\text{s}$ | $PDxB:PDxA = 11, 10$ , or $01 \rightarrow 00$ started from the rising edge of the 24 <sup>th</sup> SCK clock cycle;<br>Volatile DAC register = FFFh, $V_{OUT} = 10\text{ mV}$ , $V_{OUT}$ not connected |

**Note 1** This parameter is ensured by design.

## 1.2 SPI Mode Timing Waveforms and Requirements



**FIGURE 1-5:** SPI Timing Waveform (Mode = 11).

**TABLE 1-4: SPI REQUIREMENTS (MODE = 11)**

| #  | Characteristic <sup>(2)</sup>                                           | Sym.          | Min. | Max. | Units | Conditions                                                |
|----|-------------------------------------------------------------------------|---------------|------|------|-------|-----------------------------------------------------------|
| —  | SCK Input Frequency                                                     | $F_{SCK}$     | —    | 25   | MHz   | $V_{DD} = 2.7V$ to $5.5V$ – read command, $C_L = 20$ pF   |
|    |                                                                         |               | —    | 50   | MHz   | $V_{DD} = 2.7V$ to $5.5V$ – write commands, $C_L = 20$ pF |
|    |                                                                         |               | —    | 10   | MHz   | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 70 | $\overline{CS}$ Active ( $V_{IL}$ ) to $SCK\uparrow$ Input              | $T_{csA2scH}$ | 15   | —    | ns    | —                                                         |
| 71 | SCK Input High Time                                                     | $T_{scH}$     | 10   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                         |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 72 | SCK Input Low Time                                                      | $T_{scL}$     | 10   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                         |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 73 | SDI Input Valid to $SCK\uparrow$ Edge (Setup Time)                      | $T_{diV2scH}$ | 5    | —    | ns    | —                                                         |
| 74 | $SCK\uparrow$ Edge to SDI Input Invalid (Hold Time)                     | $T_{scH2diL}$ | 10   | —    | ns    | —                                                         |
| 77 | $\overline{CS}$ Inactive ( $V_{IH}$ ) to SDO Output High-Impedance      | $T_{csH2doZ}$ | —    | 20   | ns    | Note 1                                                    |
| 80 | $SCK\downarrow$ Edge to SDO Data Output Valid                           | $T_{scL2doV}$ | —    | 20   | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                         |               | —    | 35   | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 83 | $SCK\uparrow$ Edge to $\overline{CS}$ Inactive ( $V_{IH}$ ) (Hold Time) | $T_{scH2csi}$ | 15   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                         |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 84 | $\overline{CS}$ Input High Time                                         | $T_{csA2csi}$ | 30   | —    | ns    | —                                                         |
| 89 | Delay from HVC $V_{IHH}$ to First Command Byte <sup>(1)</sup>           | —             | 0    | —    | ns    | —                                                         |

**Note 1** This parameter is ensured by design.

**Note 2** This parameter is ensured by characterization.

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**FIGURE 1-6:** SPI Timing Waveform (Mode = 00).

**TABLE 1-5: SPI REQUIREMENTS (MODE = 00)**

| #  | Characteristic <sup>(2)</sup>                                             | Sym.          | Min. | Max. | Units | Conditions                                                |
|----|---------------------------------------------------------------------------|---------------|------|------|-------|-----------------------------------------------------------|
| —  | SCK Input Frequency                                                       | $F_{SCK}$     | —    | 25   | MHz   | $V_{DD} = 2.7V$ to $5.5V$ – read command, $C_L = 20$ pF   |
|    |                                                                           |               | —    | 50   | MHz   | $V_{DD} = 2.7V$ to $5.5V$ – write commands, $C_L = 20$ pF |
|    |                                                                           |               | —    | 10   | MHz   | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 70 | $\overline{CS}$ Active ( $V_{IL}$ or $V_{IHH}$ ) to $SCK\uparrow$ Input   | $T_{csA2scH}$ | 15   | —    | ns    | —                                                         |
| 71 | SCK Input High Time                                                       | $T_{scH}$     | 10   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                           |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 72 | SCK Input Low Time                                                        | $T_{scL}$     | 10   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                           |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 73 | SDI Input Valid to $SCK\uparrow$ Edge (Setup Time)                        | $T_{dIV2scH}$ | 5    | —    | ns    | —                                                         |
| 74 | $SCK\uparrow$ Edge to SDI Input Invalid (Hold Time)                       | $T_{scH2dIL}$ | 10   | —    | ns    | —                                                         |
| 77 | $\overline{CS}$ Inactive ( $V_{IH}$ ) to SDO Output High-Impedance        | $T_{csH2doZ}$ | —    | 20   | ns    | <b>Note 1</b>                                             |
| 80 | $SCK\downarrow$ Edge to SDO Data Output Valid                             | $T_{scL2doV}$ | —    | 20   | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                           |               | —    | 35   | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 82 | $\overline{CS}$ Active ( $V_{IL}$ ) to SDO Data Output Valid              | $T_{csL2doV}$ | —    | 20   | ns    | $2.7V$ to $5.5V$                                          |
|    |                                                                           |               | —    | 35   | ns    | $1.8V$ to $2.7V$                                          |
| 83 | $SCK\downarrow$ Edge to $\overline{CS}$ Inactive ( $V_{IH}$ ) (Hold Time) | $T_{scH2csi}$ | 15   | —    | ns    | $V_{DD} = 2.7V$ to $5.5V$                                 |
|    |                                                                           |               | 20   | —    | ns    | $V_{DD} = 1.8V$ to $2.7V$                                 |
| 84 | $\overline{CS}$ Input High Time                                           | $T_{csA2csi}$ | 30   | —    | ns    | —                                                         |
| 89 | Delay from HVC $V_{IHH}$ to First Command Byte <sup>(1)</sup>             | —             | 0    | —    | ns    | —                                                         |

**Note 1** This parameter is ensured by design.

**Note 2** This parameter is ensured by characterization.

**Timing Notes:**

1. This parameter is ensured by design.
2. This parameter is ensured by characterization.
3. Within 1/2 LSb of the final value when code changes from 1/4 to 3/4 of FSR.

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## TEMPERATURE SPECIFICATIONS

| Electrical Specifications: Unless otherwise indicated, $V_{DD} = +2.7V$ to $+5.5V$ , $V_{SS} = GND$ . |               |      |      |      |       |            |
|-------------------------------------------------------------------------------------------------------|---------------|------|------|------|-------|------------|
| Parameters                                                                                            | Sym.          | Min. | Typ. | Max. | Units | Conditions |
| <b>Temperature Ranges</b>                                                                             |               |      |      |      |       |            |
| Specified Temperature Range                                                                           | $T_A$         | -40  | —    | +125 | °C    | —          |
| Operating Temperature Range                                                                           | $T_A$         | -40  | —    | +125 | °C    | Note 1     |
| Storage Temperature Range                                                                             | $T_A$         | -65  | —    | +150 | °C    | —          |
| <b>Thermal Package Resistances</b>                                                                    |               |      |      |      |       |            |
| Thermal Resistance, 10L-MSOP                                                                          | $\theta_{JA}$ | —    | 206  | —    | °C/W  | —          |
| Thermal Resistance, 10L-DFN (3 x 3)                                                                   | $\theta_{JA}$ | —    | 91   | —    | °C/W  | —          |
| Thermal Resistance, 16L-QFN                                                                           | $\theta_{JA}$ | —    | 58   | —    | °C/W  | —          |

**Note 1:** The MCP48CXDX1/2 devices operate over this extended temperature range, but with reduced performance. Operation in this range must not cause  $T_J$  to exceed the Maximum Junction Temperature of +150°C.

## 2.0 TYPICAL PERFORMANCE CURVES

**Note:** The graphs and tables provided following this note are a statistical summary based on a limited number of samples and are provided for informational purposes only. The performance characteristics listed herein are not tested or guaranteed. In some graphs or tables, the data presented may be outside the specified operating range (e.g., outside specified power supply range) and, therefore, outside the warranted range.

### 2.1 Electrical Data

**Note:** Unless otherwise indicated,  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-1:** Average Device Supply Current vs.  $F_{SCL}$  Frequency, Voltage and Temperature – Active Interface,  $VRxB:VRxA = 00$  ( $V_{DD}$  Mode).



**FIGURE 2-4:** Average Device Supply Current vs. Voltage and Temperature – Inactive Interface ( $SCL = V_{IH}$  or  $V_{IL}$ ),  $VRxB:VRxA = 00$  ( $V_{DD}$  Mode).



**FIGURE 2-2:** Average Device Supply Current vs.  $F_{SCL}$  Frequency, Voltage and Temperature – Active Interface,  $VRxB:VRxA = 01$  (Band Gap Mode).



**FIGURE 2-5:** Average Device Supply Current vs. Voltage and Temperature – Inactive Interface ( $SCL = V_{IH}$  or  $V_{IL}$ ),  $VRxB:VRxA = 01$  (Band Gap Mode).



**FIGURE 2-3:** Average Device Supply Current vs.  $F_{SCL}$  Frequency, Voltage and Temperature – Active Interface,  $VRxB:VRxA = 10$  ( $V_{REF}$  Unbuffered Mode).



**FIGURE 2-6:** Average Device Supply Current vs. Voltage and Temperature – Inactive Interface ( $SCL = V_{IH}$  or  $V_{IL}$ ),  $VRxB:VRxA = 10$  ( $V_{REF}$  Unbuffered Mode).

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**Note:** Unless otherwise indicated,  $T_A = +25^{\circ}\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-7:** Average Device Supply Current vs.  $F_{SCL}$  Frequency, Voltage and Temperature – Active Interface,  $VRxB:VRxA = 11$  ( $V_{REF}$  Buffered Mode).



**FIGURE 2-9:** Average Device Supply Current vs. Voltage and Temperature – Inactive Interface ( $SCL = V_{IH}$  or  $V_{IL}$ ),  $VRxB:VRxA = 11$  ( $V_{REF}$  Buffered Mode).



**FIGURE 2-8:** Average Device Supply Active Current ( $I_{DDA}$ ) (at 5.5V and  $F_{SCK} = 50\text{MHz}$ ) vs. Temperature and DAC Reference Voltage Mode.

## 2.2 Linearity Data

### 2.2.1 TOTAL UNADJUSTED ERROR (TUE) – $V_{DD}$ MODE ( $VRXB:VRXA = 00$ ) MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-10:** TUE ( $V_{OUT}$ ) vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 5.5\text{V}$ ,  $G = 1x$ .



**FIGURE 2-12:** TUE ( $V_{OUT}$ ) vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 1.8\text{V}$ ,  $G = 1x$ .



**FIGURE 2-11:** TUE ( $V_{OUT}$ ) vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1x$ .

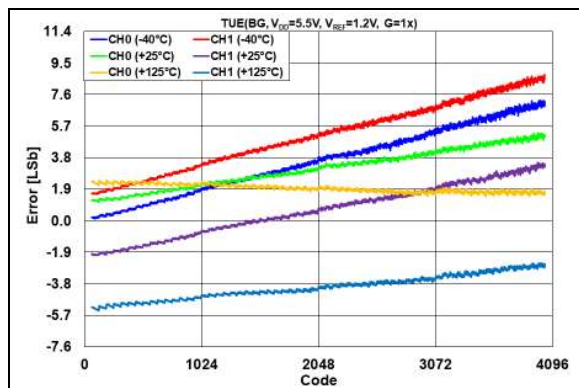


**FIGURE 2-13:** TUE ( $V_{OUT}$ ) vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

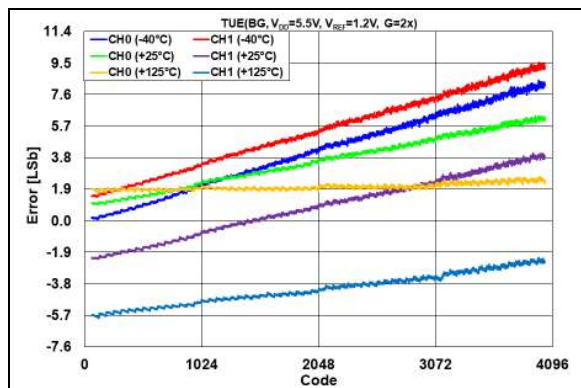
# MCP48CXDX1/2

## 2.2.2 TOTAL UNADJUSTED ERROR (TUE) – INTERNAL BAND GAP MODE (VRXB:VRXA = 01), MCP48CXD22 (12-BIT), CODE 64-4032

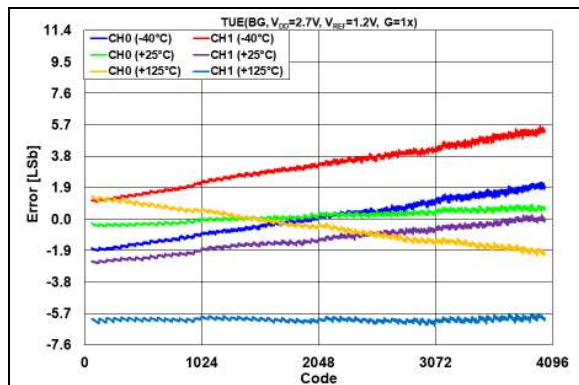
**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



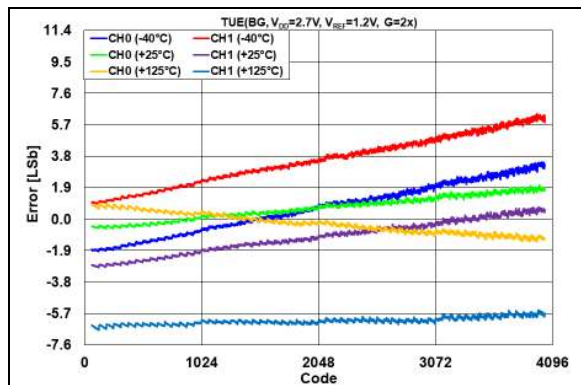
**FIGURE 2-14:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



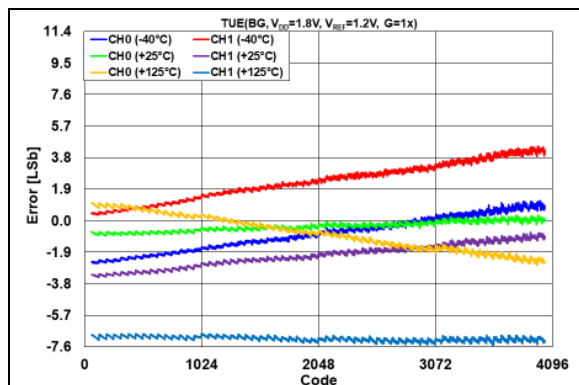
**FIGURE 2-17:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



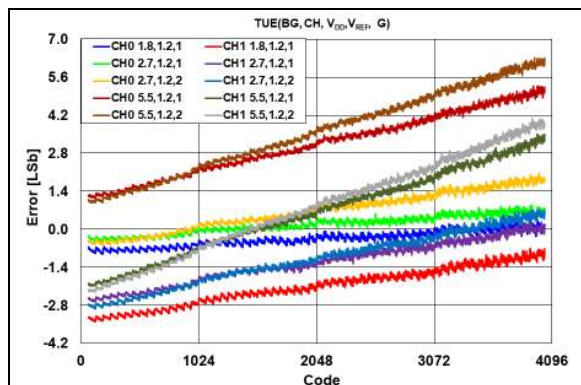
**FIGURE 2-15:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-18:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-16:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 1.8\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



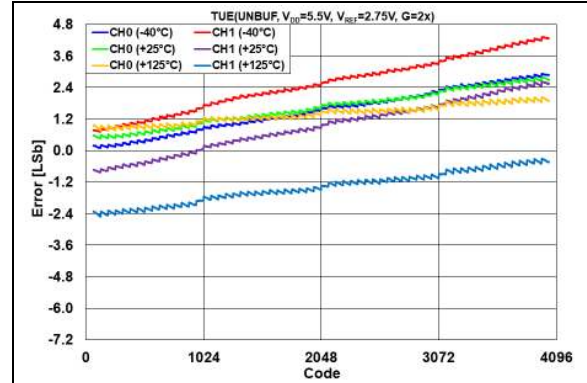
**FIGURE 2-19:**  $TUE(V_{OUT})$  vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 2.2.3 TOTAL UNADJUSTED ERROR (TUE) – EXTERNAL UNBUFFERED $V_{REF}$ MODE (VRXB:VRXA = 10), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



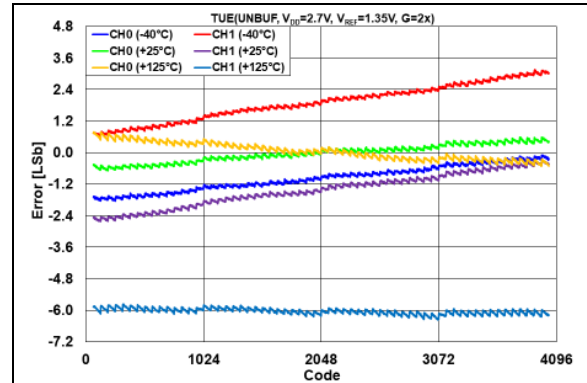
**FIGURE 2-20:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 5.5\text{V}$ ,  $G = 1x$ .



**FIGURE 2-23:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 2.75\text{V}$ ,  $G = 2x$ .



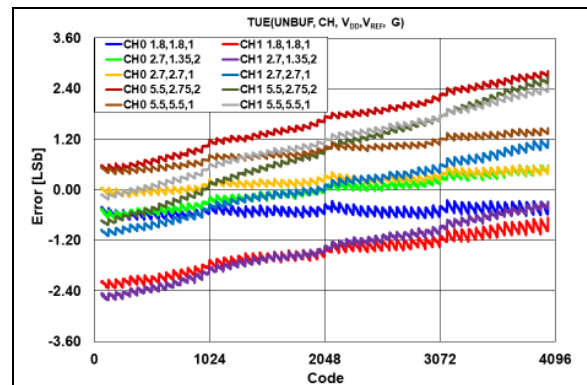
**FIGURE 2-21:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1x$ .



**FIGURE 2-24:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.35\text{V}$ ,  $G = 2x$ .



**FIGURE 2-22:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 1.8\text{V}$ ,  $G = 1x$ .



**FIGURE 2-25:**  $TUE(V_{OUT})$  vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

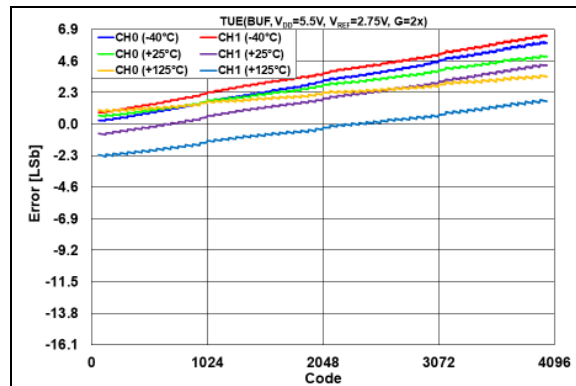
# MCP48CXDX1/2

## 2.2.4 TOTAL UNADJUSTED ERROR (TUE) – EXTERNAL BUFFERED $V_{REF}$ MODE (VRXB:VRXA = 11), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



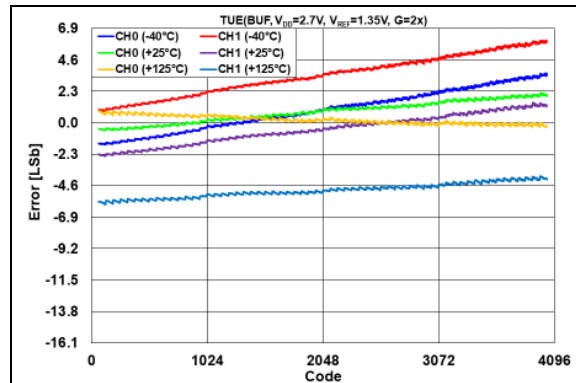
**FIGURE 2-26:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 5.5\text{V}$ ,  $G = 1x$ .



**FIGURE 2-29:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = 2.75\text{V}$ ,  $V_{DD} = 5.5\text{V}$ ,  $G = 2x$ .



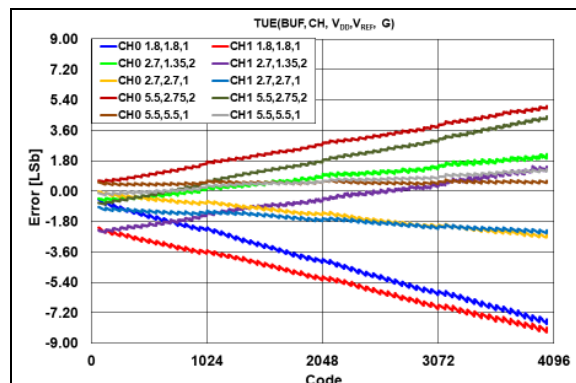
**FIGURE 2-27:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1x$ .



**FIGURE 2-30:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = 1.35\text{V}$ ,  $V_{DD} = 2.7\text{V}$ ,  $G = 2x$ .



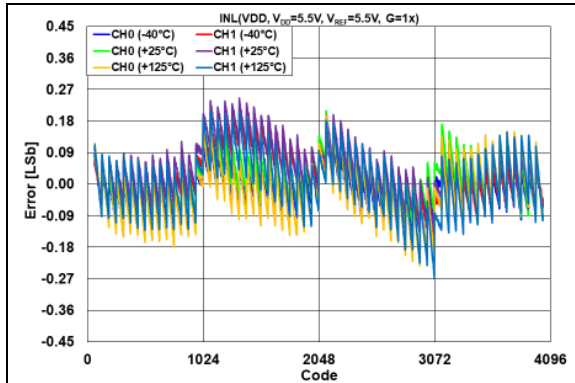
**FIGURE 2-28:**  $TUE(V_{OUT})$  vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1x$ .



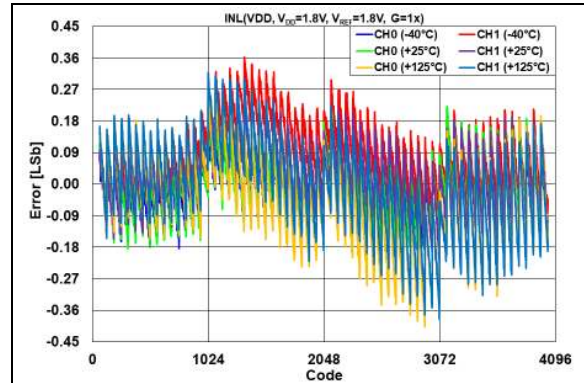
**FIGURE 2-31:**  $TUE(V_{OUT})$  vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 2.2.5 INTEGRAL NONLINEARITY (INL) – $V_{DD}$ MODE (VRXB:VRXA = 00) MCP48CXD22 (12-BIT), CODE 64-4032

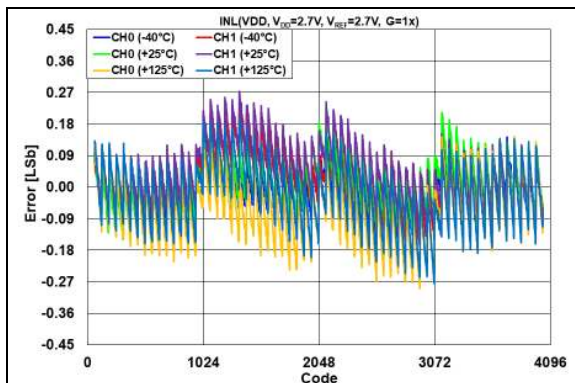
**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



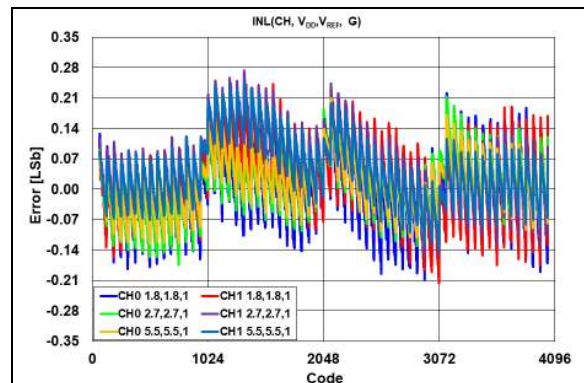
**FIGURE 2-32:** INL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-34:** INL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 1.8\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-33:** INL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-35:** INL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

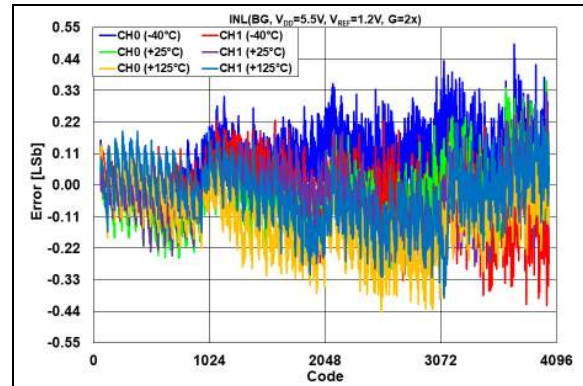
# MCP48CXDX1/2

## 2.2.6 INTEGRAL NONLINEARITY (INL) – INTERNAL BAND GAP MODE (VRXB:VRXA = 01), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



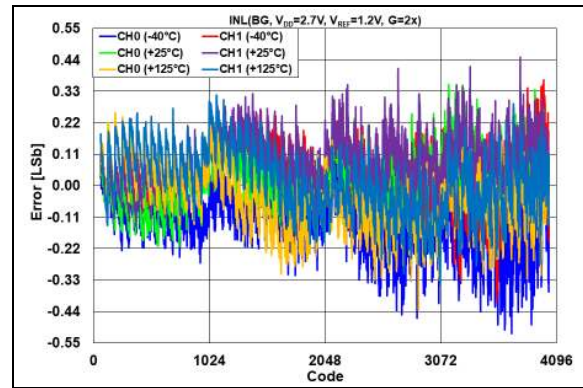
**FIGURE 2-36:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



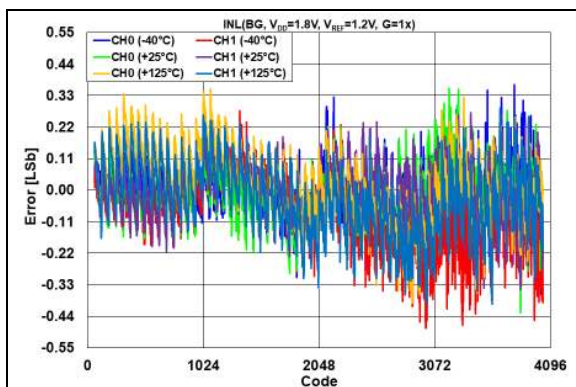
**FIGURE 2-39:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



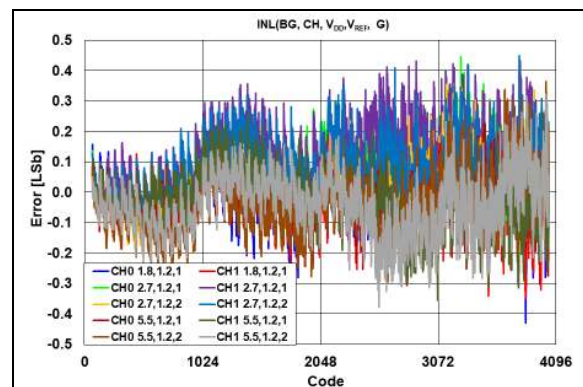
**FIGURE 2-37:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-40:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-38:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 1.8\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-41:** INL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 2.2.7 INTEGRAL NONLINEARITY (INL) – EXTERNAL UNBUFFERED $V_{REF}$ MODE (VRXB:VRXA = 10), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-42:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-45:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 2.75\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-43:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-46:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.35\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-44:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 1.8\text{V}$ ,  $G = 1\text{x}$ .

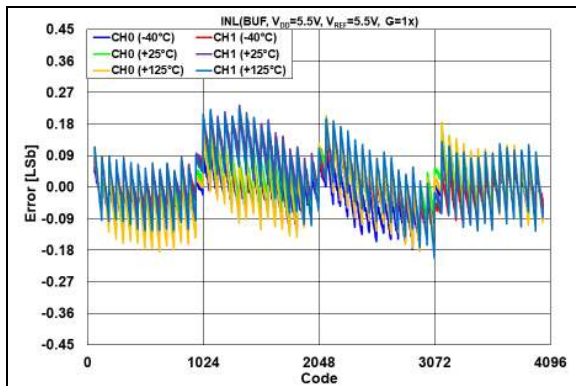


**FIGURE 2-47:** INL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

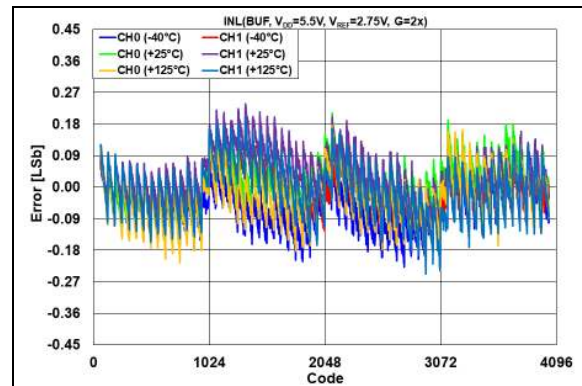
# MCP48CXDX1/2

## 2.2.8 INTEGRAL NONLINEARITY (INL) – EXTERNAL BUFFERED $V_{REF}$ MODE (VRXB:VRXA = 11), MCP48CXD22 (12-BIT), CODE 64-4032

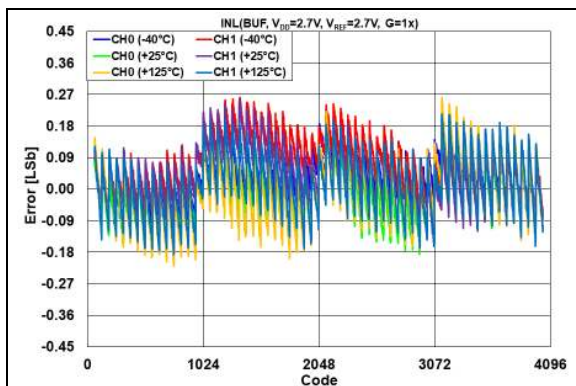
**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



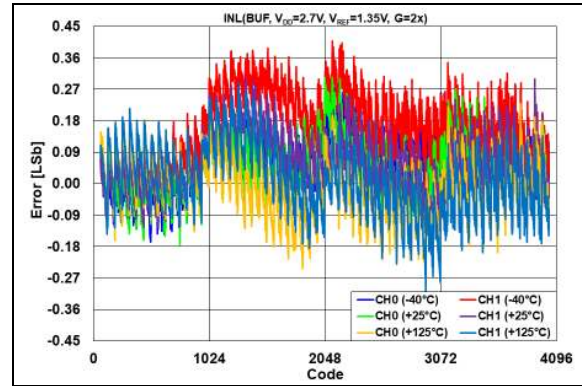
**FIGURE 2-48:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



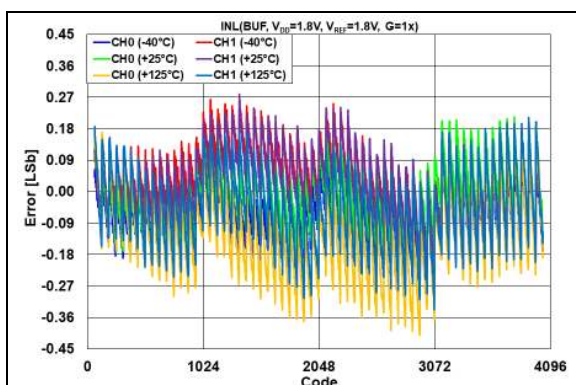
**FIGURE 2-51:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 2.75\text{V}$ ,  $G = 2\text{x}$ .



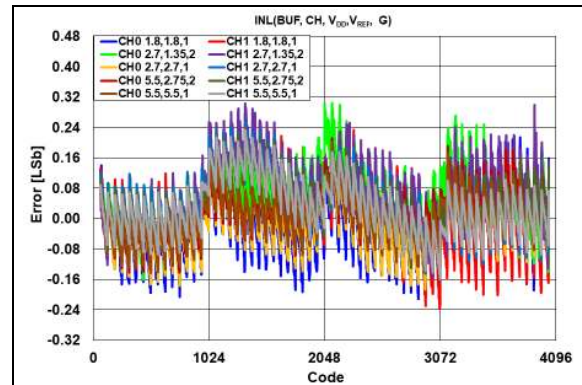
**FIGURE 2-49:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-52:** INL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.35\text{V}$ ,  $G = 2\text{x}$ .



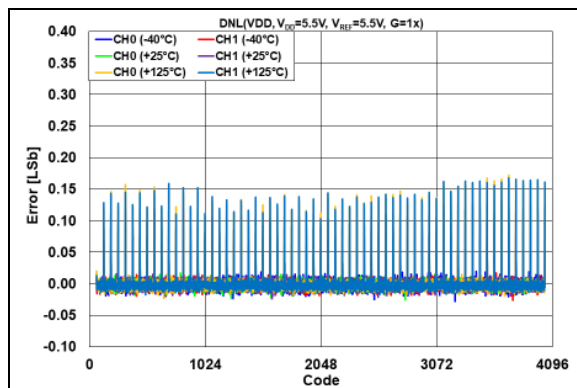
**FIGURE 2-50:** INL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 1.8\text{V}$ ,  $G = 1\text{x}$ .



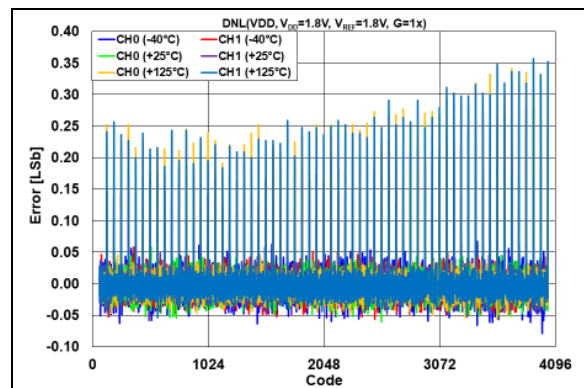
**FIGURE 2-53:** INL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 2.2.9 DIFFERENTIAL NONLINEARITY (DNL) – $V_{DD}$ MODE (VRXB:VRXA = 00) MCP48CXD22 (12-BIT), CODE 64-4032

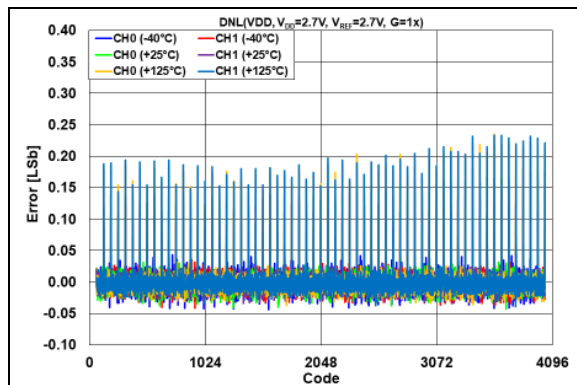
**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



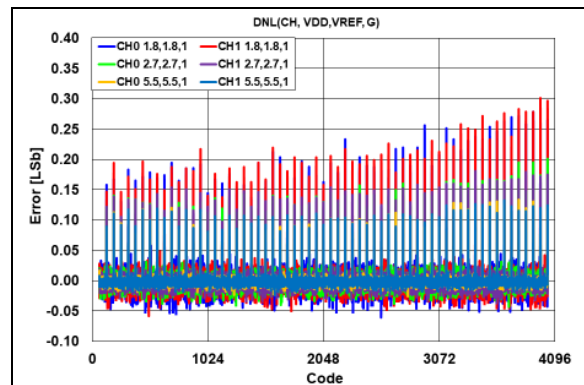
**FIGURE 2-54:** DNL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-56:** DNL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 1.8\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-55:** DNL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-57:** DNL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

# MCP48CXDX1/2

## 2.2.10 DIFFERENTIAL NONLINEARITY (DNL) – INTERNAL BAND GAP MODE (VRXB:VRXA = 01), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-58:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-61:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-59:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-62:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-60:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 1.8\text{V}$ ,  $V_{REF} = 1.2\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-63:** DNL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 2.2.11 DIFFERENTIAL NONLINEARITY (DNL) – EXTERNAL UNBUFFERED $V_{REF}$ MODE (VRXB:VRXA = 10), MCP48CXD22 (12-BIT), CODE 64-4032

**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



**FIGURE 2-64:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-67:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 2.75\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-65:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-68:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.35\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-66:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-69:** DNL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

# MCP48CXDX1/2

## 2.2.12 DIFFERENTIAL NONLINEARITY (DNL) – EXTERNAL BUFFERED $V_{REF}$ MODE (VRXB:VRXA = 11), MCP48CXD22 (12-BIT), CODE 64-4032

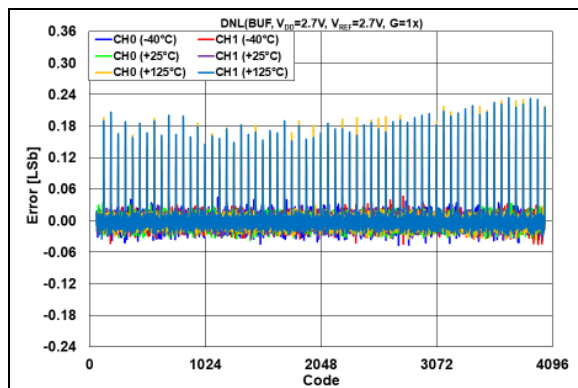
**Note:** Unless otherwise indicated:  $T_A = +25^\circ\text{C}$ ,  $V_{DD} = 5.5\text{V}$ .



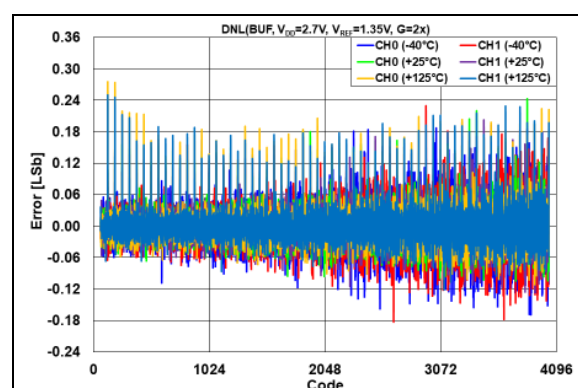
**FIGURE 2-70:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 5.5\text{V}$ ,  $G = 1\text{x}$ .



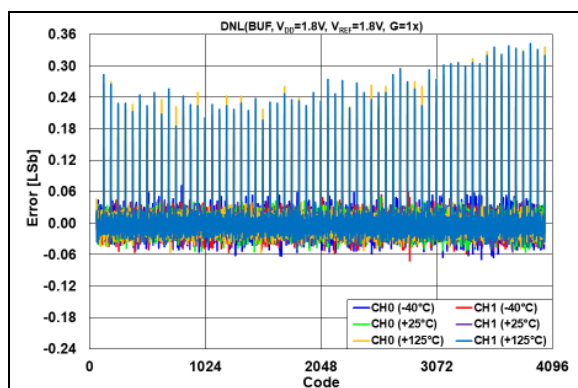
**FIGURE 2-73:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 5.5\text{V}$ ,  $V_{REF} = 2.75\text{V}$ ,  $G = 1\text{x}$ .



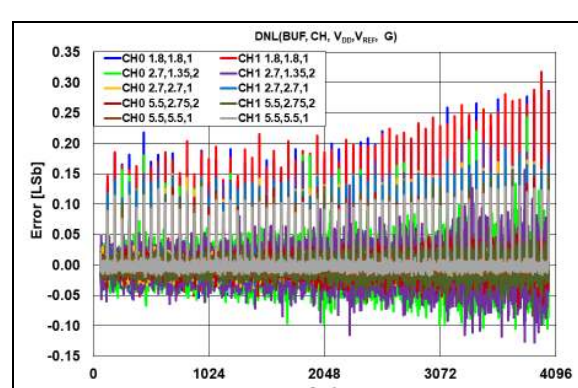
**FIGURE 2-71:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = V_{REF} = 2.7\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-74:** DNL Error vs. DAC Code and Temperature,  $V_{DD} = 2.7\text{V}$ ,  $V_{REF} = 1.35\text{V}$ ,  $G = 2\text{x}$ .



**FIGURE 2-72:** DNL Error vs. DAC Code and Temperature,  $V_{REF} = V_{DD} = 1.8\text{V}$ ,  $G = 1\text{x}$ .



**FIGURE 2-75:** DNL Error vs. DAC Code, Temperature =  $+25^\circ\text{C}$ .

## 3.0 PIN DESCRIPTIONS

Overviews of the pin functions are provided from [Section 3.1 “Positive Power Supply Input \( \$V\_{DD}\$ \)”](#) to [Section 3.10 “SPI – Serial Data Out Pin \(SDO\)”](#).

The descriptions of the pins for the single DAC output device are listed in [Table 3-1](#), and descriptions for the dual DAC output device are listed in [Table 3-2](#).

**TABLE 3-1: MCP48CXDX1 (SINGLE DAC) PIN FUNCTION TABLE**

| Pin      |         |                 | Symbol          | I/O | Buffer Type | Description                                                                                                                                                                                                                                                                        |
|----------|---------|-----------------|-----------------|-----|-------------|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| MSOP 10L | DFN 10L | QFN 16L         |                 |     |             |                                                                                                                                                                                                                                                                                    |
| 1        | 1       | 16              | $V_{DD}$        | —   | P           | Supply Voltage                                                                                                                                                                                                                                                                     |
| 2        | 2       | 1               | $\overline{CS}$ | I   | ST          | SPI Chip Select                                                                                                                                                                                                                                                                    |
| 3        | 3       | 2               | $V_{REF}$       | A   | Analog      | Voltage Reference Input/Output                                                                                                                                                                                                                                                     |
| 4        | 4       | 3               | $V_{OUT}$       | A   | Analog      | Buffered Analog Voltage Output                                                                                                                                                                                                                                                     |
| 5        | 5       | 4,5,6,7,8,14,15 | NC              | —   | —           | Not Internally Connected                                                                                                                                                                                                                                                           |
| 6        | 6       | 9               | LAT/HVC         | I   | ST          | DAC Wiper Register Latch/High-Voltage Command Pin.<br>The Latch Pin allows the value in the volatile DAC registers (wiper and configuration bits) to be transferred to the DAC output ( $V_{OUT}$ ).<br>High-Voltage commands allow the User MTP configuration bits to be written. |
| 7        | 7       | 10              | $V_{SS}$        | —   | P           | Ground Reference for all circuitries on the device                                                                                                                                                                                                                                 |
| 8        | 8       | 11              | SDO             | O   | ST          | SPI Serial Data Output                                                                                                                                                                                                                                                             |
| 9        | 9       | 12              | SCK             | I   | ST          | SPI Serial Clock                                                                                                                                                                                                                                                                   |
| 10       | 10      | 13              | SDI             | I   | ST          | SPI Serial Data Input                                                                                                                                                                                                                                                              |
| —        | —       | 17              | EP              | —   | P           | Exposed Thermal Pad, must be connected to $V_{SS}$                                                                                                                                                                                                                                 |

**Note 1:** A = Analog, I = Input, ST = Schmitt Trigger, O = Output, I/O = Input/Output, P = Power

# MCP48CXDX1/2

**TABLE 3-2: MCP48CXDX2 (DUAL DAC) PIN FUNCTION TABLE**

| Pin      |         |            | Symbol                | I/O | Buffer Type | Description                                                                                                                                                                                                                                                                            |
|----------|---------|------------|-----------------------|-----|-------------|----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| MSOP 10L | DFN 10L | QFN 16L    |                       |     |             |                                                                                                                                                                                                                                                                                        |
| 1        | 1       | 16         | $V_{DD}$              | —   | P           | Supply Voltage                                                                                                                                                                                                                                                                         |
| 2        | 2       | 1          | $\overline{CS}$       | I   | ST          | SPI Chip Select                                                                                                                                                                                                                                                                        |
| 3        | 3       | —          | $V_{REF}$             | A   | Analog      | Voltage Reference Input/Output                                                                                                                                                                                                                                                         |
| —        | —       | 2          | $V_{REF0}$            | A   | Analog      | Voltage Reference Input/Output for DAC0                                                                                                                                                                                                                                                |
| —        | —       | 4          | $V_{REF1}$            | A   | Analog      | Voltage Reference Input/Output for DAC1                                                                                                                                                                                                                                                |
| 4        | 4       | 3          | $V_{OUT0}$            | A   | Analog      | Buffered Analog Voltage Output 0                                                                                                                                                                                                                                                       |
| 5        | 5       | 5          | $V_{OUT1}$            | A   | Analog      | Buffered Analog Voltage Output 1                                                                                                                                                                                                                                                       |
| —        | —       | 6,7,14, 15 | NC                    | —   | —           | Not Internally Connected                                                                                                                                                                                                                                                               |
| 6        | 6       | —          | $\overline{LAT}/HVC$  | I   | ST          | DAC Wiper Register Latch/High-Voltage Command Pin.<br>The Latch Pin allows the value in the volatile DAC registers (wiper and configuration bits) to be transferred to the DAC output ( $V_{OUT}$ ).<br>High-Voltage commands allow the User MTP configuration bits to be written.     |
| —        | —       | 9          | $\overline{LAT0}/HVC$ | I   | ST          | DAC0 Wiper Register Latch/High-Voltage Command Pin.<br>The Latch Pin allows the value in the volatile DAC0 registers (wiper and configuration bits) to be transferred to the DAC0 output ( $V_{OUT0}$ ).<br>High-Voltage commands allow the User MTP configuration bits to be written. |
| —        | —       | 8          | $\overline{LAT1}$     | I   | ST          | DAC1 Wiper Register Latch.<br>The Latch Pin allows the value in the volatile DAC1 registers (wiper and configuration bits) to be transferred to the DAC1 output ( $V_{OUT1}$ ).                                                                                                        |
| 7        | 7       | 10         | $V_{SS}$              | —   | P           | Ground Reference for all circuitries on the device                                                                                                                                                                                                                                     |
| 8        | 8       | 11         | SDO                   | O   | ST          | SPI Serial Data Output                                                                                                                                                                                                                                                                 |
| 9        | 9       | 12         | SCK                   | I   | ST          | SPI Serial Clock                                                                                                                                                                                                                                                                       |
| 10       | 10      | 13         | SDI                   | I   | ST          | SPI Serial Data Input                                                                                                                                                                                                                                                                  |
| —        | —       | 17         | EP                    | —   | P           | Exposed Thermal Pad, must be connected to $V_{SS}$                                                                                                                                                                                                                                     |

**Note 1:** A = Analog, I = Input, ST = Schmitt Trigger, O = Output, I/O = Input/Output, P = Power

## 3.1 Positive Power Supply Input ( $V_{DD}$ )

$V_{DD}$  is the positive supply voltage input pin. The input supply voltage is relative to  $V_{SS}$ .

The power supply at the  $V_{DD}$  pin must be as clean as possible for good DAC performance. It is recommended to use an appropriate bypass capacitor of about 0.1  $\mu$ F (ceramic) to ground as close as possible to the pin. An additional 10  $\mu$ F capacitor (tantalum) in parallel is also recommended to further attenuate noise present in application boards.

## 3.2 Ground ( $V_{SS}$ )

The  $V_{SS}$  pin is the device ground reference.

The user must connect the  $V_{SS}$  pin to a ground plane through a low-impedance connection. If an analog ground path is available in the application PCB (Printed Circuit Board), it is highly recommended that the  $V_{SS}$  pin be tied to the analog ground path or isolated within an analog ground plane of the circuit board.

## 3.3 Voltage Reference Pin ( $V_{REF}$ )

The  $V_{REF}$  pin is either an input or an output. When the DAC's voltage reference is configured as the  $V_{REF}$  pin, the pin is an input. When the DAC's voltage reference is configured as the internal band gap, the pin is an output.

When the DAC's voltage reference is configured as the  $V_{REF}$  pin, there are two options for this voltage input: the  $V_{REF}$  pin voltage is buffered or unbuffered. The buffered option is offered in cases where the external reference voltage does not have sufficient current capability to not drop its voltage when connected to the internal resistor ladder circuit.

When the DAC's voltage reference is configured as the device  $V_{DD}$ , the  $V_{REF}$  pin is disconnected from the internal circuit.

When the DAC's voltage reference is configured as the internal band gap, the  $V_{REF}$  pin's drive capability is minimal, so the output signal must be buffered.

See [Section 5.2 "Voltage Reference Selection"](#) and [Register 4-2](#) for more details on the configuration bits.

## 3.4 No Connect (NC)

The NC pin is not internally connected to the device.

## 3.5 Analog Output Voltage Pins ( $V_{OUT0}$ , $V_{OUT1}$ )

$V_{OUT0}$  and  $V_{OUT1}$  are the DAC analog voltage output pins. Each DAC output has an output amplifier. The DAC output range depends on the selection of the voltage reference source (and potential output gain selection). These are:

- Device  $V_{DD}$  – The full-scale range of the DAC output is from  $V_{SS}$  to approximately  $V_{DD}$ .
- $V_{REF}$  pin – The full-scale range of the DAC output is from  $V_{SS}$  to  $G \times V_{RL}$ , where  $G$  is the gain selection option (1X or 2X).
- Internal band gap – The full-scale range of the DAC output is from  $V_{SS}$  to  $G \times V_{BG}$ , where  $G$  is the gain selection option (1X or 2X).

In Normal mode, the DC impedance of the output pin is about 1 $\Omega$ . In Power-Down mode, the output pin is internally connected to a known pull-down resistor of 1 k $\Omega$ , 100 k $\Omega$  or open. The Power-Down selection bits settings are shown in [Register 4-3 \(Table 5-5\)](#).

## 3.6 Latch/High-Voltage Command Pin ( $\overline{LAT}/HVC$ )

The DAC output value update event can be controlled and synchronized using the  $\overline{LAT}$  pin, for one or both channels, on a single or different devices.

The  $\overline{LAT}$  pin controls the effect of the volatile wiper registers,  $VRxB:VRxA$ ,  $PDxB:PDxA$  and  $Gx$  bits on the DAC output.

If the  $\overline{LAT}$  pin is held at  $V_{IH}$ , the values sent to the volatile wiper registers and configuration bits have no effect on the DAC outputs.

Once voltage on the pin transitions to  $V_{IL}$ , the values in the volatile wiper registers and configuration bits are transferred to the DAC outputs.

The pin is level-sensitive, so writing to the volatile wiper registers and configuration bits, while it is being held at  $V_{IL}$ , will trigger an immediate change in the outputs.

For dual output devices in MSOP and DFN packages, the  $\overline{LAT}$  pin controls both channels at the same time.

The HVC pin allows the device's MTP memory to be programmed for the MCP48CXDX1/2 devices. The programming voltage supply must provide 7.5V and at least 6.4 mA.

**Note:** The HVC pin must have voltages greater than 5.5V present only during the MTP programming operation. Using voltages greater than 5.5V for an extended time on the pin may cause device reliability issues.

# MCP48CXDX1/2

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## 3.7 SPI – Chip Select Pin ( $\overline{\text{CS}}$ )

The  $\overline{\text{CS}}$  pin enables/disables the serial interface (SDI, SDO and SCK). The serial interface must be enabled for the device to receive any serial commands.

See [Section 6.4 “Interface Pins \(CS, SCK, SDI, SDO and LAT/HVC\)”](#) for more details regarding the SPI serial interface communication.

## 3.8 SPI – Serial Clock Pin (SCK)

The SCK pin is the serial clock pin of the SPI interface. The MCP48CXDX1/2 SPI interface only accepts external serial clocks.

## 3.9 SPI – Serial Data In Pin (SDI)

The SDI pin is the serial data input pin of the SPI interface. The SDI pin is used to write the DAC wiper registers and configuration bits.

## 3.10 SPI – Serial Data Out Pin (SDO)

The SDO pin is the serial data output pin of the SPI interface. The SDO pin is used to read the DAC wiper registers and configuration bits.

## 4.0 GENERAL DESCRIPTION

The MCP48CXDX1 devices are single channel voltage output devices.

The MCP48CXDX2 devices are dual channel voltage output devices.

These devices are offered with 8-bit (MCP48CXDX0X), 10-bit (MCP48CXDX1X) and 12-bit (MCP48CXDX2X) resolutions.

The family offers two memory options: the MCP48CVDXX devices have a volatile memory, while the MCP48CMDXX have a 32-times programmable nonvolatile memory (MTP).

All devices include an SPI serial interface and a write latch (LAT) pin to control the update of the analog output voltage value from the value written in the volatile DAC output register.

The devices use a resistor ladder architecture. The resistor ladder DAC is driven from a software-selectable voltage reference source. The source can be either the device's internal  $V_{DD}$ , an external  $V_{REF}$  pin voltage (buffered or unbuffered) or an internal band gap voltage source.

The DAC output is buffered with a low power and precision output amplifier. This output amplifier provides a rail-to-rail output with low offset voltage and low noise. The gain (1X or 2X) of the output buffer is software configurable.

The devices operate from a single supply voltage. This voltage is specified from 2.7V to 5.5V for full specified operation, and from 1.8V to 5.5V for digital operation. The device operates between 1.8V and 2.7V, but some device parameters are not specified.

The MCP48CXDX1/2 devices also have user-programmable nonvolatile configuration memory (MTP). This allows the device's desired POR values to be saved. The device also has general purpose MTP memory locations for storing system-specific information (calibration data, serial numbers, system ID information). A high-voltage requirement for programming the nonvolatile locations on the HVC pin ensures that these device settings are not accidentally modified during normal system operation. Therefore, it is recommended that the MTP memory be only programmed at the user's factory.

The main functional blocks are:

- **Power-on Reset/Brown-out Reset (POR/BOR)**
- **Device Memory**
- **Resistor Ladder**
- **Output Buffer/ $V_{OUT}$  Operation**
- **SPI Serial Interface Module**

## 4.1 Power-on Reset/Brown-out Reset (POR/BOR)

The internal POR/BOR circuit monitors the power supply voltage ( $V_{DD}$ ) during operation. This circuit ensures correct device start-up at system power-up and power-down events.

The device's RAM retention voltage ( $V_{RAM}$ ) is lower than the POR/BOR voltage trip point ( $V_{POR}/V_{BOR}$ ). The maximum  $V_{POR}/V_{BOR}$  voltage is less than 1.8V.

The POR and BOR trip points are at the same voltage, and the condition is determined by whether the  $V_{DD}$  voltage is rising or falling (see Figure 4-1). What occurs is different depending on whether the reset is a POR or BOR.

POR occurs as the voltage rises (typically from 0V), while BOR occurs as the voltage falls (typically from  $V_{DD(MIN)}$  or higher).

When  $V_{POR}/V_{BOR} < V_{DD} < 2.7V$ , the electrical performance may not meet the data sheet specifications. In this region, the device is capable of reading and writing to its volatile memory if the proper serial command is executed.

### 4.1.1 POWER-ON RESET

The POR is the case where the device's  $V_{DD}$  has power applied to it from the  $V_{SS}$  voltage level. As the device powers-up, the  $V_{OUT}$  pin floats to an unknown value. When the device's  $V_{DD}$  is above the transistor threshold voltage of the device, the output starts to be pulled low.

After the  $V_{DD}$  is above the POR/BOR trip point ( $V_{BOR}/V_{POR}$ ), the resistor network's wiper is loaded with the POR value. The POR value is either mid-scale (MCP48CVDXX) or the user's MTP programmed value (MCP48CMDXX).

**Note:** In order to have the MCP48CMDXX devices load the values from nonvolatile memory locations at POR, they have to be programmed at least once by the user; otherwise, the loaded values will be the default ones. After MTP programming, a POR event is required to load the written values from the nonvolatile memory.

Volatile memory determines the analog output ( $V_{OUT}$ ) pin voltage. After the device is powered-up, the user can update the device memory.

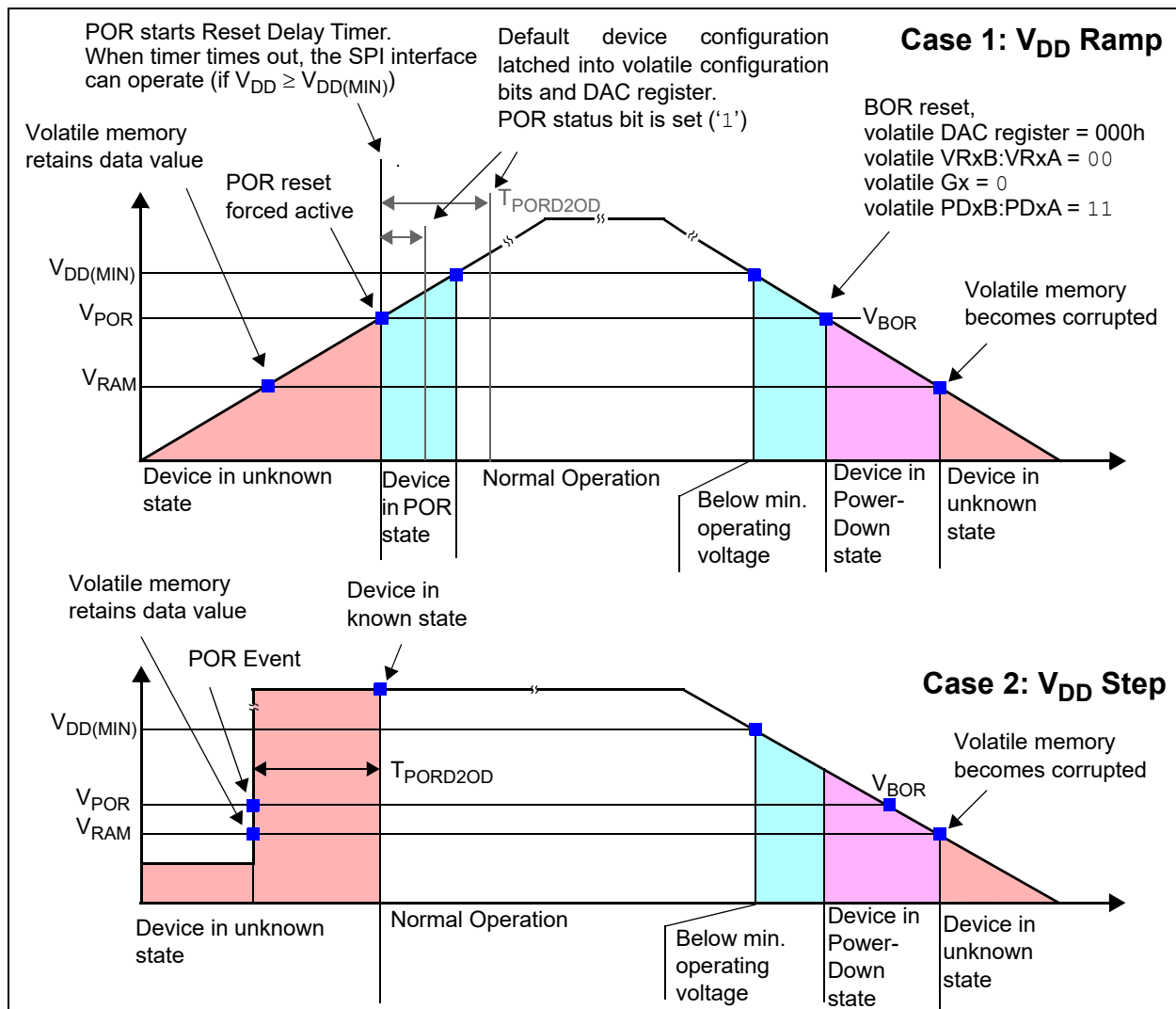
# MCP48CXDX1/2

When the rising  $V_{DD}$  voltage crosses the  $V_{POR}$  trip point, the following occurs:

- The default DAC POR value is latched into the volatile DAC register.
- The default DAC POR Configuration bit values are latched into the volatile configuration bits.
- POR Status bit is set ('1').
- The reset delay timer ( $t_{POR\Delta}$ ) starts; when the reset delay timer times out, the SPI serial interface is operational. During this delay time, the SPI interface will not accept commands.
- The device memory address pointer is forced to 00h.

The analog output ( $V_{OUT}$ ) state is determined by the state of the volatile configuration bits and the DAC register. This is called a Power-on Reset (event).

Figure 4-1 illustrates the conditions for power-up and power-down events under typical conditions.



**FIGURE 4-1:** Power-on Reset Operation.

## 4.1.2 BROWN-OUT RESET

A BOR occurs when a device has power applied to it and that power (voltage) drops below the specified range.

When the falling  $V_{DD}$  voltage crosses the  $V_{POR}$  trip point (BOR event), the following occurs:

- Serial interface is disabled.
- MTP writes are disabled.
- Device is forced into a Power-Down state ( $PDxB:PDxA = 11$ ). Analog circuitry is turned off.
- Volatile DAC register is forced to 000h.

Volatile configuration bits  $VRxB:VRxA$  and  $GX$  are forced to '0'.

If the  $V_{DD}$  voltage decreases below the  $V_{RAM}$  voltage, all volatile memory may become corrupted.

As the voltage recovers above the  $V_{POR}/V_{BOR}$  voltage, see [Section 4.1.1 "Power-on Reset"](#) for further details.

Serial commands not completed due to a brown-out condition may cause the memory location to become corrupted.

[Figure 4-1](#) illustrates the conditions for power-up and power-down events under typical conditions.

## 4.2 Device Memory

User memory includes the following types:

- **Volatile Register Memory (RAM)**
- **Nonvolatile Register Memory (MTP)**

MTP memory is present just for the MCP48CMDXX devices and has three groupings:

- NV DAC Output Values (loaded on POR event)
- Device Configuration Memory
- General Purpose NV Memory

Each memory location is up to 16 bits wide. The memory mapped register space is shown in [Table 4-1](#).

The SPI interface depends on how this memory is read and written. Refer to [Section 6.0 "SPI Serial Interface Module"](#) and [Section 7.0 "Device Commands"](#) for more details on reading and writing the device's memory.

### 4.2.1 VOLATILE REGISTER MEMORY (RAM)

The MCP48CXDX1/2 devices have volatile memory to directly control the operation of the DACs. There are up to five volatile memory locations:

- DAC0 and DAC1 Output Value registers
- $V_{REF}$  Select register
- Power-Down Configuration register
- Gain and Status register

The volatile memory starts functioning when the device  $V_{DD}$  is at (or above) the RAM retention voltage ( $V_{RAM}$ ). The volatile memory will be loaded with the default device values when the  $V_{DD}$  rises across the  $V_{POR}/V_{BOR}$  voltage trip point.

After the device is powered-up, the user can update the device memory. [Table 4-2](#) shows the volatile memory locations and their interaction due to a POR event.

### 4.2.2 NONVOLATILE REGISTER MEMORY (MTP)

This memory option is available only for the MCP48CMDXX devices.

MTP memory starts functioning below the device's  $V_{POR}/V_{BOR}$  trip point and, once the  $V_{POR}$  event occurs, the volatile memory registers are loaded with the corresponding MTP register memory values.

Memory addresses 0Ch through 1Fh are nonvolatile memory locations. These locations contain the DAC POR/BOR wiper values, the DAC POR/BOR configuration bits and 8 general purpose memory locations for storing user-defined data as calibration constants or identification numbers.

The nonvolatile wiper registers and configuration bits determine the DAC Output and Configuration values at the POR event.

These nonvolatile values will overwrite the factory default values. If these MTP addresses are unprogrammed, the factory default values define the output state.

The nonvolatile DAC registers enable the standalone operation of the device (without microcontroller control), after being programmed to the desired values.

To program nonvolatile memory locations, a high-voltage source on the  $\overline{LAT}/HVC$  pin is required. Each register/MTP location can be programmed 32 times. After 32 writes, a new write operation will not be possible and the last successful value written will remain associated with the memory location.

The device starts writing the MTP memory cells at the completion of the serial interface command. The high voltage must remain present on the  $\overline{LAT}/HVC$  pin until the write cycle is complete; otherwise, the write is unsuccessful and the location is compromised (cannot be used again and the number of available writes decreases by one).

To recover from an aborted MTP write operation, the following procedure must be used:

- Write any valid value to the same address again
- Force a POR condition
- Write the desired value to the MTP location again

It is recommended to keep high voltage on only during the MTP write command and programming cycle; otherwise, the reliability of the device could be affected.

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## 4.2.3 UNIMPLEMENTED LOCATIONS

### 4.2.3.1 Unimplemented Register Bits

When issuing read commands to a valid memory location with unimplemented bits, the unimplemented bits will be read as '0'.

## 4.2.4 UNIMPLEMENTED (RESERVED) LOCATIONS

There are a number of unimplemented memory locations that are reserved for future use. Normal (voltage) commands (read or write) to any unimplemented memory address will result in a Command Error condition (SPI Command Error – CMDERR). High-Voltage commands to any unimplemented configuration bit(s) will also result in a Command Error condition.

## 4.2.5 POR/BOR OPERATION WITH WIPERLOCK™ TECHNOLOGY ENABLED

Regardless of the WiperLock Technology state, a POR event will load the Volatile DACx Wiper register value with the Nonvolatile DACx Wiper register value. Refer to [Section 4.1 “Power-on Reset/Brown-out Reset \(POR/BOR\)”](#) for further information.

**TABLE 4-1: MCP48CXDX1/2 MEMORY MAP (16-BIT)**

| Address | Function                          | Single | Dual | Address | Function                          | Single <sup>(1)</sup> | Dual <sup>(1)</sup> |
|---------|-----------------------------------|--------|------|---------|-----------------------------------|-----------------------|---------------------|
| 00h     | Volatile DAC Wiper Register 0     | Y      | Y    | 10h     | Nonvolatile DAC Wiper Register 0  | Y                     | Y                   |
| 01h     | Volatile DAC Wiper Register 1     | —      | Y    | 11h     | Nonvolatile DAC Wiper Register 1  | —                     | Y                   |
| 02h     | Reserved                          | —      | —    | 12h     | Reserved                          | —                     | —                   |
| 03h     | Reserved                          | —      | —    | 13h     | Reserved                          | —                     | —                   |
| 04h     | Reserved                          | —      | —    | 14h     | Reserved                          | —                     | —                   |
| 05h     | Reserved                          | —      | —    | 15h     | Reserved                          | —                     | —                   |
| 06h     | Reserved                          | —      | —    | 16h     | Reserved                          | —                     | —                   |
| 07h     | Reserved                          | —      | —    | 17h     | Reserved                          | —                     | —                   |
| 08h     | Volatile VREF Register            | Y      | Y    | 18h     | Nonvolatile VREF Register         | Y                     | Y                   |
| 09h     | Volatile Power-Down Register      | Y      | Y    | 19h     | Nonvolatile Power-Down Register   | Y                     | Y                   |
| 0Ah     | Volatile Gain and Status Register | Y      | Y    | 1Ah     | NV Gain                           | Y                     | Y                   |
| 0Bh     | Reserved                          | —      | —    | 1Bh     | NV WiperLock™ Technology Register | Y                     | Y                   |
| 0Ch     | General Purpose MTP               | (1)    |      | 1Ch     | General Purpose MTP               | (1)                   |                     |
| 0Dh     | General Purpose MTP               | (1)    |      | 1Dh     | General Purpose MTP               | (1)                   |                     |
| 0Eh     | General Purpose MTP               | (1)    |      | 1Eh     | General Purpose MTP               | (1)                   |                     |
| 0Fh     | General Purpose MTP               | (1)    |      | 1Fh     | General Purpose MTP               | (1)                   |                     |

Legend:

|  |                                                        |
|--|--------------------------------------------------------|
|  | Volatile memory addresses                              |
|  | MTP memory addresses                                   |
|  | Memory locations not implemented on this device family |

**Note 1:** On nonvolatile memory devices only (MCP48CMDXX)

**TABLE 4-2: FACTORY DEFAULT POR/BOR VALUES (MTP MEMORY UNPROGRAMMED)**

| Address | Function                                         | POR/BOR Value |        |        |
|---------|--------------------------------------------------|---------------|--------|--------|
|         |                                                  | 8-bit         | 10-bit | 12-bit |
| 00h     | Volatile DAC0 Register                           | 7Fh           | 1FFh   | 7FFh   |
| 01h     | Volatile DAC1 Register                           | 7Fh           | 1FFh   | 7FFh   |
| 02h     | Reserved                                         | —             | —      | —      |
| 03h     | Reserved                                         | —             | —      | —      |
| 04h     | Reserved                                         | —             | —      | —      |
| 05h     | Reserved                                         | —             | —      | —      |
| 06h     | Reserved                                         | —             | —      | —      |
| 07h     | Reserved                                         | —             | —      | —      |
| 08h     | Volatile VREF Register                           | 0000h         | 0000h  | 0000h  |
| 09h     | Volatile Power-Down Register                     | 0000h         | 0000h  | 0000h  |
| 0Ah     | Volatile Gain and Status Register <sup>(2)</sup> | 0080h         | 0080h  | 0080h  |
| 0Bh     | Reserved                                         | 0000h         | 0000h  | 0000h  |
| 0Ch     | General Purpose MTP <sup>(1)</sup>               | —             | —      | —      |
| 0Dh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |
| 0Eh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |
| 0Fh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |

| Address | Function                                         | POR/BOR Value |        |        |
|---------|--------------------------------------------------|---------------|--------|--------|
|         |                                                  | 8-bit         | 10-bit | 12-bit |
| 10h     | Nonvolatile DAC0 Wiper Register <sup>(1)</sup>   | 7Fh           | 1FFh   | 7FFh   |
| 11h     | Nonvolatile DAC1 Wiper Register <sup>(1)</sup>   | 7Fh           | 1FFh   | 7FFh   |
| 12h     | Reserved                                         | —             | —      | —      |
| 13h     | Reserved                                         | —             | —      | —      |
| 14h     | Reserved                                         | —             | —      | —      |
| 15h     | Reserved                                         | —             | —      | —      |
| 16h     | Reserved                                         | —             | —      | —      |
| 17h     | Reserved                                         | —             | —      | —      |
| 18h     | Nonvolatile VREF register <sup>(1)</sup>         | 0000h         | 0000h  | 0000h  |
| 19h     | Nonvolatile Power-Down Register <sup>(1)</sup>   | 0000h         | 0000h  | 0000h  |
| 1Ah     | NV Gain <sup>(1)</sup>                           | 0000h         | 0000h  | 0000h  |
| 1Bh     | NV WiperLock™ Technology Register <sup>(1)</sup> | 0000h         | 0000h  | 0000h  |
| 1Ch     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |
| 1Dh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |
| 1Eh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |
| 1Fh     | General Purpose MTP <sup>(1)</sup>               | 0000h         | 0000h  | 0000h  |

Legend:

|  |                                  |
|--|----------------------------------|
|  | Volatile memory address range    |
|  | Nonvolatile memory address range |
|  | Not implemented                  |

**Note 1:** On nonvolatile devices only (MCP48CMDXX).

**2:** The '1' bit is the POR status bit, which is set after the POR event and cleared after address 0Ah is read.

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## 4.2.6 DEVICE REGISTERS

Register 4-1 shows the format of the DAC Output Value registers for the volatile memory locations. These registers will be either 8 bits, 10 bits or 12 bits wide. The values are right justified.

**REGISTER 4-1: DAC0 (00H/10H) AND DAC1 (01H/11H) OUTPUT VALUE REGISTERS (VOLATILE/NONVOLATILE)**

|        | U-0    | U-0 | U-0 | U-0 | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n | R/W-n |
|--------|--------|-----|-----|-----|-------|-------|-------|-------|-------|-------|-------|-------|-------|-------|-------|-------|
| 12-bit | —      | —   | —   | —   | D11   | D10   | D09   | D08   | D07   | D06   | D05   | D04   | D03   | D02   | D01   | D00   |
| 10-bit | —      | —   | —   | —   | —     | —     | D09   | D08   | D07   | D06   | D05   | D04   | D03   | D02   | D01   | D00   |
| 8-bit  | —      | —   | —   | —   | —     | —     | —     | —     | D07   | D06   | D05   | D04   | D03   | D02   | D01   | D00   |
|        | bit 15 |     |     |     | bit 0 |       |       |       |       |       |       |       |       |       |       |       |

### Legend:

R = Readable bit      W = Writable bit      U = Unimplemented bit, read as '0'  
 -n = Value at POR      '1' = Bit is set      '0' = Bit is cleared      x = Bit is unknown  
 = 12-bit device       = 10-bit device       = 8-bit device

### 12-bit      10-bit      8-bit

bit 15-12    bit 15-10    bit 15-8    **Unimplemented:** Read as '0'

bit 11-0    —    —    **D11-D00:** DAC Output value – 12-bit devices

FFFh = Full-Scale output value

7FFh = Mid-Scale output value

000h = Zero-Scale output value

—    bit 9-0    —    **D09-D00:** DAC Output value – 10-bit devices

3FFh = Full-Scale output value

1FFh = Mid-Scale output value

000h = Zero-Scale output value

—    —    bit 7-0    **D07-D00:** DAC Output value – 8-bit devices

FFh = Full-Scale output value

7Fh = Mid-Scale output value

00h = Zero-Scale output value

**Note 1:** Unimplemented bit, read as '0'.

Register 4-2 shows the format of the Voltage Reference Control register. Each DAC has two bits to control the source of the voltage reference of the DAC. This register is for the volatile memory locations. The width of this register is two times the number of DACs for the device.

## REGISTER 4-2: VOLTAGE REFERENCE (VREF) CONTROL REGISTERS (08h/18h) (VOLATILE/NONVOLATILE)

|        | U-0    | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-n            | R/W-n            | R/W-n | R/W-n |
|--------|--------|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|------------------|------------------|-------|-------|
| Single | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | — <sup>(1)</sup> | — <sup>(1)</sup> | VR0B  | VR0A  |
| Dual   | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | VR1B             | VR1A             | VR0B  | VR0A  |
|        | bit 15 |     |     |     |     |     |     |     |     |     |     | bit 0            |                  |       |       |

### Legend:

R = Readable bit

W = Writable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

  = Single channel device

  = Dual channel device

### Single Dual

bit 15-2 bit 15-4 **Unimplemented:** Read as '0'

bit 1-0 bit 3-0

**VRxB-VRxA:** DAC Voltage Reference Control bits

11 = V<sub>REF</sub> pin (Buffered); V<sub>REF</sub> buffer enabled

10 = V<sub>REF</sub> pin (Unbuffered); V<sub>REF</sub> buffer disabled

01 = Internal band gap<sup>(2)</sup> (1.214V typical); V<sub>REF</sub> buffer enabled. V<sub>REF</sub> voltage driven when powered-down.

00 = V<sub>DD</sub> (Unbuffered); V<sub>REF</sub> buffer disabled. Use this state with power-down bits for lowest current.

**Note 1:** Unimplemented bit, read as '0'.

**2:** When the internal band gap is selected, the band gap voltage source will continue to output the voltage on the V<sub>REF</sub> pin in any of the Power-Down modes. To reduce the power consumption to its lowest level (band gap disabled), after selecting the desired Power-Down mode, the voltage reference must be changed to V<sub>DD</sub> or V<sub>REF</sub> pin unbuffered ('00' or '10'), which turns off the internal band gap circuitry. After wake-up, the user needs to reselect the internal band gap ('01') for the voltage reference source.

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Register 4-3 shows the format of the Power-Down Control register. Each DAC has two bits to control the Power-Down state of the DAC. This register is for the volatile and nonvolatile memory locations. The width of this register is two times the number of DACs for the device.

**REGISTER 4-3: POWER-DOWN CONTROL REGISTERS (09h/19h)  
(VOLATILE/NONVOLATILE)**

|        | U-0    | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-n            | R/W-n            | R/W-n | R/W-n |
|--------|--------|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|------------------|------------------|-------|-------|
| Single | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | — <sup>(1)</sup> | — <sup>(1)</sup> | PD0B  | PD0A  |
| Dual   | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | PD1B             | PD1A             | PD0B  | PD0A  |
|        | bit 15 |     |     |     |     |     |     |     |     |     |     | bit 0            |                  |       |       |

**Legend:**

R = Readable bit      W = Writable bit      U = Unimplemented bit, read as '0'  
 -n = Value at POR      '1' = Bit is set      '0' = Bit is cleared      x = Bit is unknown  
 [Single Channel Device Icon] = Single channel device      [Dual Channel Device Icon] = Dual channel device

**Single**

bit 15-2

bit 1-0

**Dual**

bit 15-4

bit 3-0

**Unimplemented:** Read as '0'

**PDxB-PDxA:** DAC Power-Down Control bits<sup>(2)</sup>

11 =Powered-Down –  $V_{OUT}$  is open circuit

10 =Powered-Down –  $V_{OUT}$  is loaded with a 100 k $\Omega$  resistor to ground

01 =Powered-Down –  $V_{OUT}$  is loaded with a 1 k $\Omega$  resistor to ground

00 =Normal Operation (not powered-down)

**Note 1:** Unimplemented bit, read as '0'.

**2:** See Table 5-5 for more details.

Register 4-4 shows the format of the Gain Control and System Status register. Each DAC has one bit to control the gain of the DAC and two Status bits.

## REGISTER 4-4: GAIN CONTROL AND SYSTEM STATUS REGISTER (0Ah) (VOLATILE)

|        | U-0    | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-n            | R/W-n | R/C-1 | R     | U-0 | U-0 | U-0 | U-0 | U-0 |
|--------|--------|-----|-----|-----|-----|-----|------------------|-------|-------|-------|-----|-----|-----|-----|-----|
| Single | —      | —   | —   | —   | —   | —   | — <sup>(1)</sup> | G0    | POR   | MTPMA | —   | —   | —   | —   | —   |
| Dual   | —      | —   | —   | —   | —   | —   | G1               | G0    | POR   | MTPMA | —   | —   | —   | —   | —   |
|        | bit 15 |     |     |     |     |     | bit 0            |       |       |       |     |     |     |     |     |

### Legend:

R = Readable bit      W = Writable bit      C = Clearable bit      U = Unimplemented bit, read as '0'  
 -n = Value at POR      '1' = Bit is set      '0' = Bit is cleared      x = Bit is unknown  
 = Single channel device       = Dual channel device

| Single   | Dual      |                                                                                                                                                                                                                                                                                                                                                                                                                             |
|----------|-----------|-----------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| bit 15-9 | bit 15-10 | <b>Unimplemented:</b> Read as '0'                                                                                                                                                                                                                                                                                                                                                                                           |
| —        | bit 9     | <b>G1:</b> DAC1 Output Driver Gain control bits<br>1 = 2x Gain. Not applicable when $V_{DD}$ is used as $V_{RL}$ <sup>(2)</sup><br>0 = 1x Gain                                                                                                                                                                                                                                                                              |
| bit 8    | bit 8     | <b>G0:</b> DAC0 Output Driver Gain control bits<br>1 = 2x Gain. Not applicable when $V_{DD}$ is used as $V_{RL}$ <sup>(2)</sup><br>0 = 1x Gain                                                                                                                                                                                                                                                                              |
| bit 7    | bit 7     | <b>POR:</b> Power-on Reset (Brown-out Reset) Status bit<br>This bit indicates if a POR or BOR event has occurred since the last read command of this register. Reading this register clears the state of the POR Status bit.<br>1 = A POR (BOR) event occurred since the last read of this register. Reading this register clears this bit.<br>0 = A POR (BOR) event has not occurred since the last read of this register. |
| bit 6    | bit 6     | <b>MTPMA:</b> MTP Memory Access Status bit <sup>(3)</sup><br>This bit indicates if the MTP Memory Access is occurring.<br>1 = An MTP Memory Access is currently occurring (during the POR MTP read cycle or an MTP write cycle is occurring). Only serial commands addressing the volatile memory are allowed.<br>0 = An MTP Memory Access is NOT currently occurring                                                       |
| bit 5-0  | bit 5-0   | <b>Unimplemented:</b> Read as '0'                                                                                                                                                                                                                                                                                                                                                                                           |

**Note 1:** Unimplemented bit, read as '0'.

**2:** The DAC's Gain bit is ignored and the gain is forced to 1x ( $G_x = 0$ ) when the DAC voltage reference is selected as  $V_{DD}$  ( $VRxB:VRxA = 00$ ).

**3:** For volatile memory devices, this bit is read as '0'.

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Register 4-5 shows the format of the Nonvolatile Gain Control register. Each DAC has one bit to control the gain of the DAC.

## REGISTER 4-5: GAIN CONTROL REGISTER (1Ah) (NONVOLATILE)

|        | U-0    | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-n            | R/W-n | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 |
|--------|--------|-----|-----|-----|-----|-----|------------------|-------|-----|-----|-----|-----|-----|-----|-----|
| Single | —      | —   | —   | —   | —   | —   | — <sup>(1)</sup> | G0    | —   | —   | —   | —   | —   | —   | —   |
| Dual   | —      | —   | —   | —   | —   | —   | G1               | G0    | —   | —   | —   | —   | —   | —   | —   |
|        | bit 15 |     |     |     |     |     | bit 0            |       |     |     |     |     |     |     |     |

### Legend:

R = Readable bit      W = Writable bit      C = Clearable bit      U = Unimplemented bit, read as '0'  
 -n = Value at POR      '1' = Bit is set      '0' = Bit is cleared      x = Bit is unknown  
 = Single channel device       = Dual channel device

### Single

bit 15-10

bit 9-8

bit 7-0

### Dual

bit 15-10

bit 9-8

bit 7-0

**Unimplemented:** Read as '0'

**Gx<sup>(2)</sup>:** DAC Output Driver Gain control bits

1 = 2x Gain

0 = 1x Gain

**Unimplemented:** Read as '0'

**Note 1:** Unimplemented bit, read as '0'.

**Note 2:** When the DAC voltage reference is selected as  $V_{DD}$  ( $VRxB:VRxA = 00$ ), the DAC's Gain bit is ignored and the gain is forced to 1x ( $Gx = 0$ ).

Register 4-6 shows the format of the DAC WiperLock Technology Status register. The width of this register is two times the number of DACs for the device.

WiperLock Technology bits only control access to volatile memory. Nonvolatile memory write access is controlled by the requirement of high voltage on the HVC pin, which is recommended to not be available during normal device operation.

## REGISTER 4-6: WIPERLOCK™ TECHNOLOGY CONTROL REGISTER (1BH) (NONVOLATILE)

|        | U-0    | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | U-0 | R/W-n            | R/W-n            | R/W-n | R/W-n |
|--------|--------|-----|-----|-----|-----|-----|-----|-----|-----|-----|-----|------------------|------------------|-------|-------|
| Single | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | — <sup>(1)</sup> | — <sup>(1)</sup> | WL0B  | WL0A  |
| Dual   | —      | —   | —   | —   | —   | —   | —   | —   | —   | —   | —   | WL1B             | WL1A             | WL0B  | WL0A  |
|        | bit 15 |     |     |     |     |     |     |     |     |     |     | bit 0            |                  |       |       |

### Legend:

R = Readable bit

W = Writable bit

C = Clearable bit

U = Unimplemented bit, read as '0'

-n = Value at POR

'1' = Bit is set

'0' = Bit is cleared

x = Bit is unknown

  = Single channel device

  = Dual channel device

### Single

bit 15-2

bit 1-0

### Dual

bit 15-4

bit 3-0

**Unimplemented:** Read as '0'

**WLXB-WLXA:** WiperLock™ Technology Status bits<sup>(2)</sup>

11 = Vol. DAC Wiper register and Vol. DAC configuration bits are locked

10 = Vol. DAC Wiper register is locked, and Vol. DAC configuration bits are unlocked

01 = Vol. DAC Wiper register is unlocked, and Vol. DAC configuration bits are locked

00 = Vol. DAC Wiper register and Vol. DAC configuration bits are unlocked

**Note 1:** Unimplemented bit, read as '0'.

**2:** The volatile PDxB:PDxA bits are NOT locked due to the requirement of being able to exit Power-Down mode.

# MCP48CXX1/2

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NOTES:

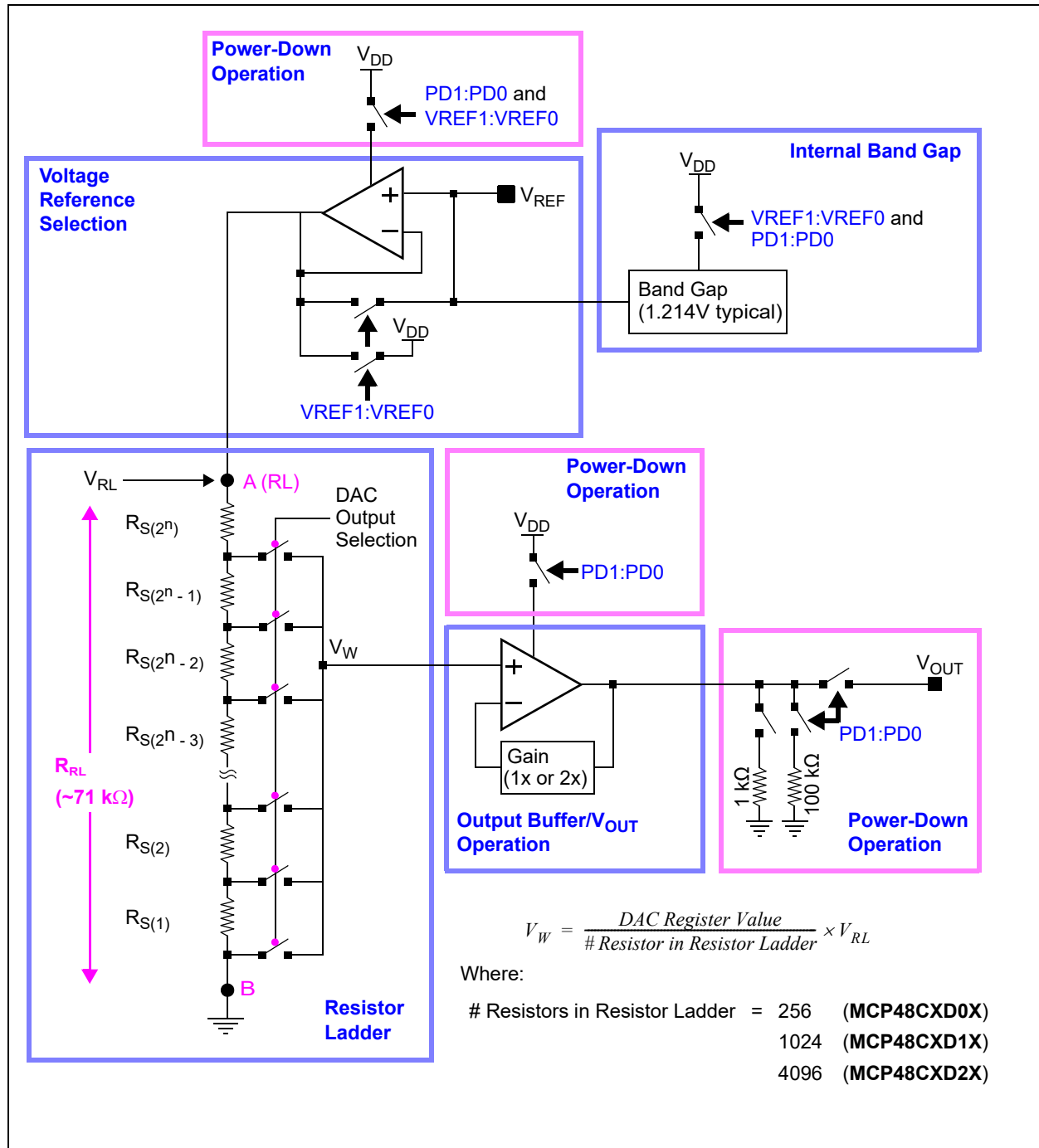
## 5.0 DAC CIRCUITRY

The Digital to Analog Converter circuitry converts a digital value into its analog representation. The description describes the functional operation of the device.

The DAC circuit uses a resistor ladder implementation. Devices have up to two DACs. The figure below shows the functional block diagram for the MCP48CXDX1/2 DAC circuitry.

The functional blocks of the DAC include:

- **Resistor Ladder**
- **Voltage Reference Selection**
- **Output Buffer/V<sub>OUT</sub> Operation**
- **Latch Pin (LAT)**
- **Power-Down Operation**



**FIGURE 5-1:** MCP48CXDX1/2 DAC Module Block Diagram.

# MCP48CXDX1/2

## 5.1 Resistor Ladder

The resistor ladder is a digital potentiometer with the A Terminal connected to the selected reference voltage and the B Terminal internally grounded (see Figure 5-2). The volatile DAC register controls the wiper position. The wiper voltage ( $V_W$ ) is proportional to the DAC register value divided by the number of resistor elements ( $R_S$ ) in the ladder (256, 1024 or 4096) related to the  $V_{RL}$  voltage.

The output of the resistor network will drive the input of an output buffer.

The Resistor Network is made up of three parts:

- Resistor ladder (string of  $R_S$  elements)
- Wiper switches
- DAC register decode

The resistor ladder has a typical impedance ( $R_{RL}$ ) of approximately 71 k $\Omega$ . This resistor ladder resistance ( $R_{RL}$ ) may vary from device to device up to  $\pm 10\%$ . This is a voltage divider configuration; therefore, the actual  $R_{RL}$  resistance does not affect the output, given a fixed voltage at  $V_{RL}$ .

Equation 5-1 shows the calculation for the step resistance.

**Note:** The maximum wiper position is  $2^n - 1$ , while the number of resistors in the resistor ladder is  $2^n$ . This means that when the DAC register is at full scale, there is one resistor element ( $R_S$ ) between the wiper and the  $V_{RL}$  voltage.

If the unbuffered  $V_{REF}$  pin is used as the  $V_{RL}$  voltage source, the external voltage source must have a low output impedance.

When the DAC is powered down, the resistor ladder is disconnected from the selected reference voltage.

### EQUATION 5-1: $R_S$ CALCULATION

|                               |               |
|-------------------------------|---------------|
| $R_S = \frac{R_{RL}}{(256)}$  | 8-bit Device  |
| $R_S = \frac{R_{RL}}{(1024)}$ | 10-bit Device |
| $R_S = \frac{R_{RL}}{(4096)}$ | 12-bit Device |

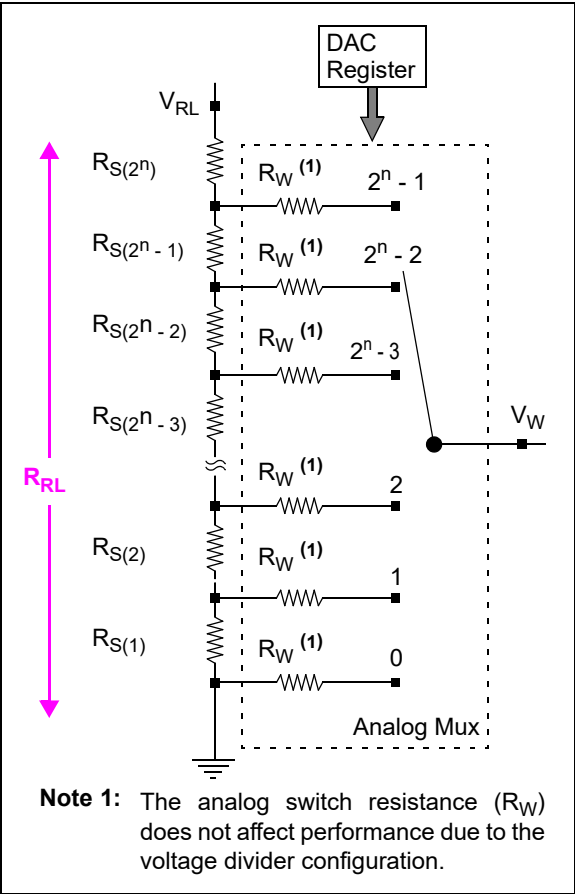


FIGURE 5-2: Resistor Ladder Model Block Diagram.

## 5.2 Voltage Reference Selection

The resistor ladder has up to four sources for the reference voltage. The selection of the voltage reference source is specified with the volatile VREF1:VREF0 configuration bits (see Register 4-2). The selected voltage source is connected to the V<sub>RL</sub> node (see Figure 5-3 and Figure 5-4).



**FIGURE 5-3:** Resistor Ladder Reference Voltage Selection Block Diagram.

The four voltage source options for the resistor ladder are:

1. V<sub>DD</sub> pin voltage
2. Internal band gap voltage reference (V<sub>BG</sub>)
3. V<sub>REF</sub> pin voltage – unbuffered
4. V<sub>REF</sub> pin voltage – internally buffered

On a POR/BOR event, the default configuration state or the value written in the nonvolatile register is latched into the volatile VREF1:VREF0 configuration bits.

If the V<sub>REF</sub> pin is used with an external voltage source, then the user must select between Buffered or Unbuffered mode.

### 5.2.1 USING V<sub>DD</sub> AS V<sub>REF</sub>

When the user selects the V<sub>DD</sub> as the reference, the V<sub>REF</sub> pin voltage is not connected to the resistor ladder. The V<sub>DD</sub> voltage is internally connected to the resistor ladder.

### 5.2.2 USING AN EXTERNAL V<sub>REF</sub> SOURCE IN UNBUFFERED MODE

In this case, the V<sub>REF</sub> pin voltage may vary from V<sub>SS</sub> to V<sub>DD</sub>. The voltage source must have a low-output impedance. If the voltage source has a high-output impedance, the voltage on the V<sub>REF</sub> pin can be lower than expected. The resistor ladder has a typical impedance of 71 kΩ and a typical capacitance of 29 pF.

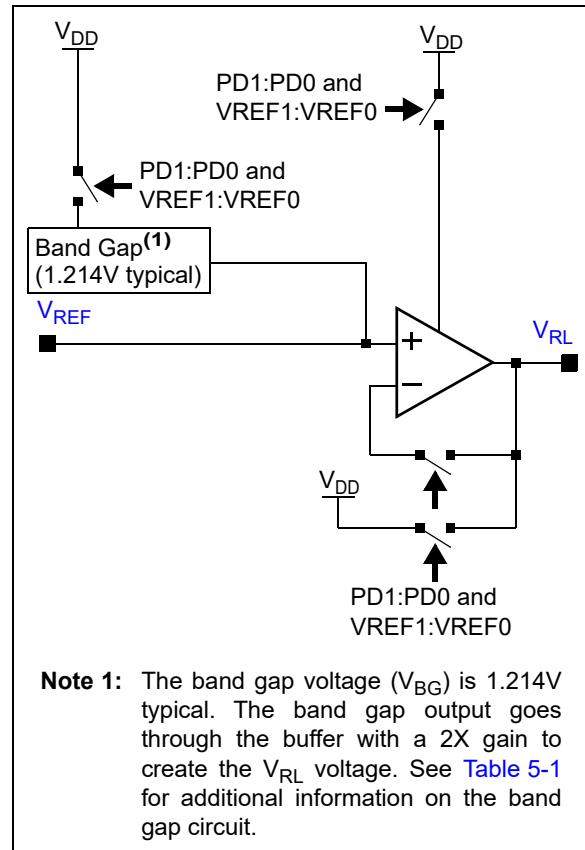
If a single V<sub>REF</sub> pin is supplying multiple DACs, the V<sub>REF</sub> pin source must have adequate current capability to support the number of DACs. It must be assumed that the resistor ladder resistance (R<sub>RL</sub>) of each DAC is at the minimum specified resistance and these resistances are in parallel.

If the V<sub>REF</sub> pin is tied to the V<sub>DD</sub> voltage, selecting the V<sub>DD</sub> Reference mode (VREF1:VREF0 = 00) is recommended.

### 5.2.3 USING AN EXTERNAL V<sub>REF</sub> SOURCE IN BUFFERED MODE

The V<sub>REF</sub> pin voltage may be from 0V to V<sub>DD</sub>. The input buffer (amplifier) provides low offset voltage, low noise and a very high input impedance, with only minor limitations on the input range and frequency response.

Any variation or noises on the reference source can directly affect the DAC output. The reference voltage needs to be as clean as possible for accurate DAC performance.



**FIGURE 5-4:** Reference Voltage Selection Implementation Block Diagram.

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## 5.2.4 USING THE INTERNAL BAND GAP AS VOLTAGE REFERENCE

The internal band gap is designed to drive the resistor ladder buffer.

If the internal band gap is selected, the band gap voltage source will drive the external  $V_{REF}$  pins. The  $V_{REF1}$  pin must be left unloaded in this mode. The voltage reference source can be independently selected on devices with two DAC channels, but restrictions apply:

- The  $V_{DD}$  mode can be used without issues on any channel.
- When the internal band gap is selected as the voltage source, all the  $V_{REF}$  pins are connected to its output. The use of the Unbuffered mode is only possible on  $V_{REF0}$ , because it is the only one that can be loaded.
- When using the Internal Band Gap mode on channel 0, channel 1 must be put in Buffered External  $V_{REF}$  mode or  $V_{DD}$  Reference mode and the  $V_{REF1}$  pin must be left unloaded.

The resistance of the resistor ladder ( $R_{RL}$ ) is targeted to be 71 k $\Omega$  ( $\pm 10\%$ ), which means a minimum resistance of 63.9 k $\Omega$ .

The band gap selection can be used across the  $V_{DD}$  voltages while maximizing the  $V_{OUT}$  voltage ranges. For  $V_{DD}$  voltages below the Gain \*  $V_{BG}$  voltage, the output for the upper codes will be clipped to the  $V_{DD}$  voltage. The table below shows the maximum DAC register code given device  $V_{DD}$  and Gain bit setting.

**TABLE 5-1:  $V_{OUT}$  USING BAND GAP**

| $V_{DD}$ | DAC Gain         | Max DAC Code <sup>(1)</sup> |        |       | Comment                       |
|----------|------------------|-----------------------------|--------|-------|-------------------------------|
|          |                  | 12-bit                      | 10-bit | 8-bit |                               |
| 5.5      | 1                | FFFh                        | 3FFh   | FFh   | $V_{OUT(max)} = 1.214V^{(3)}$ |
|          | 2                | FFFh                        | 3FFh   | FFh   | $V_{OUT(max)} = 2.428V^{(3)}$ |
| 2.7      | 1                | FFFh                        | 3FFh   | FFh   | $V_{OUT(max)} = 1.214V^{(3)}$ |
|          | 2                | FFFh                        | 3FFh   | FFh   | $V_{OUT(max)} = 2.428V$       |
| 1.8      | 1                | FFFh                        | 3FFh   | FFh   | $V_{OUT(max)} = 1.214V$       |
|          | 2 <sup>(2)</sup> | BBCh                        | 2EFh   | BBh   | 1.8V                          |

- Note**
- 1: Without the  $V_{OUT}$  pin voltage being clipped.
  - 2: Recommended to use the Gain = 1 setting.
  - 3: When  $V_{BG} = 1.214V$  typical.

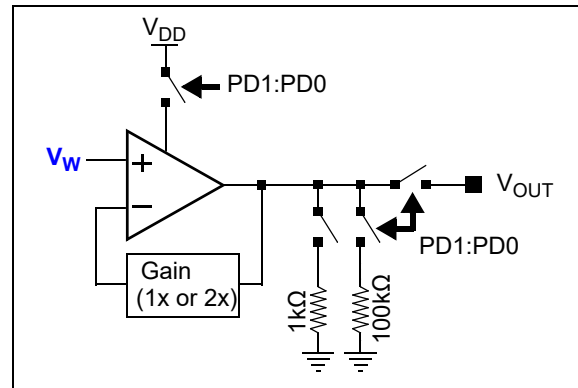
## 5.3 Output Buffer/ $V_{OUT}$ Operation

The output driver buffers the wiper voltage ( $V_W$ ) of the resistor ladder.

The DAC output is buffered with a low-power, precision output amplifier with selectable gain. This amplifier provides a rail-to-rail output with low offset voltage and low noise. The amplifier's output can drive the resistive and high-capacitive loads without oscillation. The amplifier provides a maximum load current that is enough for most programmable voltage reference applications. Refer to [Section 1.0 "Electrical Characteristics"](#) for the specifications of the output amplifier.

**Note:** The load resistance must be kept higher than 2 k $\Omega$  to maintain stability of the analog output and have it meet electrical specifications.

The figure below shows a block diagram of the output driver circuit.



**FIGURE 5-5:** Output Driver Block Diagram.

Power-Down logic also controls the output buffer operation (see [Section 5.5 "Power-Down Operation"](#) for additional information on Power-Down). In any of the three Power-Down modes, the output amplifier is powered down and its output becomes a high impedance to the  $V_{OUT}$  pin.

### 5.3.1 PROGRAMMABLE GAIN

The amplifier's gain is controlled by the Gain (G) Configuration bit (see [Register 4-4](#)) and the  $V_{RL}$  reference selection (see [Register 4-2](#)).

The Gain options are:

- a) Gain of '1', with either the  $V_{DD}$  or  $V_{REF}$  pin used as reference voltage.
- b) Gain of '2', only when the  $V_{REF}$  pin or the internal band gap is used as reference voltage. The  $V_{REF}$  pin voltage must be limited to  $V_{DD}/2$ . When the reference voltage selection ( $V_{RL}$ ) is the device's  $V_{DD}$  voltage, the G bit is ignored and a gain of '1' is used.

The table below shows the gain bit operation.

**TABLE 5-2: OUTPUT DRIVER GAIN**

| Gain Bit | Gain | Comment                                                           |
|----------|------|-------------------------------------------------------------------|
| 0        | 1    |                                                                   |
| 1        | 2    | Limits $V_{REF}$ pin voltages relative to device $V_{DD}$ voltage |

The volatile G bit value can be modified by:

- POR event
- BOR event
- SPI write commands

## 5.3.2 OUTPUT VOLTAGE

The volatile DAC register values, along with the device's configuration bits, control the analog  $V_{OUT}$  voltage. The volatile DAC register's value is unsigned binary. The formula for the output voltage is provided in the equation below. Examples of volatile DAC register values and the corresponding theoretical  $V_{OUT}$  voltage for the MCP48CXDX1/2 devices are shown in [Table 5-6](#).

**EQUATION 5-2: CALCULATING OUTPUT VOLTAGE ( $V_{OUT}$ )**

$$V_{OUT} = \frac{V_{RL} \times \text{DAC Register Value}}{\# \text{ Resistor in Resistor Ladder}} \times \text{Gain}$$

Where:

# Resistors in R-Ladder = 4096 (MCP48CXDX2X)  
 1024 (MCP48CXDX1X)  
 256 (MCP48CXDX0X)

When Gain = 2 ( $V_{RL} = V_{REF}$ ), if  $V_{REF} > V_{DD}/2$ , the  $V_{OUT}$  voltage is limited to  $V_{DD}$ . So, if  $V_{REF} = V_{DD}$ , the  $V_{OUT}$  voltage does not change for volatile DAC register values mid-scale and greater because the output amplifier is at full-scale output.

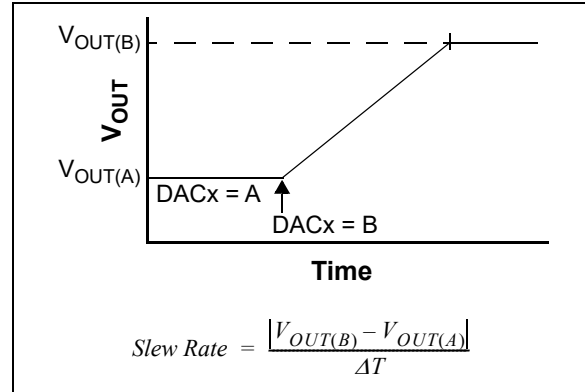
The following events update the DAC register value and, therefore, the analog voltage output ( $V_{OUT}$ ):

- Power-on Reset
- Brown-out Reset
- SPI write command (to volatile registers)

Next, the  $V_{OUT}$  voltage starts driving to the new value after the event has occurred.

## 5.3.3 OUTPUT SLEW RATE

The figure below shows an example of the slew rate of the  $V_{OUT}$  pin. The slew rate can be affected by the characteristics of the circuit connected to the  $V_{OUT}$  pin.



**FIGURE 5-6:  $V_{OUT}$  Pin Slew Rate.**

### 5.3.3.1 Small Capacitive Load

With a small capacitive load, the output buffer's current is not affected by the capacitive load ( $C_L$ ). But still, the  $V_{OUT}$  pin's voltage is not a step transition from one output value (DAC register value) to the next output value. The change of the  $V_{OUT}$  voltage is limited by the output buffer's characteristics, so the  $V_{OUT}$  pin voltage will have a slope from the old voltage to the new voltage. This slope is fixed for the output buffer, and is referred to as the buffer slew rate ( $SR_{BUF}$ ).

### 5.3.3.2 Large Capacitive Load

With a larger capacitive load, the slew rate is determined by two factors:

- The output buffer's short-circuit current ( $I_{SC}$ )
- The  $V_{OUT}$  pin's external load

$I_{OUT}$  cannot exceed the output buffer's short-circuit current ( $I_{SC}$ ), which fixes the output buffer slew rate ( $SR_{BUF}$ ). The voltage on the capacitive load ( $C_L$ ),  $V_{CL}$ , changes at a rate proportional to  $I_{OUT}$ , which fixes a capacitive load slew rate ( $SR_{CL}$ ).

So the  $V_{CL}$  voltage slew rate is limited to the slower of the output buffer's internally set slew rate ( $SR_{BUF}$ ) and the capacitive load slew rate ( $SR_{CL}$ ).

## 5.3.4 DRIVING RESISTIVE AND CAPACITIVE LOADS

The  $V_{OUT}$  pin can drive up to 100 pF of capacitive load in parallel with a 5-k $\Omega$  resistive load (to meet electrical specifications).  $V_{OUT}$  drops slowly as the load resistance decreases after about 3.5 k $\Omega$ . It is recommended that a load with  $R_L$  greater than 2 k $\Omega$  be used.

Refer to the Characterization Data documents for a detailed  $V_{OUT}$  vs. resistive load characterization graph.

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Driving large capacitive loads can cause stability problems for voltage feedback output amplifiers. As the load capacitance increases, the feedback loop's phase margin decreases and the closed-loop bandwidth is reduced. This produces gain peaking in the frequency response with overshoot and ringing in the step response. That is, the  $V_{OUT}$  pin's voltage does not quickly follow the buffer's input voltage (due to the large capacitive load); therefore, the output buffer will overshoot the desired target voltage. Once the driver detects this overshoot, it compensates by forcing it to a voltage below the target. This causes voltage ringing on the  $V_{OUT}$  pin.

So, when driving large capacitive loads with the output buffer, a small series resistor ( $R_{ISO}$ ) at the output (see the figure below) improves the output buffer's stability (feedback loop's phase margin) by making the output load resistive at higher frequencies. The bandwidth will generally be lower than the bandwidth with no capacitive load.



**FIGURE 5-7:** Circuit to Stabilize Output Buffer for Large Capacitive Loads ( $C_L$ ).

The  $R_{ISO}$  resistor value for your circuit needs to be selected. The resulting frequency response peaking and step response overshoot for this  $R_{ISO}$  resistor value must be verified on the bench. Modify the  $R_{ISO}$ 's resistance value until the output characteristics meet your requirements.

A method to evaluate the system's performance is to inject a step voltage on the  $V_{REF}$  pin and observe the  $V_{OUT}$  pin's characteristics.

**Note:** Additional insight into circuit design for driving capacitive loads can be found in AN884 – “Driving Capacitive Loads With Op Amps” (DS00000884).

### 5.3.5 STEP VOLTAGE ( $V_S$ )

The step voltage depends on the device resolution and the calculated output voltage range. 1 LSB is defined as the ideal voltage difference between two successive codes. The step voltage can easily be calculated using the equation below (the DAC register value is equal to 1). Theoretical step voltages are shown in Table 5-3 for several  $V_{REF}$  voltages.

#### EQUATION 5-3: $V_S$ CALCULATION

$$V_S = \frac{V_{RL}}{\# \text{ Resistor in Resistor Ladder}} \times \text{Gain}$$

Where:

|                         |   |      |          |
|-------------------------|---|------|----------|
| # Resistors in R-Ladder | = | 4096 | (12-bit) |
|                         |   | 1024 | (10-bit) |
|                         |   | 256  | (8-bit)  |

**TABLE 5-3:** THEORETICAL STEP VOLTAGE ( $V_S$ )<sup>(1)</sup>

| Step Voltage | $V_{REF}$ |             |             |             |             | #bits  |
|--------------|-----------|-------------|-------------|-------------|-------------|--------|
|              | 5.0       | 2.7         | 1.8         | 1.5         | 1.0         |        |
| $V_S$        | 1.22 mV   | 659 $\mu$ V | 439 $\mu$ V | 366 $\mu$ V | 244 $\mu$ V | 12-bit |
|              | 4.88 mV   | 2.64 mV     | 1.76 mV     | 1.46 mV     | 977 $\mu$ V | 10-bit |
|              | 19.5 mV   | 10.5 mV     | 7.03 mV     | 5.86 mV     | 3.91 mV     | 8-bit  |

**Note 1:** When Gain = 1X,  $V_{FS} = V_{RL}$  and  $V_{ZS} = 0V$ .

## 5.4 Latch Pin ( $\overline{\text{LAT}}$ )

The Latch pin controls when the volatile DAC register value is transferred to the DAC wiper. This is useful for applications that need to synchronize the wiper(s) updates to an external event, such as zero crossing or updates to the other wipers on the device. The  $\overline{\text{LAT}}$  pin is asynchronous to the serial interface operation.

When the  $\overline{\text{LAT}}$  pin is high, transfers from the volatile DAC register to the DAC wiper are inhibited. The volatile DAC register value(s) can continue to be updated.

When the  $\overline{\text{LAT}}$  pin is low, the volatile DAC register value is transferred to the DAC wiper.

**Note:** This allows both the volatile DAC0 and DAC1 registers to be updated while the  $\overline{\text{LAT}}$  pin is high and to have outputs synchronously updated as the  $\overline{\text{LAT}}$  pin is driven low.

The figure below shows the interaction of the  $\overline{\text{LAT}}$  pin and the loading of the DAC wiper x (from the volatile DAC register x). The transfers are level driven. If the  $\overline{\text{LAT}}$  pin is held low, the corresponding DAC wiper is updated as soon as the volatile DAC register value is updated.

The  $\overline{\text{LAT}}$  pin allows the DAC wiper to be updated to an external event and to have multiple DAC channels/devices update at a common event.



**FIGURE 5-8:**  $\overline{\text{LAT}}$  and DAC Interaction.

The DAC wiper x is updated from the volatile DAC register x; therefore, all DACs that are associated with a given  $\overline{\text{LAT}}$  pin can be updated synchronously.

If the application does not require synchronization, this signal must be tied low.

The figure below shows two cases of using the  $\overline{\text{LAT}}$  pin to control when the wiper register is updated relative to the value of a sine wave signal.



**FIGURE 5-9:** Example Use of  $\overline{\text{LAT}}$  Pin Operation.

## 5.5 Power-Down Operation

To allow the application to conserve power when DAC operation is not required, three Power-Down modes are available. On devices with multiple DACs, each DAC's Power-Down mode is individually controllable.

All Power-Down modes do the following:

- Turn off most of the DAC module's internal circuits
- Op amp is powered down and the  $V_{\text{OUT}}$  pin becomes high-impedance

- Retain the value of the volatile DAC register and configuration bits

Depending on the selected Power-Down mode, the following will occur:

- $V_{\text{OUT}}$  pin is switched to one of the two resistive pull-downs:
  - 100 k $\Omega$  (typical)
  - 1 k $\Omega$  (typical)

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The Power-Down configuration bits (PD1:PD0) control the power-down operation, shown in the table below.

**TABLE 5-4: POWER-DOWN BITS AND OUTPUT RESISTIVE LOAD**

| PD1 | PD0 | Function                  |
|-----|-----|---------------------------|
| 0   | 0   | Normal operation          |
| 0   | 1   | 1 kΩ resistor to ground   |
| 1   | 0   | 100 kΩ resistor to ground |
| 1   | 1   | Open circuit              |

There is a delay ( $T_{PDD}$ ) between the PD1:PD0 bits changing from '00' to either '01', '10' or '11' and the op amp no longer driving the  $V_{OUT}$  output and the pull-down resistors sinking current.

In any of the Power-Down modes where the  $V_{OUT}$  pin is not externally connected (sinking or sourcing current), as the number of DACs increases, the device's power-down current will also increase.

The following table shows the current sources for the DAC based on the selected source of the DAC's reference voltage and if the device is in normal operating mode or one of the Power-Down modes.

**TABLE 5-5: DAC CURRENT SOURCES**

| Device<br>$V_{DD}$<br>Current<br>Source | PDxB:xA = 00,<br>VREFx:B:xA = |    |                  |    | PDxB:xA ≠ 00,<br>VREFx:B:xA = |                  |                  |                  |
|-----------------------------------------|-------------------------------|----|------------------|----|-------------------------------|------------------|------------------|------------------|
|                                         | 00                            | 01 | 10               | 11 | 00                            | 01               | 10               | 11               |
| Output<br>Op Amp                        | Y                             | Y  | Y                | Y  | N                             | N                | N                | N                |
| Resistor<br>Ladder                      | Y                             | Y  | N <sup>(1)</sup> | Y  | N                             | N                | N <sup>(1)</sup> | N                |
| VREF<br>Selection<br>Buffer             | N                             | Y  | N                | Y  | N                             | N                | N                | N                |
| Band Gap                                | N                             | Y  | N                | N  | N <sup>(2)</sup>              | Y <sup>(2)</sup> | N <sup>(2)</sup> | N <sup>(2)</sup> |

**Note 1:** The current is sourced from the  $V_{REF}$  pin, not the device  $V_{DD}$ .

**2:** If DAC0 and DAC1 are in one of the Power-Down modes, MTP write operations are not recommended.

The power-down bits are modified by using a write command to the volatile Power-Down register or a POR event, which transfers the nonvolatile Power-Down Register to the volatile Power-Down Register.

**Section 7.0 "Device Commands"** describes the SPI command for writing the power-down bits.

**Note 1:** The SPI serial interface circuit is not affected by the Power-Down mode. This circuit remains active in order to receive any command that might come from the SPI Host device.

## 5.5.1 EXITING POWER-DOWN

The following event changes the PD1:PD0 bits to '00' and, therefore, exits the Power-Down mode. This is any SPI write command where the PD1:PD0 bits are '00'.

When the device exits Power-Down mode, the following occurs:

- Disabled internal circuits are turned on
- Resistor ladder is connected to the selected reference voltage ( $V_{RL}$ )
- Selected pull-down resistor is disconnected
- The  $V_{OUT}$  output is driven to the voltage represented by the volatile DAC register's value and configuration bits

DAC Wiper register and DAC Wiper value may be different due to the DAC Wiper register being modified while the  $\overline{LAT}$  pin was driven to (and remaining at)  $V_{IH}$ .

The  $V_{OUT}$  output signal requires time as these circuits are powered-up and the output voltage is driven to the specified value as determined by the volatile DAC register and configuration bits.

**Note:** The op amp and resistor ladder were powered off (0V); therefore, the op amp's input voltage ( $V_W$ ) can be considered 0V. There is a delay ( $T_{PDE}$ ) between the PD1:PD0 bits updating to '00' and the op amp driving the  $V_{OUT}$  output. The op amp's settling time (from 0V) needs to be taken into account to ensure the  $V_{OUT}$  voltage reflects the selected value.

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**TABLE 5-6: DAC INPUT CODE VS. CALCULATED ANALOG OUTPUT ( $V_{OUT}$ ) ( $V_{DD} = 5.0V$ )**

| Device              | Volatile DAC Register Value | $V_{RL}^{(1)}$ | LSb         |          | Gain Selection <sup>(2)</sup> | $V_{OUT}^{(3)}$            |          |
|---------------------|-----------------------------|----------------|-------------|----------|-------------------------------|----------------------------|----------|
|                     |                             |                | Equation    | $\mu V$  |                               | Equation                   | V        |
| MCP48CVD2X (12-bit) | 1111 1111 1111              | 5.0V           | $5.0V/4096$ | 1,220.7  | 1x                            | $V_{RL} * (4095/4096) * 1$ | 4.998779 |
|                     |                             | 2.5V           | $2.5V/4096$ | 610.4    | 1x                            | $V_{RL} * (4095/4096) * 1$ | 2.499390 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (4095/4096) * 2$ | 4.998779 |
|                     | 0111 1111 1111              | 5.0V           | $5.0V/4096$ | 1,220.7  | 1x                            | $V_{RL} * (2047/4096) * 1$ | 2.498779 |
|                     |                             | 2.5V           | $2.5V/4096$ | 610.4    | 1x                            | $V_{RL} * (2047/4096) * 1$ | 1.249390 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (2047/4096) * 2$ | 2.498779 |
|                     | 0011 1111 1111              | 5.0V           | $5.0V/4096$ | 1,220.7  | 1x                            | $V_{RL} * (1023/4096) * 1$ | 1.248779 |
|                     |                             | 2.5V           | $2.5V/4096$ | 610.4    | 1x                            | $V_{RL} * (1023/4096) * 1$ | 0.624390 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (1023/4096) * 2$ | 1.248779 |
|                     | 0000 0000 0000              | 5.0V           | $5.0V/4096$ | 1,220.7  | 1x                            | $V_{RL} * (0/4096) * 1$    | 0        |
|                     |                             | 2.5V           | $2.5V/4096$ | 610.4    | 1x                            | $V_{RL} * (0/4096) * 1$    | 0        |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (0/4096) * 2$    | 0        |
| MCP48CVD1X (10-bit) | 11 1111 1111                | 5.0V           | $5.0V/1024$ | 4,882.8  | 1x                            | $V_{RL} * (1023/1024) * 1$ | 4.995117 |
|                     |                             | 2.5V           | $2.5V/1024$ | 2,441.4  | 1x                            | $V_{RL} * (1023/1024) * 1$ | 2.497559 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (1023/1024) * 2$ | 4.995117 |
|                     | 01 1111 1111                | 5.0V           | $5.0V/1024$ | 4,882.8  | 1x                            | $V_{RL} * (511/1024) * 1$  | 2.495117 |
|                     |                             | 2.5V           | $2.5V/1024$ | 2,441.4  | 1x                            | $V_{RL} * (511/1024) * 1$  | 1.247559 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (511/1024) * 2$  | 2.495117 |
|                     | 00 1111 1111                | 5.0V           | $5.0V/1024$ | 4,882.8  | 1x                            | $V_{RL} * (255/1024) * 1$  | 1.245117 |
|                     |                             | 2.5V           | $2.5V/1024$ | 2,441.4  | 1x                            | $V_{RL} * (255/1024) * 1$  | 0.622559 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (255/1024) * 2$  | 1.245117 |
|                     | 00 0000 0000                | 5.0V           | $5.0V/1024$ | 4,882.8  | 1x                            | $V_{RL} * (0/1024) * 1$    | 0        |
|                     |                             | 2.5V           | $2.5V/1024$ | 2,441.4  | 1x                            | $V_{RL} * (0/1024) * 1$    | 0        |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (0/1024) * 1$    | 0        |
| MCP48CVD0X (8-bit)  | 1111 1111                   | 5.0V           | $5.0V/256$  | 19,531.3 | 1x                            | $V_{RL} * (255/256) * 1$   | 4.980469 |
|                     |                             | 2.5V           | $2.5V/256$  | 9,765.6  | 1x                            | $V_{RL} * (255/256) * 1$   | 2.490234 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (255/256) * 2$   | 4.980469 |
|                     | 0111 1111                   | 5.0V           | $5.0V/256$  | 19,531.3 | 1x                            | $V_{RL} * (127/256) * 1$   | 2.480469 |
|                     |                             | 2.5V           | $2.5V/256$  | 9,765.6  | 1x                            | $V_{RL} * (127/256) * 1$   | 1.240234 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (127/256) * 2$   | 2.480469 |
|                     | 0011 1111                   | 5.0V           | $5.0V/256$  | 19,531.3 | 1x                            | $V_{RL} * (63/256) * 1$    | 1.230469 |
|                     |                             | 2.5V           | $2.5V/256$  | 9,765.6  | 1x                            | $V_{RL} * (63/256) * 1$    | 0.615234 |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (63/256) * 2$    | 1.230469 |
|                     | 0000 0000                   | 5.0V           | $5.0V/256$  | 19,531.3 | 1x                            | $V_{RL} * (0/256) * 1$     | 0        |
|                     |                             | 2.5V           | $2.5V/256$  | 9,765.6  | 1x                            | $V_{RL} * (0/256) * 1$     | 0        |
|                     |                             |                |             |          | 2x <sup>(2)</sup>             | $V_{RL} * (0/256) * 2$     | 0        |

- Note 1:**  $V_{RL}$  is the resistor ladder's reference voltage. It is independent of the VREF1:VREF0 selection.
- Note 2:** Gain selection of 2X ( $GX = 1$ ) requires the voltage reference source to come from the  $V_{REF}$  pin ( $VREF1:VREF0 = 10$  or  $11$ ) and requires  $V_{REF}$  pin voltage (or  $V_{RL}$ )  $\leq V_{DD}/2$  or from the internal band gap ( $VREF1:VREF0 = 01$ ).
- Note 3:** These theoretical calculations do not take into account the Offset, Gain and Nonlinearity errors.

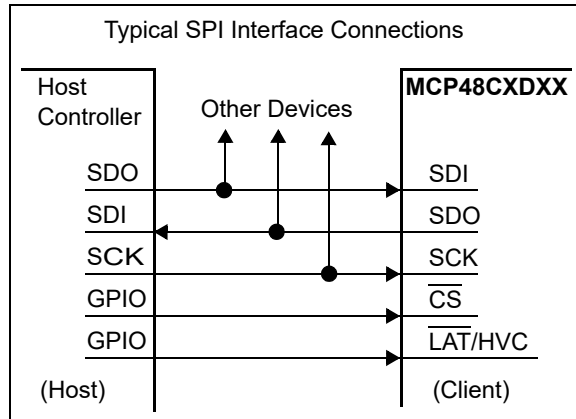
# MCP48CXX1/2

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NOTES:

## 6.0 SPI SERIAL INTERFACE MODULE

The MCP48CXDX1/2's SPI serial interface module is a four-wire interface. The devices operate only as Clients (do not generate the Host clock). The figure below shows a typical SPI interface connection.



**FIGURE 6-1:** Typical SPI Interface.

The frame content (commands) for the MCP48CXDX1/2 are defined in [Section 7.0 “Device Commands”](#).

### 6.1 Overview

This section discusses some of the specific characteristics of the MCP48CXDX1/2's SPI serial interface module. This is to assist in the development of your application.

The following sections discuss some of these device-specific characteristics:

- [Communication Data Rates](#)
- [POR/BOR](#)
- [Interface Pins \(CS, SCK, SDI, SDO and LAT/HVC\)](#)
- [Device Memory Address](#)
- [SPI Modes](#)

The MCP48CXDX1/2 devices support the SPI serial protocol. This SPI operates in Client mode (does not generate the serial clock).

The SPI interface uses four pins. These are:

- $\overline{\text{CS}}$  – Chip Select
- SCK – Serial Clock
- SDI – Serial Data In
- SDO – Serial Data Out

A fifth pin is used if a write is being done in the MTP memory. This pin is HVC – High-Voltage command (customer manufacturing only, multiplexed with LAT0 functionality).

The HVC pin is used to program the MTP memory. This is intended to be used only during the customer's factory production flow. On volatile devices, the HVC pin is high-voltage tolerant. To enter a high-voltage command, the HVC pin must be greater than the  $V_{\text{IH}}$  voltage.

Typical SPI Interfaces are shown in [Figure 6-1](#). In the SPI interface, the Host's Output pin is connected to the Client's Input pin and the Host's Input pin is connected to the Client's Output pin.

The MCP48CXDX1/2 SPI module supports two (of the four) standard SPI modes. These are Mode 0, 0 and 1, 1. The SPI mode is determined by the state of the SCK pin ( $V_{\text{IH}}$  or  $V_{\text{IL}}$ ) when the  $\overline{\text{CS}}$  pin transitions from inactive ( $V_{\text{IH}}$ ) to active ( $V_{\text{IL}}$ ).

### 6.2 Communication Data Rates

The MCP48CXDX1/2 supports clock rates (bit rates) of up to 25 MHz for read and 50 MHz for write commands.

For most applications, the write time will be considered more important because that is how the device operation is controlled.

### 6.3 POR/BOR

On a POR/BOR event, the SPI serial interface module state machine is reset, which includes forcing the device's memory address pointer to 00h.

### 6.4 Interface Pins ( $\overline{\text{CS}}$ , SCK, SDI, SDO and LAT/HVC)

The operation of the four interface pins and the High-Voltage command (HVC) pin is discussed in this section. The serial interface works on 24-bit boundaries. The  $\overline{\text{CS}}$  pin frames the SPI commands.

#### 6.4.1 SERIAL DATA IN (SDI)

The Serial Data In (SDI) signal is the data signal in the device. The value on this pin is latched on the rising edge of the SCK signal.

#### 6.4.2 SERIAL DATA OUT (SDO)

The Serial Data Out (SDO) signal is the data signal out of the device. The value on this pin is driven on the falling edge of the SCK signal.

Once the  $\overline{\text{CS}}$  pin is forced to the active level ( $V_{\text{IL}}$ ), the SDO pin is driven. The state of the SDO pin is determined by the serial bit's position in the command, the selected command and if there is a state of command error (CMDERR).

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## 6.4.3 SERIAL CLOCK (SCK) (SPI FREQUENCY OF OPERATION)

The SPI interface is specified to operate up to 50 MHz for write commands and 25 MHz for read commands. The actual clock rate depends on the configuration of the system and the serial command used. The table below shows the SCK frequency for different configurations.

**TABLE 6-1: SCK FREQUENCY**

| Memory Type Access |          | Command |                       |
|--------------------|----------|---------|-----------------------|
|                    |          | Read    | Write                 |
| Nonvolatile Memory | SDI, SDO | 25 MHz  | 50 MHz <sup>(1)</sup> |
| Volatile Memory    | SDI, SDO | 25 MHz  | 50 MHz <sup>(1)</sup> |

**Note 1:** After issuing a write command to the NV locations, the internal write cycle must be completed before the next SPI command addressing the NV locations is received ( $t_{wc}$ ).

## 6.4.4 THE $\overline{CS}$ SIGNAL

The  $\overline{CS}$  signal is used to select the device and frame a command sequence. To start a command, or sequence of commands, the  $\overline{CS}$  signal must transition from the inactive state ( $V_{IH}$ ) to an active state ( $V_{IL}$ ).

After the  $\overline{CS}$  signal goes active, the SDO pin is driven and the clock bit counter is reset.

**Note:** There is a required delay after the  $\overline{CS}$  pin goes active to the 1<sup>st</sup> edge of the SCK pin.

If an error condition occurs for an SPI command, the Command byte's Command Error (CMDERR) bit (on the SDO pin) will be driven low ( $V_{IL}$ ). To exit the error condition, the user must take the  $\overline{CS}$  pin to the  $V_{IH}$  level.

When the  $\overline{CS}$  pin returns to the inactive state ( $V_{IH}$ ), the SPI module resets (including the address pointer). While the  $\overline{CS}$  pin is in the inactive state ( $V_{IH}$ ), the serial interface is ignored. This allows the Host Controller to interface to other SPI devices using the same SDI, SDO and SCK signals.

## 6.4.5 HVC SIGNAL

The high-voltage requirement of the HVC pin for programming MTP registers ensures that a device in normal operation does not corrupt the values.

## 6.5 Device Memory Address

The memory address is the 5-bit value that specifies the location in the device's memory that the specified command will operate on.

On a POR/BOR event, the device's memory address pointer is forced to 00h.

## 6.6 SPI Modes

The SPI module supports two (of the four) standard SPI modes. These are Mode 0,0 and 1,1. The MCP48CXDX1/2's SPI mode is automatically determined based on the Host's configured mode.

## 6.6.1 OPERATION IN SPI MODE 0,0

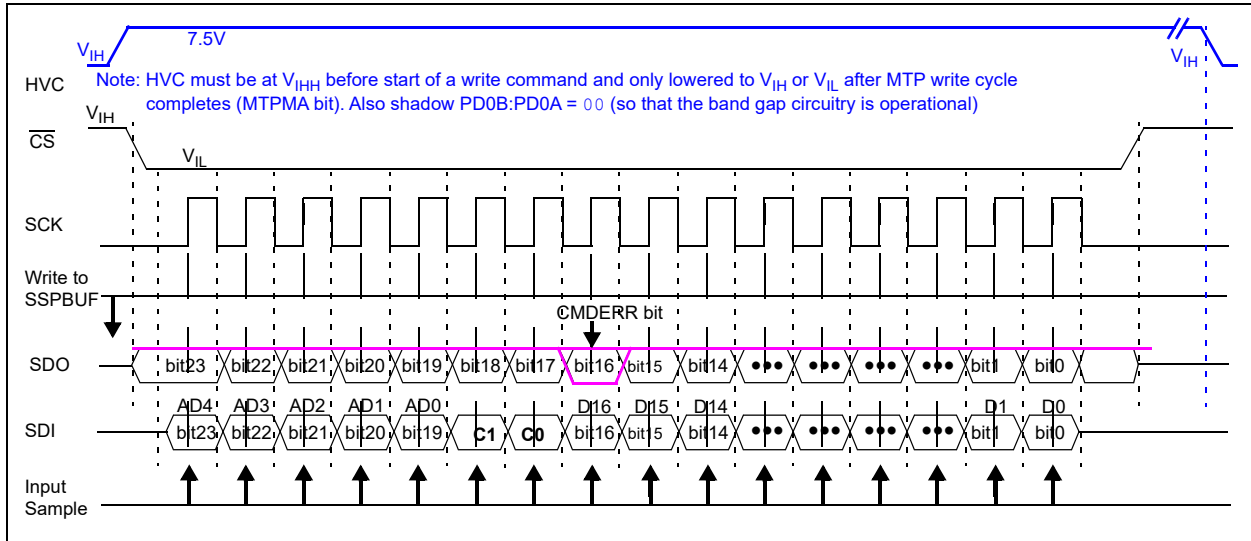
In SPI Mode 0,0:

- SCK idle state = Low ( $V_{IL}$ )
- Data are clocked in on the SDI pin on the rising edge of SCK
- Data are clocked out on the SDO pin on the falling edge of SCK

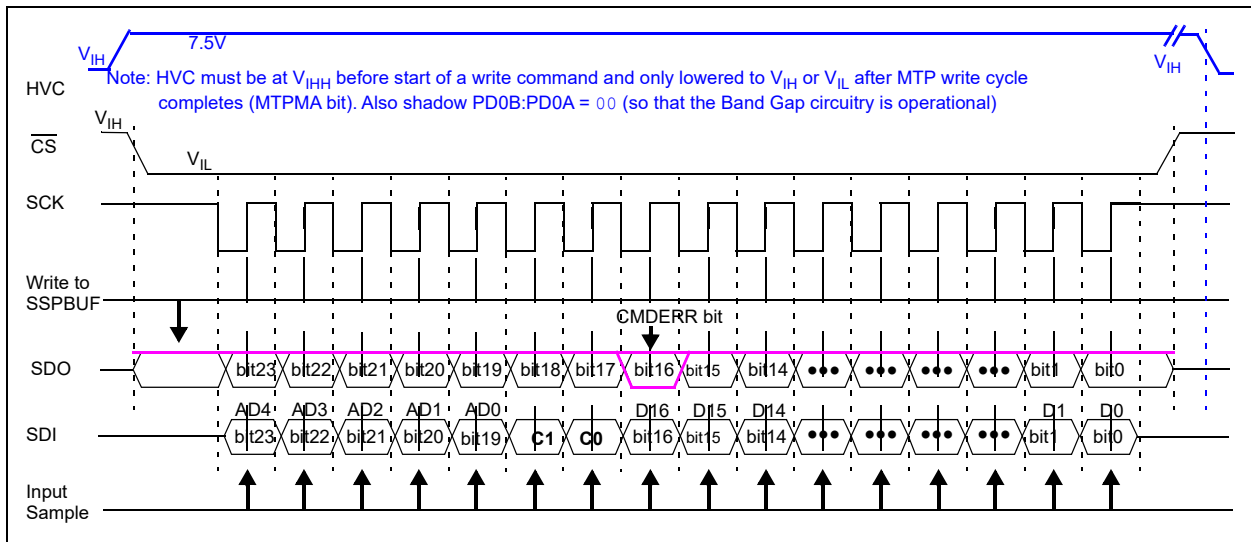
## 6.6.2 OPERATION IN SPI MODE 1,1

In SPI Mode 1,1:

- SCK idle state = High ( $V_{IH}$ )
- Data are clocked in on the SDI pin on the rising edge of SCK
- Data are clocked out on the SDO pin on the falling edge of SCK



**FIGURE 6-2:** 24-Bit Commands (Write, Read) – SPI Waveform (Mode 0,0).



**FIGURE 6-3:** 24-Bit Commands (Write, Read) – SPI Waveform (Mode 1,1).

# MCP48CXX1/2

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NOTES:

## 7.0 DEVICE COMMANDS

The MCP48CXDX1/2's SPI command format supports 32 memory address locations and two commands. The command may have two modes:

- Normal Serial Commands
- MTP Programming (HV) Serial Commands

Normal Serial commands are those where the HVC pin is driven to either  $V_{IH}$  or  $V_{IL}$ . With High-Voltage Serial commands, the HVC pin is driven to 7.5V. These commands are shown in Table 7-1.

Table 7-2 shows an overview of all the SPI commands and their interaction with other device features.

The 24-bit commands ([Read Command](#) and [Write Command](#)) contain a Command byte and a Data Word. The Command byte contains one reserved bit, see Figure 7-1.

TABLE 7-1: COMMAND BITS OVERVIEW

| Bit States<br>C1:C0 | Command    | # of Bits | Normal or HV               |
|---------------------|------------|-----------|----------------------------|
| 11                  | Read Data  | 24-Bits   | Normal only <sup>(1)</sup> |
| 00                  | Write Data | 24-Bits   | Both                       |
| 01                  | Reserved   | —         | —                          |
| 10                  | Reserved   | —         | —                          |

**Note 1:** Reading from the NV memory locations will return the shadow RAM value of the NV memory, not the NV memory contents. Once a write cycle starts, no other commands accessing NV memory locations are allowed.

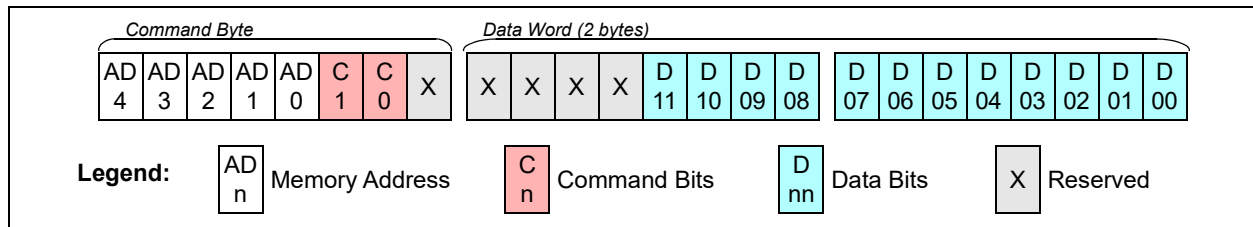


FIGURE 7-1: 24-bit SPI Command Format.

TABLE 7-2: SPI COMMANDS – NUMBER OF CLOCKS

| Command                         |      |    |              |                     | # of Bit Clocks <sup>(2)</sup> | Data Update Rate<br>(8-bit/10-bit/12-bit)<br>(Data Words/Second) |         |           | Comments          |
|---------------------------------|------|----|--------------|---------------------|--------------------------------|------------------------------------------------------------------|---------|-----------|-------------------|
| Operation                       | Code |    | HV<br>(4, 7) | Mode <sup>(1)</sup> |                                | 1 MHz                                                            | 10 MHz  | 50 MHz    |                   |
|                                 | C1   | C0 |              |                     |                                |                                                                  |         |           |                   |
| Write Command <sup>(3, 6)</sup> | 0    | 0  | Y            | Single              | 24                             | 41,666                                                           | 416,666 | 2,083,333 | —                 |
|                                 | 0    | 0  | N            | Continuous          | 24 * n                         | 41,666                                                           | 416,666 | 2,083,333 | For 10 data words |
| Read Command <sup>(5)</sup>     | 1    | 1  | N            | Random              | 24                             | 41,666                                                           | 416,666 | 2,083,333 | —                 |
|                                 | 1    | 1  | N            | Continuous          | 24 * n                         | 41,666                                                           | 416,666 | 2,083,333 | For 10 data words |

**Note 1:** Nonvolatile registers can only use the Single mode.

**2:** “n” indicates the number of times the command operation is to be repeated.

**3:** The registers are updated after the 24<sup>th</sup> clock bit or after the  $\overline{CS}$  rising, depending on mode.

**4:** If the state of the HVC pin is  $V_{IHH}$ , the command is ignored, but a Command Error condition (CMDERR) will NOT be generated.

**5:** This command is useful to determine when an MTP programming cycle is complete.

**6:** This command can be either normal voltage or high voltage.

**7:** The MTP write cycle starts after the  $\overline{CS}$  rising edge. A high-voltage command requires the HVC pin to be at  $V_{IHH}$  for the entire command, until the completion of the MTP write cycle.

# MCP48CXDX1/2

## 7.1 Command Byte

The Command byte has three fields, the Address (5 bits), the Command (2 bits) and one Reserved bit (see [Figure 7-1](#)).

The device memory is accessed when the Host sends a proper Command byte to select the desired operation. The memory location getting accessed is contained in the Command byte's AD4:AD0 bits. The action desired is contained in the Command byte's C1:C0 bits, see [Figure 7-5](#). C1:C0 determines if the desired memory location will be read or written.

As the Command byte is loaded into the device (on the SDI pin), the device's SDO pin drives. The SDO pin will output high bits for the first seven bits of that command. On the 8<sup>th</sup> bit, the SDO pin will output the CMDERR bit state.

## 7.2 Data Bytes

The read and write commands use Data bytes, see [Figure 7-1](#). The D16 bit is currently unused and corresponds to the position on the SDO data of the CMDERR bit.

## 7.3 Error Condition

The CMDERR bit indicates if the five address bits received (AD4:AD0) and the two command bits received (C1:C0) are a valid combination. The CMDERR bit is high if the combination is valid and low if the combination is invalid.

The command error bit will also be low if a write to a nonvolatile address is specified and another SPI command occurs before the  $\overline{\text{CS}}$  pin is driven inactive ( $V_{IH}$ ).

SPI commands that do not have a multiple of 24 clocks are ignored.

Once an error condition has occurred, any following commands are ignored. All following SDO bits will be low until the CMDERR condition is cleared by forcing the  $\overline{\text{CS}}$  pin to the Inactive state ( $V_{IH}$ ) or doing a POR.

### 7.3.1 ABORTING A TRANSMISSION

All SPI transmissions must have the correct number of SCK pulses to be executed. The command is not executed until the complete number of clocks are received. Some commands also require the  $\overline{\text{CS}}$  pin to be forced inactive ( $V_{IH}$ ). If the  $\overline{\text{CS}}$  pin is forced to the Inactive state ( $V_{IH}$ ), the serial interface is reset. Partial commands are not executed.

SPI is more susceptible to noise than other bus protocols. The most likely case is that this noise corrupts the value of the data being clocked into the MCP48CXDX1/2 or the SCK pin is injected with extra clock pulses. This may cause data to be corrupted in the device or a command error to occur because the address and command bits were not a valid combination. The extra SCK pulse will also cause the SPI data (SDI) and clock (SCK) to be out of sync. Forcing the  $\overline{\text{CS}}$  pin to the Inactive state ( $V_{IH}$ ) resets the serial interface. The SPI interface will ignore activity on the SDI and SCK pins until the  $\overline{\text{CS}}$  pin transition to the active state is detected ( $V_{IH}$  to  $V_{IL}$ ).

**Note 1:** When the MCP48CXDX1/2 does not receive data, it is recommended that the  $\overline{\text{CS}}$  pin be forced to the inactive level ( $V_{IL}$ ).

## 7.4 Continuous Commands

The device supports the ability to execute commands continuously. While the  $\overline{\text{CS}}$  pin is in Active state ( $V_{IL}$ ), any sequence of valid commands may be received.

The following example is a valid sequence of events:

1.  $\overline{\text{CS}}$  pin driven active ( $V_{IL}$ )
2. Read command
3. Write command (volatile memory)
4. Write command (nonvolatile memory)
5.  $\overline{\text{CS}}$  pin driven inactive ( $V_{IH}$ )

**Note 1:** While the  $\overline{\text{CS}}$  pin is active, only one type of command must be issued. When changing commands, it is recommended that the  $\overline{\text{CS}}$  pin be made inactive, then forced back to the active state.

- 2: Long command strings must be broken down into shorter command strings. This reduces the probability of noise on the SCK pin corrupting the desired SPI command string.

## 7.5 Write Command

The write command is a 24-bit command. The write command can be issued to both the volatile and non-volatile memory locations. The format of the command is shown in [Figure 7-2](#).

A write command to a volatile memory location changes that location after a properly formatted write command (24-clock) is received.

A write command to a nonvolatile memory location will only start a write cycle after a properly formatted write command (24-clock) is received and the  $\overline{\text{CS}}$  pin transitions to the Inactive state ( $V_{IH}$ ).

**Note:** Writes to volatile memory locations depend on the state of the WiperLock™ Technology bits.

## 7.5.1 SINGLE WRITE TO VOLATILE MEMORY

The write operation requires that the  $\overline{CS}$  pin be in the Active state ( $V_{IL}$ ). Typically, the  $\overline{CS}$  pin is in the Inactive state ( $V_{IH}$ ) and is driven to the Active state ( $V_{IL}$ ). The 24-bit write command (Command byte and Data bytes) is, then, clocked in on the SCK and SDI pins. Once all 24 bits are received, the specified volatile address is updated. A write will not occur if the write command is not exactly 24 clock pulses. This prevents system issues from corrupting the nonvolatile memory locations.

Figure 7-2 shows the waveform for a single write.

## 7.5.2 SINGLE WRITE TO NONVOLATILE MEMORY

The sequence to write to a single nonvolatile memory location is the same as a single write to volatile memory with the exception that before the command, the HVC pin must be driven to  $V_{IH}$ . After the command, the  $\overline{CS}$  pin is driven inactive ( $V_{IH}$ ), which then starts the MTP write cycle ( $t_{wc}$ ). The HVC pin must remain at the  $V_{IH}$  level until the completion of the MTP write cycle.

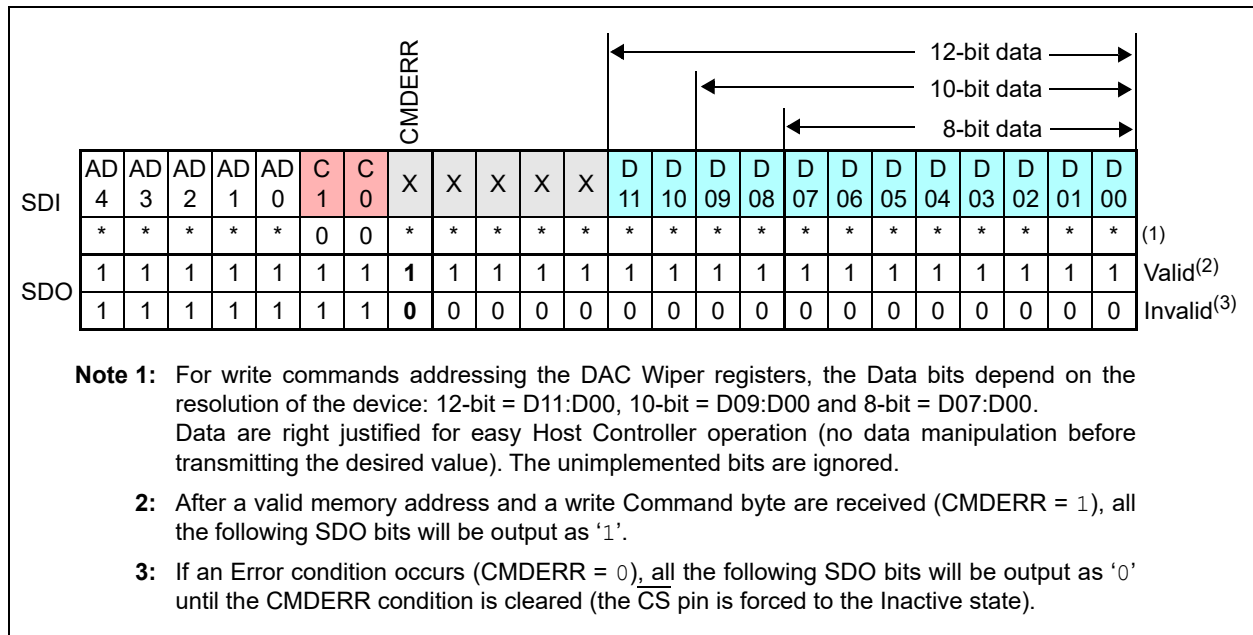
A write cycle will not start if the write command is not exactly 24 clock pulses. This protects against system issues from corrupting the nonvolatile memory locations.

After the  $\overline{CS}$  pin is driven inactive ( $V_{IH}$ ), the serial interface may immediately be re-enabled by driving the  $\overline{CS}$  pin to the Active state ( $V_{IL}$ ).

During an MTP write cycle, only serial commands to volatile memory are accepted. All other serial commands are ignored until the MTP write cycle ( $t_{wc}$ ) is completed. The MTPMA bit in the Status register indicates the status of an MTP write cycle.

Once a write command to a nonvolatile memory location is received, no other SPI commands can be received before the  $\overline{CS}$  pin transitions to the Inactive state ( $V_{IH}$ ) or a CMDERR on the current SPI command occurs.

The write to a nonvolatile memory command has the same format as the write to a volatile memory command (see the figure below).



**FIGURE 7-2:** Write Single Memory Location Command – SDI and SDO States.

# MCP48CXDX1/2

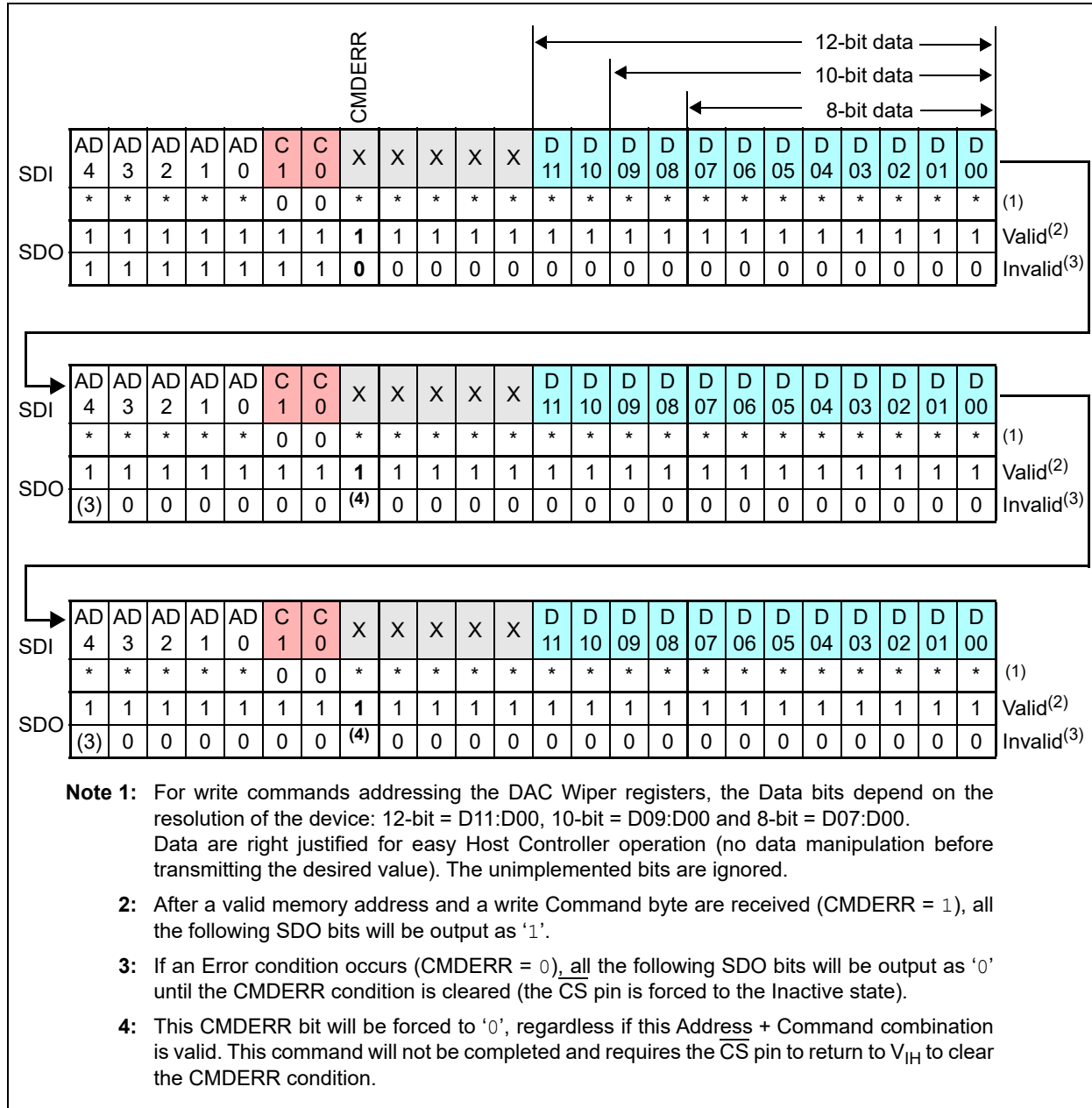
## 7.5.3 CONTINUOUS WRITES TO VOLATILE MEMORY

Continuous writes are possible only when writing to the volatile memory registers.

The figure below shows the sequence for three continuous writes. The writes do not need to be to the same volatile memory address.

## 7.5.4 CONTINUOUS WRITES TO NONVOLATILE MEMORY

Continuous writes to nonvolatile memory are not allowed, and attempts to do so will result in a CMDERR condition.



**FIGURE 7-3:** Continuous Write Sequence (Volatile Memory Only).

## 7.6 Read Command

The read command is a 24-bit command. The read command can be issued to both the volatile and nonvolatile memory locations. The format of the command is shown in the figure below.

The first seven bits of the read command determine the address and the command. The 8<sup>th</sup> clock will output the CMDERR bit on the SDO pin. For the remaining 16 clocks, the device will transmit the data bits of the specified address (AD4:AD0).

The figure below shows the SDI and SDO information for a read command.

During an MTP write cycle, the read command can only be issued to the volatile memory locations. By reading the Status register, the Host Controller can determine when the write cycle is completed (via the state of the MTPMA bit).

### 7.6.1 SINGLE READ

The read operation requires that the  $\overline{CS}$  pin be in the Active state ( $V_{IL}$ ). Typically, the  $\overline{CS}$  pin will be in the Inactive state ( $V_{IH}$ ) and is driven to the Active state ( $V_{IL}$ ). The 24-bit read command (Command byte and Data Word) is then clocked in on the SCK and SDI pins. The SDO pin starts driving high ( $V_{IH}$ ) when the  $\overline{CS}$  goes active and starts driving data on the 8<sup>th</sup> bit (CMDERR bit); the addressed data come out on the 9<sup>th</sup> through 24<sup>th</sup> clocks.



**FIGURE 7-4:** Read Single Memory Location Command – SDI and SDO States.

# MCP48CXDX1/2

## 7.6.2 CONTINUOUS READS

Continuous reads allow the device's memory to be read quickly. Continuous reads are possible to all memory locations. Read commands may only access volatile memory locations during an MTP write cycle.

The figure below shows the sequence for three continuous reads. The reads do not need to be to the same memory address.



**FIGURE 7-5:** Continuous Read Sequence.

## 8.0 TYPICAL APPLICATIONS

The MCP48CXDX1/2 devices are general purpose, single/dual channel voltage output DACs for various applications where a precision operation with low power is needed.

Applications generally suited for the devices are:

- Set Point or Offset Trimming
- Sensor Calibration
- Portable Instrumentation (Battery-Powered)
- Motor Control

## 8.1 Design Considerations

In the design of a system with the MCP48CXDX1/2 devices, the following considerations must be taken into account:

- **Power Supply Considerations**
- **Layout Considerations**

### 8.1.1 POWER SUPPLY CONSIDERATIONS

The power source supplying these devices must be as clean as possible. If the application circuit has separate digital and analog power supplies,  $V_{DD}$  and  $V_{SS}$  may reside on the analog plane.

The power supply to the device is also used for the DAC voltage reference internally if the internal  $V_{DD}$  is selected as the resistor ladder's reference voltage.

The typical application requires a bypass capacitor in order to filter high-frequency noise, which can be induced onto the power supply's traces. The bypass capacitor helps to minimize the effect of these noise sources on signal integrity.

Any noise induced on the  $V_{DD}$  line can affect the DAC performance. Typical applications require a bypass capacitor in order to filter out high-frequency noise on the  $V_{DD}$  line. The noise can be induced onto the power supply's traces or as a result of changes on the DAC output. The bypass capacitor helps to minimize the effect of these noise sources on signal integrity. The figure below shows an example of using two bypass capacitors (a 10  $\mu\text{F}$  tantalum capacitor and a 0.1  $\mu\text{F}$  ceramic capacitor) in parallel on the  $V_{DD}$  line. These capacitors must be placed as close to the  $V_{DD}$  pin as possible (within 4 mm). If the application circuit has separate digital and analog power supplies, the  $V_{DD}$  and  $V_{SS}$  pins of the device must reside on the analog plane.



**FIGURE 8-1:** Example Circuit.

# MCP48CXDX1/2

## 8.1.2 LAYOUT CONSIDERATIONS

Several layout considerations may be applicable to your application. These may include:

- [Noise](#)
- [PCB Area Requirements](#)

### 8.1.2.1 Noise

Inductively-coupled AC transients and digital switching noise can degrade the input and output signal integrity, potentially masking the MCP48CXDX1/2's performance. Careful board layout minimizes these effects and increases the Signal-to-Noise Ratio (SNR). Multi-layer boards utilizing a low-inductance ground plane, isolated inputs, isolated outputs and proper decoupling are critical to achieving the performance that the silicon is capable of providing. Particularly harsh environments may require shielding of critical signals.

Separate digital and analog ground planes are recommended. In this case, the  $V_{SS}$  pin and the ground pins of the  $V_{DD}$  capacitors must be terminated to the analog ground plane.

**Note:** Breadboards and wire-wrapped boards are not recommended.

### 8.1.2.2 PCB Area Requirements

In some applications, PCB area is a criteria for device selection. The table below shows the typical package dimensions and area for the different package options.

**TABLE 8-1: PACKAGE FOOTPRINT<sup>(1)</sup>**

| Package |      |      | Package Footprint |       |                         |
|---------|------|------|-------------------|-------|-------------------------|
| Pins    | Type | Code | Dimensions (mm)   |       | Area (mm <sup>2</sup> ) |
|         |      |      | Length            | Width |                         |
| 10      | MSOP | UN   | 3.00              | 4.90  | 14.70                   |
| 10      | DFN  | MF   | 3.00              | 3.00  | 9.00                    |
| 16      | QFN  | MG   | 3.00              | 3.00  | 9.00                    |

**Note 1:** Does not include recommended land pattern dimensions. Dimensions are typical values.

## 8.2 Application Examples

The MCP48CXDX1/2 devices are rail-to-rail output DACs designed to operate with a  $V_{DD}$  range of 2.7V to 5.5V. The internal output amplifier is robust enough to drive common, small-signal loads directly, thus eliminating the cost and size of the external buffers for most applications. The user can select the gain of 1 or 2 of the output op amp by setting the Configuration register bits. The internal  $V_{DD}$  or an external reference can be used. There are various user options and easy-to-use features that make the devices suitable for various modern DAC applications.

Application examples include:

- [Decreasing Output Step Size](#)
- [Building a "Window" DAC](#)
- [Bipolar Operation](#)
- [Selectable Gain and Offset Bipolar Voltage Output](#)
- [Designing a Double-Precision DAC](#)
- [Building Programmable Current Source](#)
- [Serial Interface Communication Times](#)
- [Development Support](#)
- [Power Supply Considerations](#)
- [Layout Considerations](#)

### 8.2.1 DC SET POINT OR CALIBRATION

A common application for the devices is a digitally-controlled set point and/or calibration of variable parameters, such as sensor offset or slope. For example, the MCP48CVD2X provides 4096 output steps. If voltage reference is 4.096V (where  $G_X = 0$ ), the  $LSb$  size is 1 mV. If a smaller output step size is desired, a lower external voltage reference is needed.

## 8.2.1.1 Decreasing Output Step Size

If the application calibrates the bias voltage of a diode or transistor, a bias voltage range of 0.8V may be desired with about 200  $\mu$ V resolution per step. Two common methods to achieve a small step size are to use a lower  $V_{REF}$  pin voltage or a voltage divider on the DAC's output.

Using an external voltage reference ( $V_{REF}$ ) is an option if the external reference is available with the desired output voltage range. However, when using a low-voltage reference voltage, occasionally the noise floor causes an SNR error that is intolerable. Using a voltage divider method is another option and provides some advantages when the external voltage reference needs to be very low, or when the desired output voltage is not available. In this case, a larger value reference voltage is used, while two resistors scale the output range down to the precise desired level.

The figure below illustrates this concept. A bypass capacitor on the output of the voltage divider plays a critical function in attenuating the output noise of the DAC and the induced noise from the environment.



**FIGURE 8-2:** Example Circuit Of Set Point or Threshold Calibration.

### EQUATION 8-1: $V_{OUT}$ AND $V_{TRIP}$ CALCULATIONS

$$V_{OUT} = V_{REF} \cdot G \cdot \frac{DAC \text{ Register Value}}{2^N}$$

$$V_{trip} = V_{OUT} \left( \frac{R_2}{R_1 + R_2} \right)$$

## 8.2.1.2 Building a “Window” DAC

When calibrating a set point or threshold of a sensor, typically only a small portion of the DAC output range is utilized. If the LSb size is adequate enough to meet the application's accuracy needs, the unused range is sacrificed without consequences. If greater accuracy is needed, the output range will need to be reduced to increase the resolution around the desired threshold.

If the threshold is not near  $V_{REF}$ ,  $2 \cdot V_{REF}$  or  $V_{SS}$ , then creating a “window” around the threshold has several advantages. One simple method to create this “window” is to use a voltage divider network with a pull-up and pull-down resistor. The figure below and Figure 8-5 illustrate this concept.



**FIGURE 8-3:** Single Supply “Window” DAC.

### EQUATION 8-2: $V_{OUT}$ AND $V_{TRIP}$ CALCULATIONS

$$V_{OUT} = V_{REF} \cdot G \cdot \frac{DAC \text{ Register Value}}{2^N}$$

$$V_{TRIP} = \frac{V_{OUT} R_{23} + V_{23} R_1}{R_1 + R_{23}}$$

$$\text{Thevenin Equivalent} \left\{ \begin{array}{l} R_{23} = \frac{R_2 R_3}{R_2 + R_3} \\ V_{23} = \frac{(V_{CC+} R_2) + (V_{CC-} R_3)}{R_2 + R_3} \end{array} \right.$$



# MCP48CXDX1/2

## 8.3 Bipolar Operation

Bipolar operation is achievable by utilizing an external operational amplifier. This configuration is desirable due to the wide variety and availability of op amps. This allows a general purpose DAC, with its cost and availability advantages, to meet almost any desired output voltage range, power and noise performance.

The figure below illustrates a simple bipolar voltage source configuration.  $R_1$  and  $R_2$  allow the gain to be selected, while  $R_3$  and  $R_4$  shift the DAC's output to a selected offset. Note that  $R_4$  can be tied to  $V_{DD}$  instead of  $V_{SS}$  if a higher offset is desired.



**FIGURE 8-4:** Digitally-Controlled Bipolar Voltage Source Example Circuit.

### EQUATION 8-3: $V_{OUT}$ , $V_{OA+}$ AND $V_O$ CALCULATIONS

$$V_{OUT} = V_{REF} \cdot G \cdot \frac{\text{DAC Register Value}}{2^N}$$

$$V_{OA+} = \frac{V_{OUT} \cdot R_4}{R_3 + R_4}$$

$$V_O = V_{OA+} \cdot \left(1 + \frac{R_2}{R_1}\right) - V_{DD} \cdot \left(\frac{R_2}{R_1}\right)$$

## 8.4 Selectable Gain and Offset Bipolar Voltage Output

In some applications, precision digital control of the output range is desirable. Figure 8-5 illustrates how to use the DAC devices to achieve this in a bipolar or single supply application.

This circuit is typically used for linearizing a sensor whose slope and offset vary.

The equation to design a bipolar “window” DAC is utilized if  $R_3$ ,  $R_4$  and  $R_5$  are populated.

### 8.4.1 BIPOLAR DAC EXAMPLE

An output step size of 1 mV, with an output range of  $\pm 2.05V$ , is desired for a particular application.

**Step 1:** Calculate the range:  $+2.05V - (-2.05V) = 4.1V$

**Step 2:** Calculate the resolution needed:

$$4.1V / 1 \text{ mV} = 4100$$

$2^{12} = 4096$ ; therefore, 12-bit resolution is desired.

**Step 3:** The amplifier gain ( $R_2/R_1$ ), multiplied by full-scale  $V_{OUT}$  (4.096V), must be equal to the desired minimum output to achieve bipolar operation. Any gain can be realized by choosing resistor values ( $R_1 + R_2$ ); therefore, the  $V_{REF}$  value must be selected first. If a  $V_{REF}$  of 4.096V is used, solve the amplifier's gain by setting the DAC to 0, knowing that the output needs to be -2.05V.

The equation can be simplified to:

### EQUATION 8-4:

$$\frac{-R_2}{R_1} = \frac{-2.05}{4.096V} \quad \frac{R_2}{R_1} = \frac{1}{2}$$

If  $R_1 = 20 \text{ k}\Omega$  and  $R_2 = 10 \text{ k}\Omega$ , the gain will be 0.5.

**Step 4:** Next, solve for  $R_3$  and  $R_4$  by setting the DAC to 4096, knowing that the output needs to be +2.05V.

### EQUATION 8-5:

$$\frac{R_4}{(R_3 + R_4)} = \frac{2.05V + (0.5 \cdot 4.096V)}{1.5 \cdot 4.096V} = \frac{2}{3}$$

If  $R_4 = 20 \text{ k}\Omega$ , then  $R_3 = 10 \text{ k}\Omega$



**FIGURE 8-5:** Bipolar Voltage Source with Selectable Gain and Offset.

**EQUATION 8-6:  $V_{OUT}$ ,  $V_{OA+}$  AND  $V_O$  CALCULATIONS**

$$V_{OUT} = V_{REF} \cdot G \cdot \frac{\text{DAC Register Value}}{2^N}$$

$$V_{OA+} = \frac{V_{OUT} \cdot R_4 + V_{CC-} \cdot R_5}{R_3 + R_4}$$

$$V_O = \underbrace{V_{OA+} \cdot \left(1 + \frac{R_2}{R_1}\right)}_{\text{Offset Adjust}} - \underbrace{V_{IN} \cdot \left(\frac{R_2}{R_1}\right)}_{\text{Gain Adjust}}$$

**EQUATION 8-7: BIPOLAR “WINDOW” DAC USING  $R_4$  AND  $R_5$**

Thevenin Equivalent

$$V_{45} = \frac{V_{CC+} R_4 + V_{CC-} R_5}{R_4 + R_5}$$

$$V_{IN+} = \frac{V_{OUT} R_{45} + V_{45} R_3}{R_3 + R_{45}}$$

$$R_{45} = \frac{R_4 R_5}{R_4 + R_5}$$

$$V_O = \underbrace{V_{IN+} \left(1 + \frac{R_2}{R_1}\right)}_{\text{Offset Adjust}} - \underbrace{V_A \left(\frac{R_2}{R_1}\right)}_{\text{Gain Adjust}}$$

## 8.5 Designing a Double-Precision DAC

The figure below shows an example design of a single supply voltage output capable of up to a 24-bit resolution. This requires two 12-bit DACs. This design is simply a voltage divider with a buffered output.

As an example, if a similar application to the one developed in [Section 8.4.1 “Bipolar DAC Example”](#) required a resolution of 1  $\mu V$ , instead of 1 mV, and a range of 0V to 4.1V, then a 12-bit resolution would not be adequate.

**Step 1:** Calculate the resolution needed:

$$4.1V/1 \mu V = 4.1 \times 10^6$$

$2^{22} = 4.2 \times 10^6$ ; therefore, a 22-bit resolution is desired. DNL =  $\pm 1.0$  LSB; therefore, this design can be attempted with the 12-bit DAC.

**Step 2:** DAC1's  $V_{OUT1}$  has a resolution of 1 mV; therefore, its output only needs to be “pulled” 1/1000 to meet the 1  $\mu V$  target. Dividing  $V_{OUT0}$  by 1000 would allow the application to compensate for DAC1's DNL error.

**Step 3:** If  $R_2$  is 100 $\Omega$ , then  $R_1$  needs to be 100 k $\Omega$ .

**Step 4:** The resulting transfer function is shown in [Equation 8-8](#).



**FIGURE 8-6:** Simple Double Precision DAC Using MCP48CVDX2.

**EQUATION 8-8:  $V_{OUT}$  CALCULATION**

$$V_{OUT} = \frac{V_{OUT0} \cdot R_2 + V_{OUT1} \cdot R_1}{R_1 + R_2}$$

Where:

$$V_{OUT0} = (V_{REF} \cdot G \cdot \text{DAC0 register value})/4096$$

$$V_{OUT1} = (V_{REF} \cdot G \cdot \text{DAC1 register value})/4096$$

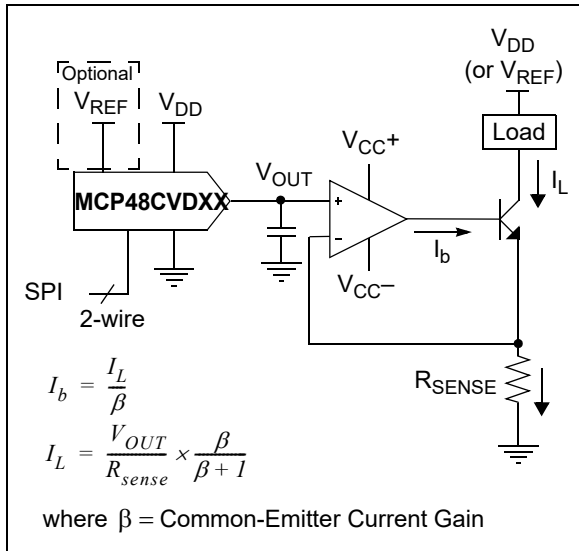
GX = Selected Op Amp Gain

# MCP48CXDX1/2

## 8.6 Building Programmable Current Source

The figure below shows an example of building a programmable current source using a voltage follower. The current sensor resistor is used to convert the DAC voltage output into a digitally-selectable current source.

The smaller  $R_{SENSE}$  is, the less power is dissipated across it. However, this also reduces the resolution that the current can be controlled at.



**FIGURE 8-7:** Digitally-Controlled Current Source.

## 8.7 Serial Interface Communication Times

Table 7-2 shows the time/frequency of the supported operations of the SPI serial interface for the different serial interface operational frequencies. This, along with the  $V_{OUT}$  output performance (such as slew rate), would be used to determine your application's volatile DAC register update rate.

## 9.0 DEVELOPMENT SUPPORT

Development support can be classified into two groups:

- [Development Tools](#)
- [Technical Documentation](#)

### 9.1 Development Tools

Several development tools are available to assist in the design and evaluation of the MCP48CXDX1/2 devices. The currently available tools display in the table below.

Figure 9-1 shows how the ADM00309 bond-out PCB can be populated to easily evaluate the MCP48CXDX1/2 devices. Device evaluation can use the PICKit™ Serial Analyzer to control the DAC output registers and state of the Configuration, Control and Status registers.

The ADM00309 boards may be purchased directly from the Microchip website at [www.microchip.com](http://www.microchip.com).

### 9.2 Technical Documentation

Several additional technical documents for design and development are available. These technical documents include Application Notes, Technical Briefs and Design Guides. Table 9-2 lists some of these documents.

**TABLE 9-1: DEVELOPMENT TOOLS (Note 1)**

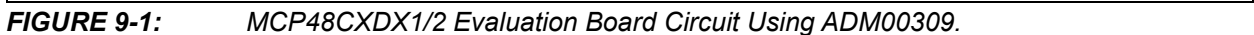
| Board Name                          | Part #   | Comment                                                                                                                                                                                                                                                                                    |
|-------------------------------------|----------|--------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|
| MSOP-8 and MSOP-10 Evaluation Board | ADM00309 | The MSOP-10 and MSOP-8 Evaluation Board is a bond-out board that allows the system designer to quickly evaluate the operation of Microchip Technology's devices in any of the following packages: <ul style="list-style-type: none"><li>• MSOP (8/10-pin)</li><li>• DIP (10-pin)</li></ul> |

**Note 1:** Supports the PICKit™ Serial Analyzer. See the User's Guide for additional information and requirements.

**TABLE 9-2: TECHNICAL DOCUMENTATION**

| Application Note Number | Title                                                                      | Literature # |
|-------------------------|----------------------------------------------------------------------------|--------------|
| AN1326                  | Using the MCP4728 12-Bit DAC for LDMOS Amplifier Bias Control Applications | DS01326      |
| —                       | Signal Chain Design Guide                                                  | DS21825      |
| —                       | Analog Solutions for Automotive Applications Design Guide                  | DS01005      |

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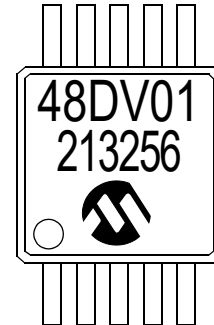
## 10.0 PACKAGING INFORMATION

### 10.1 Package Marking Information

10-Lead MSOP



Example

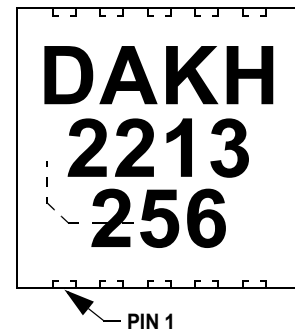


10-Lead DFN (3 x 3 mm)



| Part Number     | Code |
|-----------------|------|
| MCP48CVD01-E/MF | DAKH |
| MCP48CVD11-E/MF | DAKL |
| MCP48CVD21-E/MF | DAKN |
| MCP48CVD02-E/MF | DAKJ |
| MCP48CVD12-E/MF | DAKM |
| MCP48CVD22-E/MF | DAKP |
| MCP48CMD01-E/MF | DAKD |
| MCP48CMD11-E/MF | DAKD |
| MCP48CMD21-E/MF | DAKF |
| MCP48CMD02-E/MF | DAKC |
| MCP48CMD12-E/MF | DAKE |
| MCP48CMD22-E/MF | DAKG |

Example

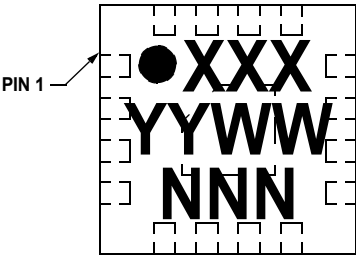


|                |        |                                                                                                                  |
|----------------|--------|------------------------------------------------------------------------------------------------------------------|
| <b>Legend:</b> | XX...X | Customer-specific information                                                                                    |
|                | Y      | Year code (last digit of calendar year)                                                                          |
|                | YY     | Year code (last 2 digits of calendar year)                                                                       |
|                | WW     | Week code (week of January 1 is week '01')                                                                       |
|                | NNN    | Alphanumeric traceability code                                                                                   |
|                | (e3)   | Pb-free JEDEC designator for Matte Tin (Sn)                                                                      |
|                | *      | This package is Pb-free. The Pb-free JEDEC designator (e3) can be found on the outer packaging for this package. |

**Note:** In the event the full Microchip part number cannot be marked on one line, it will be carried over to the next line, thus limiting the number of available characters for customer-specific information.

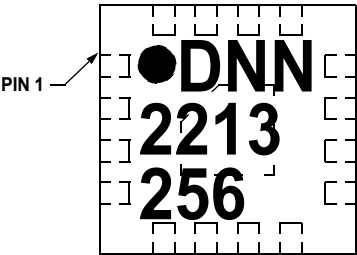
# MCP48CXDX1/2

16-Lead QFN (3 x 3 mm)



| Part Number     | Code |
|-----------------|------|
| MCP48CVD01-E/MG | DNN  |
| MCP48CVD11-E/MG | DPP  |
| MCP48CVD21-E/MG | DRR  |
| MCP48CVD02-E/MG | DAK  |
| MCP48CVD12-E/MG | DAL  |
| MCP48CVD22-E/MG | DAM  |
| MCP48CMD01-E/MG | DKK  |
| MCP48CMD11-E/MG | DLL  |
| MCP48CMD21-E/MG | DMM  |
| MCP48CMD02-E/MG | DAG  |
| MCP48CMD12-E/MG | DAH  |
| MCP48CMD22-E/MG | DAJ  |

Example





# MCP48CXDX1/2

## 10-Lead Plastic Micro Small Outline Package (UN) - 3x3 mm Body [MSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



| Units                    |    | MILLIMETERS |      |      |
|--------------------------|----|-------------|------|------|
| Dimension Limits         |    | MIN         | NOM  | MAX  |
| Number of Pins           | N  | 10          |      |      |
| Pitch                    | e  | 0.50 BSC    |      |      |
| Overall Height           | A  | -           | -    | 1.10 |
| Molded Package Thickness | A2 | 0.75        | 0.85 | 0.95 |
| Standoff                 | A1 | 0.00        | -    | 0.15 |
| Overall Width            | E  | 4.90 BSC    |      |      |
| Molded Package Width     | E1 | 3.00 BSC    |      |      |
| Overall Length           | D  | 3.00 BSC    |      |      |
| Foot Length              | L  | 0.40        | 0.60 | 0.80 |
| Footprint                | L1 | 0.95 REF    |      |      |
| Foot Angle               | Ø  | 0°          | -    | 8°   |
| Mold Draft Angle         | Ø1 | 5°          | -    | 15°  |
| Lead Thickness           | c  | 0.08        | -    | 0.23 |
| Lead Width               | b  | 0.15        | -    | 0.33 |

### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Dimensions D and E1 do not include mold flash or protrusions. Mold flash or protrusions shall not exceed 0.15mm per side.
- Dimensioning and tolerancing per ASME Y14.5M.  
BSC: Basic Dimension. Theoretically exact value shown without tolerances.  
REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-021-UN Rev F Sheet 2 of 2

## 10-Lead Plastic Micro Small Outline Package (UN) - 3x3 mm Body [MSOP]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



### RECOMMENDED LAND PATTERN

| Units                      |    | MILLIMETERS |      |      |
|----------------------------|----|-------------|------|------|
| Dimension Limits           |    | MIN         | NOM  | MAX  |
| Contact Pitch              | E  | 0.50 BSC    |      |      |
| Contact Pad Spacing        | C  |             | 4.40 |      |
| Overall Width              | Z  |             |      | 5.80 |
| Contact Pad Width (X10)    | X1 |             |      | 0.30 |
| Contact Pad Length (X10)   | Y1 |             |      | 1.40 |
| Distance Between Pads (X5) | G1 | 3.00        |      |      |
| Distance Between Pads (X8) | G  | 0.20        |      |      |

#### Notes:

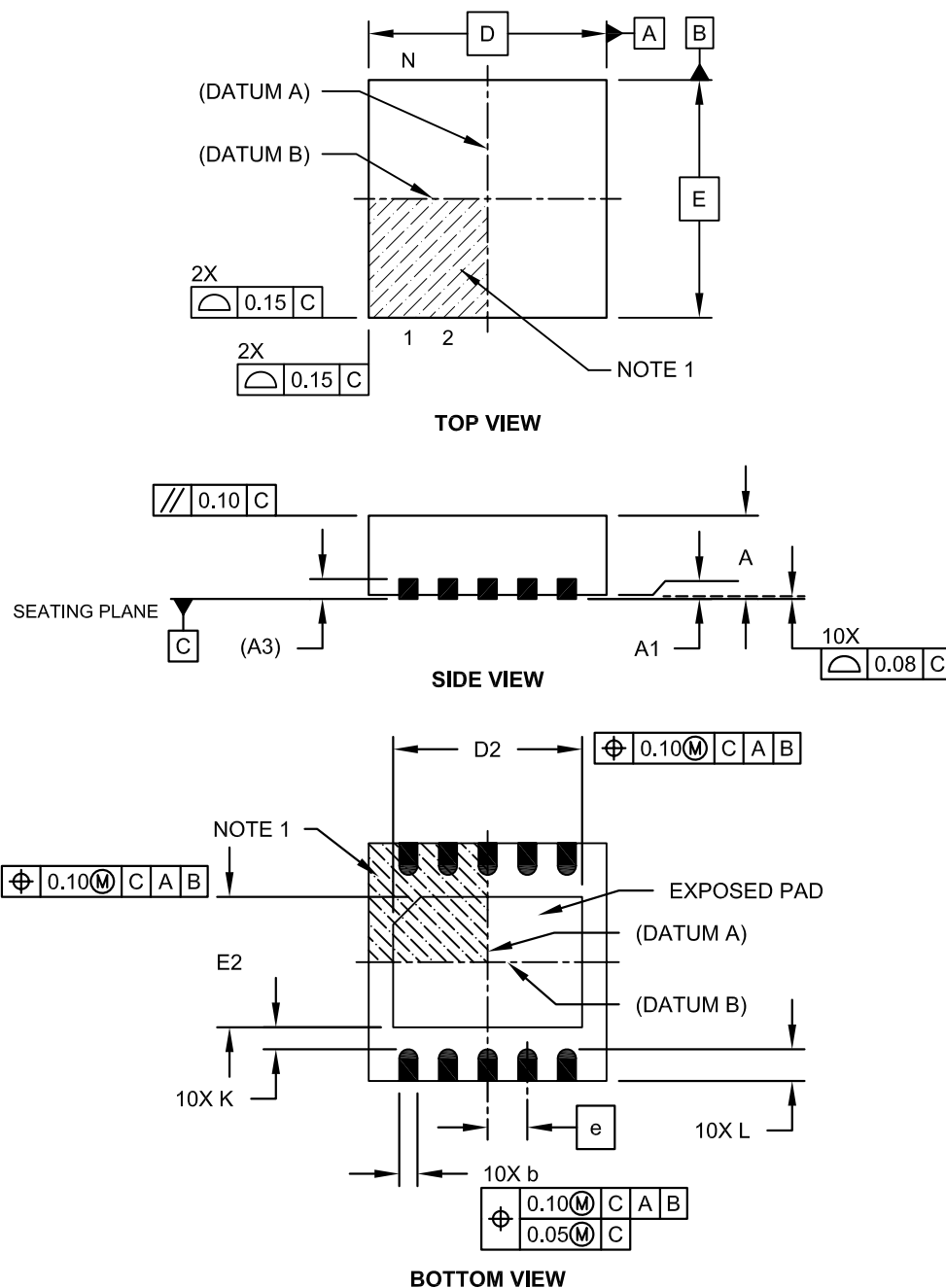
- Dimensioning and tolerancing per ASME Y14.5M  
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2021-UN Rev F

# MCP48CXDX1/2

## 10-Lead Plastic Dual Flat, No Lead Package (MF) - 3x3x0.9mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



Microchip Technology Drawing No. C04-063C Sheet 1 of 2

## 10-Lead Plastic Dual Flat, No Lead Package (MF) - 3x3x0.9mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



| Dimension Limits       | Units | MILLIMETERS |      |      |
|------------------------|-------|-------------|------|------|
|                        |       | MIN         | NOM  | MAX  |
| Number of Pins         | N     | 10          |      |      |
| Pitch                  | e     | 0.50 BSC    |      |      |
| Overall Height         | A     | 0.80        | 0.90 | 1.00 |
| Standoff               | A1    | 0.00        | 0.02 | 0.05 |
| Contact Thickness      | A3    | 0.20 REF    |      |      |
| Overall Length         | D     | 3.00 BSC    |      |      |
| Exposed Pad Length     | D2    | 2.15        | 2.35 | 2.45 |
| Overall Width          | E     | 3.00 BSC    |      |      |
| Exposed Pad Width      | E2    | 1.40        | 1.50 | 1.75 |
| Contact Width          | b     | 0.18        | 0.25 | 0.30 |
| Contact Length         | L     | 0.30        | 0.40 | 0.50 |
| Contact-to-Exposed Pad | K     | 0.20        | -    | -    |

### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Package may have one or more exposed tie bars at ends.
- Package is saw singulated.
- Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing No. C04-063C Sheet 2 of 2

# MCP48CXDX1/2

## 10-Lead Plastic Dual Flat, No Lead Package (MF) - 3x3x0.9mm Body [DFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

| Units                      |    | MILLIMETERS |      |      |
|----------------------------|----|-------------|------|------|
| Dimension Limits           |    | MIN         | NOM  | MAX  |
| Contact Pitch              | E  | 0.50 BSC    |      |      |
| Optional Center Pad Width  | W2 |             |      | 2.48 |
| Optional Center Pad Length | T2 |             |      | 1.55 |
| Contact Pad Spacing        | C1 |             | 3.10 |      |
| Contact Pad Width (X10)    | X1 |             |      | 0.30 |
| Contact Pad Length (X10)   | Y1 |             |      | 0.65 |
| Distance Between Pads      | G  | 0.20        |      |      |

Notes:

1. Dimensioning and tolerancing per ASME Y14.5M  
BSC: Basic Dimension. Theoretically exact value shown without tolerances.

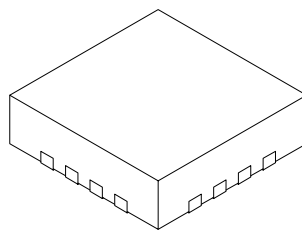
Microchip Technology Drawing No. C04-2063B



# MCP48CXDX1/2

## 16-Lead Plastic Quad Flat, No Lead Package (MG) - 3x3x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



| Units                  |    | MILLIMETERS |      |      |
|------------------------|----|-------------|------|------|
| Dimension Limits       |    | MIN         | NOM  | MAX  |
| Number of Pins         | N  | 16          |      |      |
| Pitch                  | e  | 0.50 BSC    |      |      |
| Overall Height         | A  | 0.80        | 0.85 | 0.90 |
| Standoff               | A1 | 0.00        | 0.02 | 0.05 |
| Contact Thickness      | A3 | 0.20 REF    |      |      |
| Overall Width          | E  | 3.00 BSC    |      |      |
| Exposed Pad Width      | E2 | 1.00        | 1.10 | 1.50 |
| Overall Length         | D  | 3.00 BSC    |      |      |
| Exposed Pad Length     | D2 | 1.00        | 1.10 | 1.50 |
| Contact Width          | b  | 0.18        | 0.25 | 0.30 |
| Contact Length         | L  | 0.25        | 0.35 | 0.45 |
| Contact-to-Exposed Pad | K  | 0.20        | -    | -    |

### Notes:

- Pin 1 visual index feature may vary, but must be located within the hatched area.
- Package is saw singulated.
- Dimensioning and tolerancing per ASME Y14.5M.
  - BSC: Basic Dimension. Theoretically exact value shown without tolerances.
  - REF: Reference Dimension, usually without tolerance, for information purposes only.

Microchip Technology Drawing C04-142A Sheet 2 of 2

## 16-Lead Plastic Quad Flat, No Lead Package (MG) – 3x3x0.9 mm Body [QFN]

**Note:** For the most current package drawings, please see the Microchip Packaging Specification located at <http://www.microchip.com/packaging>



RECOMMENDED LAND PATTERN

| Units                      |        | MILLIMETERS |      |      |
|----------------------------|--------|-------------|------|------|
| Dimension                  | Limits | MIN         | NOM  | MAX  |
| Contact Pitch              | E      | 0.50 BSC    |      |      |
| Optional Center Pad Width  | W2     |             |      | 1.20 |
| Optional Center Pad Length | T2     |             |      | 1.20 |
| Contact Pad Spacing        | C1     |             | 2.90 |      |
| Contact Pad Spacing        | C2     |             | 2.90 |      |
| Contact Pad Width (X16)    | X1     |             |      | 0.30 |
| Contact Pad Length (X16)   | Y1     |             |      | 0.80 |
| Distance Between Pads      | G      | 0.20        |      |      |

**Notes:**

1. Dimensioning and tolerancing per ASME Y14.5M

BSC: Basic Dimension. Theoretically exact value shown without tolerances.

Microchip Technology Drawing No. C04-2142A

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NOTES:

## APPENDIX A: REVISION HISTORY

### Revision A (April 2022)

- Original release of this document

**Note:** The SPI standard uses the terminology “Master” and “Slave”. The equivalent Microchip terminology used in this document is “Host” and “Client”, respectively.

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NOTES:

## APPENDIX B: TERMINOLOGY

### B.1 Resolution

The resolution is the number of DAC output states that divide the full-scale range. For the 12-bit DAC, the resolution is  $2^{12}$ , meaning the DAC code ranges from 0 to 4095.

**Note:** When there are  $2^N$  resistors in the resistor ladder and  $2^N$  tap points, the full-scale DAC register code is the resistor element (1 LSb) from the source reference voltage ( $V_{DD}$  or  $V_{REF}$ ).

### B.2 Least Significant Bit (LSb)

This is the voltage difference between two successive codes. For a given output voltage range, it is divided by the resolution of the device (see the equation below). The range may be  $V_{DD}$  (or  $V_{REF}$ ) to  $V_{SS}$  (ideal), the DAC register codes across the linear range of the output driver (Measured 1) or full-scale to zero-scale (Measured 2).

#### EQUATION B-1: LSb VOLTAGE CALCULATION

**Ideal:**

$$V_{LSb(IDEAL)} = \frac{V_{DD}}{2^N} \text{ or } \frac{V_{REF}}{2^N}$$

**Measured 1 (12-bit device):**

$$V_{LSb(Measured)} = \frac{V_{OUT(@4032)} - V_{OUT(@64)}}{(4032 - 64)}$$

**Measured 2:**

$$V_{LSb} = \frac{V_{OUT(@FS)} - V_{OUT(@ZS)}}{2^N - 1}$$

$$\begin{aligned} 2^N &= 4096 \text{ (MCP48CXDX2X)} \\ &= 1024 \text{ (MCP48CXDX1X)} \\ &= 256 \text{ (MCP48CXDX0X)} \end{aligned}$$

### B.3 Monotonic Operation

The monotonic operation means that the device's output voltage ( $V_{OUT}$ ) increases with every 1 code step (LSb) increment (from  $V_{SS}$  to the DAC's reference voltage ( $V_{DD}$  or  $V_{REF}$ )).



**FIGURE B-1:**  $V_W$  ( $V_{OUT}$ ).

### B.4 Full-Scale Error ( $E_{FS}$ )

The full-scale error (see Figure B-3) is the error on the  $V_{OUT}$  pin relative to the expected  $V_{OUT}$  voltage (theoretical) for the maximum device DAC register code (code FFFh for 12-bit, code 3FFh for 10-bit and code FFh for 8-bit) (see the equation below). The error depends on the resistive load on the  $V_{OUT}$  pin (and where that load is tied to, such as  $V_{SS}$  or  $V_{DD}$ ). For loads (to  $V_{SS}$ ) greater than specified, the full-scale error will be greater.

The error in bits is determined by the theoretical voltage step size to give an error in LSb.

#### EQUATION B-2: FULL-SCALE ERROR

$$E_{FS} = \frac{V_{OUT(@FS)} - V_{IDEAL(@FS)}}{V_{LSb(IDEAL)}}$$

Where:

$E_{FS}$  is expressed in LSb.

$V_{OUT(@FS)}$  = the  $V_{OUT}$  voltage when the DAC register code is at full scale

$V_{IDEAL(@FS)}$  = the ideal output voltage when the DAC register code is at full scale

$V_{LSb(IDEAL)}$  = the theoretical voltage step size

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## B.5 Zero-Scale Error ( $E_{ZS}$ )

The zero-scale error (see [Figure B-2](#)) is the difference between the ideal and measured  $V_{OUT}$  voltage with the DAC register code equal to 000h (see the equation below). The error depends on the resistive load on the  $V_{OUT}$  pin (and where that load is tied to, such as  $V_{SS}$  or  $V_{DD}$ ). For loads (to  $V_{DD}$ ) greater than specified, the zero-scale error is greater.

The error in bits is determined by the theoretical voltage step size to give an error in LSb.

### EQUATION B-3: ZERO SCALE ERROR

$$E_{ZS} = \frac{V_{OUT(@ZS)}}{V_{LSb(IDEAL)}}$$

Where:

$E_{ZS}$  is expressed in LSb.

$V_{OUT(@ZS)}$  = the  $V_{OUT}$  voltage when the DAC register code is at zero scale

$V_{LSb(IDEAL)}$  = the theoretical voltage step size

## B.6 Total Unadjusted Error ( $E_T$ )

The total unadjusted error ( $E_T$ ) is the difference between the ideal and the measured  $V_{OUT}$  voltage. Typically, calibration of the output voltage is implemented to improve the system's performance.

The error in bits is determined by the theoretical voltage step size to give an error in LSb.

The equation below shows the total unadjusted error calculation

### EQUATION B-4: TOTAL UNADJUSTED ERROR CALCULATION

$$E_T = \frac{(V_{OUT\_Actual(@code)} - V_{OUT\_Ideal(@code)})}{V_{LSb(Ideal)}}$$

Where:

$E_T$  is expressed in LSb.

$V_{OUT\_Actual(@code)}$  = the measured DAC output voltage at the specified code

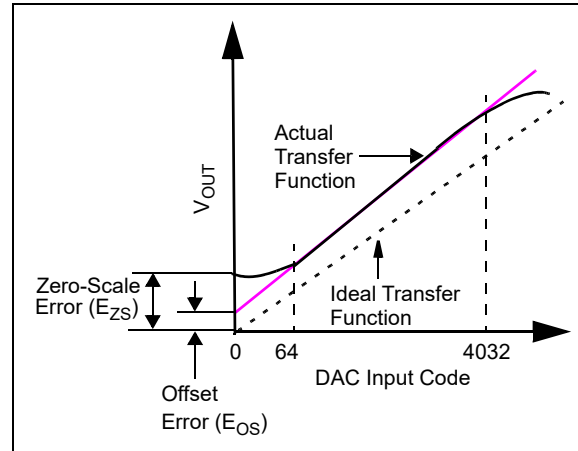
$V_{OUT\_Ideal(@code)}$  = the calculated DAC output voltage at the specified code  
(code \*  $V_{LSb(Ideal)}$ )

$V_{LSb(Ideal)}$  =  $V_{REF}/\# \text{ Steps}$   
12-bit =  $V_{REF}/4096$   
10-bit =  $V_{REF}/1024$   
8-bit =  $V_{REF}/256$

## B.7 Offset Error ( $E_{OS}$ )

The offset error is the delta voltage of the  $V_{OUT}$  voltage from the ideal output voltage at the specified code. This code is specified where the output amplifier is in the linear operating range; for the MCP48CXDX1/2, we specify code 64 (decimal). The offset error does not include gain error, which is illustrated in the figure below.

This error is expressed in mV. The offset error can be negative or positive. The error can be calibrated by software in application circuits.



**FIGURE B-2: OFFSET ERROR (ZERO GAIN ERROR).**

## B.8 Offset Error Drift ( $E_{OSD}$ )

The offset error drift is the variation in offset error due to a change in ambient temperature. The offset error drift is typically expressed in ppm/°C or  $\mu V/^\circ C$ .

## B.9 Gain Error ( $E_G$ )

Gain error is a calculation based on the ideal slope using the voltage boundaries for the linear range of the output driver (e.g., code 64 and code 4032) (see the figure below). The gain error calculation nullifies the device's offset error.

The gain error indicates how well the slope of the actual transfer function matches the slope of the ideal transfer function. The gain error is usually expressed as a percentage of full-scale range (% of FSR) or in LSb. FSR is the ideal full-scale voltage of the DAC (see the equation below).



**FIGURE B-3: GAIN ERROR AND FULL-SCALE ERROR EXAMPLE.**

## EQUATION B-5: GAIN ERROR EXAMPLE

$$E_G = \frac{(V_{OUT(@4032)} - V_{OS} - V_{OUT\_Ideal(@4032)})}{V_{Full-Scale\ Range}} * 100$$

Where:

$E_G$  is expressed in % of FSR.

- $V_{OUT(@4032)}$  = the measured DAC output voltage at the specified code
- $V_{OUT\_Ideal(@4032)}$  = the calculated DAC output voltage at the specified code ( $4032 * V_{LSb(Ideal)}$ )
- $V_{OS}$  = measured offset voltage
- $V_{Full-Scale\ Range}$  = Expected full-scale output value (such as the  $V_{REF}$  voltage)

## B.10 Gain Error Drift ( $E_{GD}$ )

The gain error drift is the variation in gain error due to a change in ambient temperature. The gain error drift is typically expressed in ppm/°C (of FSR).

## B.11 Integral Nonlinearity (INL)

The integral nonlinearity (INL) error is the maximum deviation of an actual transfer function from an ideal transfer function (straight line) passing through the defined end-points of the DAC transfer function (after offset and gain errors are removed).

For the MCP48CXDX1/2, INL is calculated using the defined end-points, DAC code 64 and code 4032. INL can be expressed as a percentage of FSR or in LSb. INL is also called relative accuracy. The equation below shows how to calculate the INL error in LSb, and the figure below shows an example of INL accuracy.

Positive INL means a  $V_{OUT}$  voltage higher than the ideal one. Negative INL means a  $V_{OUT}$  voltage lower than the ideal one.

## EQUATION B-6: INL ERROR

$$E_{INL} = \frac{(V_{OUT} - V_{Calc\_Ideal})}{V_{LSb(Measured)}}$$

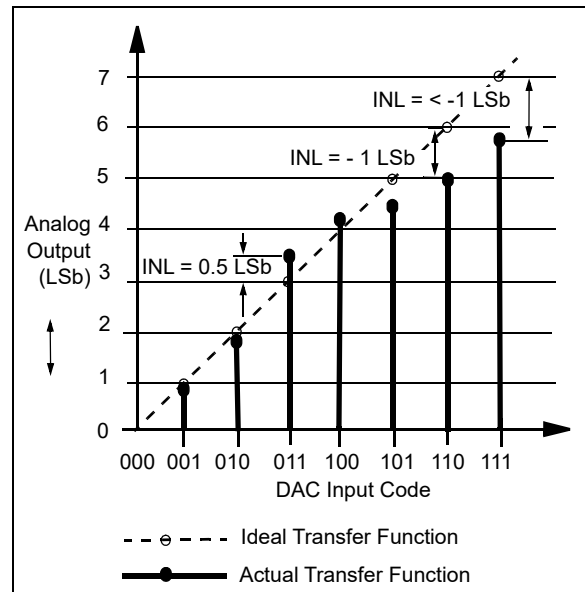
Where:

INL is expressed in LSb.

$$V_{Calc\_Ideal} = \text{Code} * V_{LSb(Measured)} + V_{OS}$$

$V_{OUT}$  = the measured DAC output voltage with a given DAC register code

$$V_{LSb(Measured)} = \text{for measured: } (V_{OUT(4032)} - V_{OUT(64)})/3968$$



**FIGURE B-4: INL ACCURACY.**

## B.12 Differential Nonlinearity (DNL)

The differential nonlinearity (DNL) error (see [Figure B-5](#)) is the measure of step size between codes in actual transfer function. The ideal step size between codes is 1 LSB. A DNL error of zero would imply that every code is exactly 1 LSB wide. If the DNL Error is less than 1 LSB, the DAC guarantees monotonic output and no missing codes. The equation below shows how to calculate the DNL Error between any two adjacent codes in LSB.

### EQUATION B-7: DNL ERROR

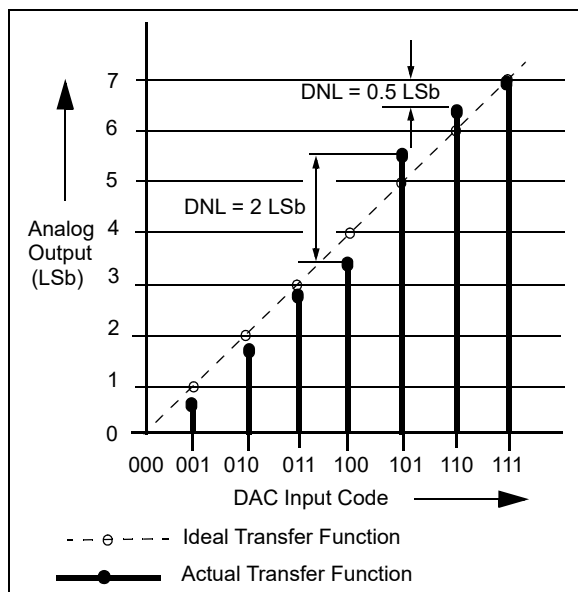
$$E_{DNL} = \frac{(V_{OUT(code=n+1)} - V_{OUT(code=n)})}{V_{LSb(Measured)}} - 1$$

Where:

DNL is expressed in LSB.

$V_{OUT(Code = n)}$  = the measured DAC output voltage with a given DAC register code

$V_{LSb(Measured)}$  = for measured:  
( $V_{OUT(4032)} - V_{OUT(64)}/3968$ )



**FIGURE B-5: DNL ACCURACY.**

## B.13 Settling Time

The settling time is the time delay required for the  $V_{OUT}$  voltage to settle into its new output value. This time is measured from the start of the code transition to when the  $V_{OUT}$  voltage is within the specified accuracy.

For the MCP48CXDX1/2, the settling time is a measure of the time delay until the  $V_{OUT}$  voltage reaches within 0.5 LSB of its final value, when the volatile DAC register changes from 1/4 to 3/4 of the FSR (12-bit device: 400h to C00h).

## B.14 Major-Code Transition Glitch

Major-Code transition glitch is the impulse energy injected into the DAC analog output when the code in the DAC register changes the state. It is normally specified as the area of the glitch in nV-Sec and is measured when the digital code is changed by 1 LSB at the major carry transition (Example: 011...111 to 100...000 or 100...000 to 011...111).

## B.15 Digital Feed-Through

The digital feed-through is the glitch that appears at the analog output caused by coupling from the digital input pins of the device. The area of the glitch is expressed in nV-Sec and is measured with a full-scale change (example: all 0s to all 1s and vice versa) on the digital input pins. The digital feed-through is measured when the DAC is not being written to the output register.

## B.16 -3 dB Bandwidth

This is the frequency of the signal at the  $V_{REF}$  pin that causes the voltage at the  $V_{OUT}$  pin to fall to -3 dB from a static value on the  $V_{REF}$  pin. The output decreases due to the RC characteristics of the resistor ladder and the characteristics of the output buffer.

## B.17 Power-Supply Sensitivity (PSS)

PSS indicates how the output of the DAC is affected by changes in the supply voltage. PSS is the ratio of the change in  $V_{OUT}$  to a change in  $V_{DD}$  for mid-scale output of the DAC. The  $V_{OUT}$  is measured while the  $V_{DD}$  is varied from 5.5V to 2.7V as a step ( $V_{REF}$  voltage held constant) and expressed in %/%, which is the % change of the DAC output voltage with respect to the % change of the  $V_{DD}$  voltage.

### EQUATION B-8: PSS CALCULATION

$$PSS = \frac{(V_{OUT(@5.5V)} - V_{OUT(@2.7V)})/V_{OUT(@5.5V)}}{(5.5V - 2.7V)/(5.5V)}$$

Where:

PSS is expressed in %/ %.

$V_{OUT(@5.5V)}$  = the measured DAC output voltage with  $V_{DD} = 5.5V$

$V_{OUT(@2.7V)}$  = the measured DAC output voltage with  $V_{DD} = 2.7V$

## B.18 Power-Supply Rejection Ratio (PSRR)

PSRR indicates how the output of the DAC is affected by changes in the supply voltage. PSRR is the ratio of the change in  $V_{OUT}$  to a change in  $V_{DD}$  for full-scale output of the DAC. The  $V_{OUT}$  is measured while the  $V_{DD}$  is varied  $\pm 10\%$  ( $V_{REF}$  voltage held constant) and expressed in dB or  $\mu V/V$ .

## B.19 $V_{OUT}$ Temperature Coefficient

The  $V_{OUT}$  temperature coefficient quantifies the error in the resistor ladder's resistance ratio (DAC register code value) and the output buffer due to temperature drift.

## B.20 Absolute Temperature Coefficient

The absolute temperature coefficient quantifies the error in the end-to-end output voltage (nominal output voltage  $V_{OUT}$ ) due to temperature drift. For a DAC, this error is typically not an issue due to the ratiometric aspect of the output.

## B.21 Noise Spectral Density

The noise spectral density is a measurement of the device's internally generated random noise and is characterized as a spectral density (voltage per  $\sqrt{\text{Hz}}$ ). It is measured by loading the DAC to the mid-scale value and measuring the noise at the  $V_{OUT}$  pin. It is measured in  $nV/\sqrt{\text{Hz}}$ .

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NOTES:

## PRODUCT IDENTIFICATION SYSTEM

To order or obtain information, e.g., on pricing or delivery, refer to the factory or the listed sales office.

| <u>PART NO.</u>                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                      | <u>X</u> <sup>(1)</sup> | — | <u>X</u>          | <u>/XX</u> |
|------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------------|-------------------------|---|-------------------|------------|
| Device                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                               | Tape and Reel           |   | Temperature Range | Package    |
| <b>Device:</b> MCP48CXDX1/2:1 LSb INL Voltage Output Digital-to-Analog Converters, with SPI Interface, 8/10/12-bit Resolution, Single/Dual Outputs and Volatile/MTP Memory                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                           |                         |   |                   |            |
| <b>Tape and Reel:</b> T = Tape and Reel                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                              |                         |   |                   |            |
| <b>Temperature Range:</b> E = -40°C to +125°C (Extended)                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                             |                         |   |                   |            |
| <b>Package:</b> MF = Plastic Dual Flat, No Lead Package (DFN), 3 x 3 x 0.9 mm, 10-Lead<br>MG = Plastic Quad Flat, No Lead Package (QFN), 3 x 3 x 0.9 mm, 16-Lead<br>UN = Plastic Micro Small Outline Package (MSOP), 10-Lead                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                                         |                         |   |                   |            |
| <b>Examples:</b><br>a) MCP48CVD01-E/MF: 1 LSb INL Voltage Output Digital-to-Analog Converter, 8-bit Resolution, Extended Temperature, 10LD DFN, with volatile memory.<br>b) MCP48CVD01T-E/MF: 1 LSb INL Voltage Output Digital-to-Analog Converter, 8-bit Resolution, Tape and Reel, Extended Temperature, 10LD DFN, with volatile memory<br>a) MCP48CVD12-E/MG: 1 LSb INL Voltage Output Digital-to-Analog Converter, 10-bit Resolution, Extended Temperature, 16LD QFN, with volatile memory.<br>b) MCP48CVD12T-E/MG: 1 LSb INL Voltage Output Digital-to-Analog Converter, 10-bit Resolution, Tape and Reel, Extended Temperature, 16LD QFN, with volatile memory.<br>a) MCP48CMD21-E/UN: 1 LSb INL Voltage Output Digital-to-Analog Converter, 12-bit Resolution, Extended Temperature, 10LD MSOP, with nonvolatile memory.<br>b) MCP48CMD21T-E/UN: 1 LSb INL Voltage Output Digital-to-Analog Converter, 12-bit Resolution, Tape and Reel, Extended Temperature, 10LD MSOP, with nonvolatile memory.<br><br><b>Note 1:</b> Tape and Reel identifier only appears in the catalog part number description. This identifier is used for ordering purposes and is not printed on the device package. Check with your Microchip Sales Office for package availability with the Tape and Reel option. |                         |   |                   |            |

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NOTES:

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